

Chapter 8

Plane Electromagnetic Waves

Ex. 1 In a source-free simple medium,

$$\nabla \cdot (\nabla \times \mathbf{A}) = \nabla \cdot \mathbf{B} = \nabla \cdot \nabla \times \mathbf{A} = \nabla \cdot \mathbf{E} + \epsilon \nabla \cdot \nabla \times \mathbf{A} \quad (1)$$

$$\nabla \cdot (\nabla \times \mathbf{A}) = \nabla \cdot \nabla \times \mathbf{E} = \nabla \cdot \nabla \times \mathbf{E} = \nabla \cdot \nabla \times \mathbf{E} = \nabla \cdot \nabla \times \mathbf{E} \quad (2)$$

Substituting (1) in (2) and noting that $\nabla \cdot \mathbf{E} = 0$:

$$\nabla^2 \mathbf{E} = \nabla \times \nabla \times \mathbf{E} = \nabla \times \nabla \times \mathbf{E} = 0$$

Similarly for \mathbf{B} .

Ex. 2 Assume that the velocity vector with a velocity u is in the z -direction, which is the direction of propagation of the incident wave.

$$(1) \quad \mathbf{E}_i = \mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}, \quad \mathbf{E}_r = \mathbf{E}_0 e^{i(\mathbf{k}' \cdot \mathbf{r} - \omega' t)}$$

$\mathbf{E}_i + \mathbf{E}_r = 0$ must be satisfied on reflecting surface for all t

and \mathbf{r} :

$$(u \cdot \mathbf{k} - \omega) e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)} = (u \cdot \mathbf{k}' - \omega') e^{i(\mathbf{k}' \cdot \mathbf{r} - \omega' t)}$$

$$\implies \omega' - \omega = -(u \cdot \mathbf{k}' - \omega') = -(\mathbf{k}' \cdot \mathbf{u} - \omega') = -(\omega' - u \cdot \mathbf{k}') \implies \omega' = \omega$$

$$\implies \frac{\omega'}{\omega} = 1 = \frac{c}{c} \left(1 + \frac{u \cdot \mathbf{k}'}{c \omega'}\right)$$

$$\implies \frac{\omega'}{\omega} = \frac{c}{c} = \frac{1}{1 - \frac{u \cdot \mathbf{k}'}{c \omega'}} \implies \omega' = \omega \left(1 - \frac{u \cdot \mathbf{k}'}{c \omega'}\right) \text{ for wave.}$$

$$\implies \omega' = \omega \left(1 - \frac{u \cdot \mathbf{k}'}{c \omega'}\right)$$

(2) For $\omega' = \omega$, $\mathbf{k}' = \mathbf{k} - \frac{2\mathbf{u} \cdot \mathbf{k}}{c} \mathbf{u}$

Ex. 3 Harmonic time dependence: $e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}$

$$\text{Assume: } \mathbf{E} = \mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}, \quad \mathbf{B} = \mathbf{B}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}, \text{ where } \mathbf{E}_0 \text{ and } \mathbf{B}_0 \text{ are constant vectors.}$$

$$\text{Max: } \nabla \cdot (\mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}) = \mathbf{k} \cdot \mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)} = \mathbf{k} \cdot \mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)} = \mathbf{k} \cdot \mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}$$

$$\text{Max: } \nabla \cdot (\mathbf{B}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}) = \mathbf{k} \cdot \mathbf{B}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)} = \mathbf{k} \cdot \mathbf{B}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}$$

$$\text{Max: } \nabla \cdot (\mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}) = \mathbf{k} \cdot \mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)} = \mathbf{k} \cdot \mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}$$

$$\text{Max: } \nabla \cdot (\mathbf{B}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}) = \mathbf{k} \cdot \mathbf{B}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)} = \mathbf{k} \cdot \mathbf{B}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}$$

$$\text{Max: } \nabla \cdot (\mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}) = \mathbf{k} \cdot \mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)} = \mathbf{k} \cdot \mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}$$

Ex. 2.1. Let $\vec{r} = \vec{r}_0 + \omega r^2 \cos(\omega t) \hat{i} + \vec{r}_0 + \frac{\pi}{2}$ (2.1.1).

(a) $\vec{r}_0 = \omega \sqrt{2} \hat{i} = \omega \frac{\sqrt{2}}{2} \hat{i} = \frac{\omega}{\sqrt{2}} \hat{i} = \omega \cos \alpha \hat{i}$

$\alpha = \cos^{-1} \frac{1}{\sqrt{2}} = 45^\circ$

At $t = \pi/2\omega$, we require the argument of \vec{r} to be

$$\omega^2 \cos \omega t = \frac{\omega}{\sqrt{2}} \cos \frac{\pi}{2} + \frac{\pi}{2} = 0 + \frac{\pi}{2} = \frac{\pi}{2} \Rightarrow \alpha = \frac{\pi}{4} \dots$$

$$\Rightarrow \gamma = \frac{\pi}{4} + \frac{\pi}{4} = \frac{\pi}{2} = 90^\circ \text{ (a)}$$

Ex. 2.2. Show: $\vec{r} = \vec{r}_0 e^{i\omega t} + \vec{r}_1 e^{-i\omega t}$ (2.2.1)

(a) $\omega = \omega^2$ (rad/s) $\Rightarrow \beta = \omega^2 \hat{i} = \omega \cos \alpha \hat{i}$

$\beta = \omega \cos \alpha$ (rad/s) $\Rightarrow \alpha = \cos^{-1} \frac{\omega}{\omega} = 0^\circ$

(b) $\omega = \frac{\omega}{2} \Rightarrow \alpha = \left(\frac{\omega}{2}\right)^{-1} = 2$

(c) Left-hand elliptically polarized.

(d) $\gamma = \frac{\pi}{4} - \frac{\pi}{4} = 0^\circ$ (2.2.2)

$$\vec{r} = \frac{1}{2} \vec{r}_0 e^{i\omega t} + \frac{\sqrt{3}}{2} \vec{r}_1 e^{i(\omega t - \pi/3)} - \vec{r}_1 e^{-i\omega t}$$

$$\vec{r}(t) = \frac{\sqrt{3}}{2} \left[\vec{r}_0 \cos(\omega t - \pi/3) + \vec{r}_1 \cos(\omega t - \pi/3) - \cos \omega t \right] \quad (2.2.3)$$

Ex. 2.3 Let $\vec{r} = \vec{r}_0 \cos \omega t + \vec{r}_1 \sin \omega t$ in $\vec{r} = \vec{r}_0 \cos \omega t + \vec{r}_1 \sin \omega t$

$$\frac{\vec{r}}{\omega} = \vec{r}_0 \cos \omega t, \quad \frac{\dot{\vec{r}}}{\omega} = \vec{r}_1 \cos \omega t = \vec{r}_0 \cos \omega t + \vec{r}_1 \sin \omega t$$

$$\frac{\dot{\vec{r}}}{\omega} = \frac{\vec{r}}{\omega} \cos \omega t + \sqrt{1 - \left(\frac{\vec{r}}{\omega}\right)^2} \sin \omega t$$

$$\left(\frac{\dot{\vec{r}}}{\omega} - \frac{\vec{r}}{\omega} \cos \omega t\right) = \left(1 - \left(\frac{\vec{r}}{\omega}\right)^2\right)^{1/2} \sin \omega t$$

$$\left(\frac{\dot{\vec{r}}}{\omega} \cos \omega t\right)^2 + \left(\frac{\dot{\vec{r}}}{\omega} \sin \omega t\right)^2 = \left(\frac{\dot{\vec{r}}}{\omega}\right)^2 \sin^2 \omega t = 1 \quad (2.3.1)$$

which is the equation of an ellipse. In order to find the

parameters of the polarization

ellipse, rotate the coordinate

axes by any counterclockwise by

an angle θ to θ' . Assume the

equation of the ellipse in terms

of the new coordinates to be

$$\left(\frac{x'}{a}\right)^2 + \left(\frac{y'}{b}\right)^2 = 1 \quad (2.3.2)$$



$$\text{where } E_x = E_0 \cos \theta - E_1 \sin \theta, \quad (3)$$

$$\text{and } E_y = -E_0 \sin \theta + E_1 \cos \theta, \quad (4)$$

Substituting (3) and (4) in (2) and rearranging

$$E_0 \left(\frac{\cos^2 \theta}{\mu^2} - \frac{\sin^2 \theta}{\epsilon^2} \right) + E_1 \left(\frac{\sin^2 \theta}{\mu^2} - \frac{\cos^2 \theta}{\epsilon^2} \right) + E_0 E_1 \sin 2\theta \left(\frac{1}{\mu} - \frac{1}{\epsilon} \right) = 0. \quad (5)$$

Comparing (5) and (1), we obtain

$$\begin{cases} \frac{\cos^2 \theta}{\mu^2} - \frac{\sin^2 \theta}{\epsilon^2} = \frac{1}{\mu^2 \cos 2\theta} & (6) \\ \frac{\sin^2 \theta}{\mu^2} - \frac{\cos^2 \theta}{\epsilon^2} = \frac{1}{\mu^2 \sin 2\theta} & (7) \\ \text{constant} \left(\frac{1}{\mu} - \frac{1}{\epsilon} \right) = \frac{\sin 2\theta}{\mu \epsilon \cos 2\theta} & (8) \end{cases}$$

Eqs. (6), (7), and (8) can be solved for three unknowns:

$$\theta = \frac{1}{2} \tan^{-1} \left(\frac{\mu \epsilon \sin 2\theta}{\mu^2 - \epsilon^2} \right),$$

$$a = \sqrt{\frac{\mu^2 \cos 2\theta (\mu^2 - \epsilon^2)}{\mu^2 \cos 2\theta - \epsilon^2 \sin 2\theta}} \sin \theta,$$

$$b = \sqrt{\frac{\mu^2 \sin 2\theta (\mu^2 - \epsilon^2)}{\mu^2 \sin 2\theta - \epsilon^2 \cos 2\theta}} \cos \theta.$$

In particular, if $\epsilon_1 = \epsilon_2 = \epsilon$, $\mu_1 = \mu_2 = \mu$, $\mu < \epsilon$, $\mu > \epsilon$, $\mu < \epsilon$, $\mu > \epsilon$.

Ex. 1. Let an elliptically polarized plane wave be represented by the phasor (with propagation factor $e^{i(kz - \omega t)}$ omitted)

$$\mathbf{E} = E_1 \hat{x} + E_2 \hat{y} e^{i\phi}$$

where E_1 , E_2 , and ϕ are arbitrary constants.

Right-hand circularly polarized wave: $E_1 = E_2$, $\phi = \pi/2$.

Left-hand circularly polarized wave: $E_1 = E_2$, $\phi = -\pi/2$.

$$\text{If } E_1 = \frac{1}{2}(E_0 + jE_0 e^{i\phi}) \text{ and } E_2 = \frac{1}{2}(E_0 + jE_0 e^{i\phi}),$$

$$\text{then } \mathbf{E} = \mathbf{E}_+ + \mathbf{E}_-$$

$$\text{(Right-hand circularly polarized wave)} \quad \mathbf{E}_+ = \frac{1}{2}(E_0 - E_0 e^{i\phi})$$

$$= \frac{1}{2}(E_0 \hat{x} - E_0 \hat{y} e^{i\phi}) + \frac{1}{2}(E_0 \hat{x} + E_0 \hat{y} e^{i\phi})$$

$$= E_0 \hat{x}, \quad \text{where } \mathbf{E}_+ \text{ and } \mathbf{E}_- \text{ are}$$

right-hand and left-hand elliptically polarized waves respectively.

Similarly, from left-hand circularly polarized wave (with $\phi = -\pi/2$)

$$\mathbf{E} = \frac{1}{2}(E_0 + jE_0 e^{-i\phi}) + \frac{1}{2}(E_0 + jE_0 e^{-i\phi}) = E_0 \hat{x} + E_0 \hat{y} e^{-i\phi}$$

Ex. 1.1 For conducting media: $k_2 = \beta - j\sigma_2$.

$$\begin{aligned} \lambda_2^2 &= \beta^2 - \alpha^2 - 2j\sigma_2\beta \\ &= \alpha^2 j\sigma_2\beta = \alpha^2 j\sigma_2 \left(1 - j \frac{\beta}{\sigma_2}\right) \end{aligned}$$

$$\therefore \beta^2 - \alpha^2 = \sigma_2 \lambda_2^2 = \alpha^2 j\sigma_2 \quad \text{--- (1)}$$

$$\beta + \alpha^2 = |\lambda_2^2| = \alpha^2 j\sigma_2 \sqrt{1 + \left(\frac{\beta}{\sigma_2}\right)^2} \quad \text{--- (2)}$$

From (1) and (2) we obtain

$$\alpha = \omega \sqrt{\frac{\mu\epsilon}{2}} \left[\sqrt{1 + \left(\frac{\beta}{\sigma_2}\right)^2} - 1 \right]^{1/2}, \quad \beta = \omega \sqrt{\frac{\mu\epsilon}{2}} \left[\sqrt{1 + \left(\frac{\beta}{\sigma_2}\right)^2} + 1 \right]^{1/2}$$

Ex. 1.2 All three media are good dielectrics, $\left(\frac{\beta}{\sigma_2}\right)^2 \ll 1$.

$$\alpha = \omega \sqrt{\frac{\mu\epsilon}{2}}, \quad \beta = \frac{\sigma_2}{2}, \quad \gamma_2 = (1 - j)\frac{\sigma_2}{2}$$

(i) $f = 40$ (MHz)

	γ_1 (dB)	n (ppm)	α (Np/m)	β (rad/m)
Upper	$1.12 \times 10^{-2} \text{ dB/m}^2$	1.12×10^{-2}	1.12×10^{-2}	1.12×10^{-2}
Lower	$1.12 \times 10^{-2} \text{ dB/m}^2$	1.12×10^{-2}	1.12×10^{-2}	1.12×10^{-2}
Average	$1.12 \times 10^{-2} \text{ dB/m}^2$	1.12×10^{-2}	1.12×10^{-2}	1.12×10^{-2}

(ii) $f = 1$ (MHz)

	γ_1 (dB)	n (ppm)	α (Np/m)	β (rad/m)
Upper	$1.12 \times 10^{-2} \text{ dB/m}^2$	1.12×10^{-2}	1.12×10^{-2}	1.12×10^{-2}
Lower	$1.12 \times 10^{-2} \text{ dB/m}^2$	1.12×10^{-2}	1.12×10^{-2}	1.12×10^{-2}
Average	$1.12 \times 10^{-2} \text{ dB/m}^2$	1.12×10^{-2}	1.12×10^{-2}	1.12×10^{-2}

(iii) $f = 1$ (kHz)

	γ_1 (dB)	n (ppm)	α (Np/m)	β (rad/m)
Upper	$1.12 \times 10^{-2} \text{ dB/m}^2$	1.12×10^{-2}	1.12×10^{-2}	1.12×10^{-2}
Lower	$1.12 \times 10^{-2} \text{ dB/m}^2$	1.12×10^{-2}	1.12×10^{-2}	1.12×10^{-2}
Average	$1.12 \times 10^{-2} \text{ dB/m}^2$	1.12×10^{-2}	1.12×10^{-2}	1.12×10^{-2}

Beispiel $f = 2 \cdot 10^8 \text{ (Hz)}$, $\lambda = 1 \text{ m}$, $\tan \zeta = \frac{f}{c} = 10^8$
 a) $\xi_p(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right) = \cos \left(\frac{2\pi}{1} x - \frac{2\pi}{10^{-8}} t \right)$

$$c^{100} = \frac{1}{2} \rightarrow x = \frac{1}{2} \text{ (m)} = 0,5 \text{ (m)}$$

b) $\xi_p(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right) = \cos \left(\frac{2\pi}{1} x - \frac{2\pi}{10^{-8}} t \right)$

$\xi_p(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right) = \cos \left(\frac{2\pi}{1} x - \frac{2\pi}{10^{-8}} t \right)$

$$x = \frac{1}{2} = 0,5 \text{ (m)}$$

$$x_1 = \frac{1}{2} = 0,5 \text{ (m)}$$

$$x_2 = \frac{1}{2} = 0,5 \text{ (m)}$$

c) $\xi(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right)$

$$\xi(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right) = \cos \left(\frac{2\pi}{1} x - \frac{2\pi}{10^{-8}} t \right)$$

$$\xi(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right) = \cos \left(\frac{2\pi}{1} x - \frac{2\pi}{10^{-8}} t \right)$$

Beispiel $\xi(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right) = \cos \left(\frac{2\pi}{1} x - \frac{2\pi}{10^{-8}} t \right)$

a) $\xi(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right) = \cos \left(\frac{2\pi}{1} x - \frac{2\pi}{10^{-8}} t \right)$

$$\xi(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right) = \cos \left(\frac{2\pi}{1} x - \frac{2\pi}{10^{-8}} t \right)$$

$$\xi(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right) = \cos \left(\frac{2\pi}{1} x - \frac{2\pi}{10^{-8}} t \right)$$

$$\xi(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right) = \cos \left(\frac{2\pi}{1} x - \frac{2\pi}{10^{-8}} t \right)$$

b) $\xi(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right) = \cos \left(\frac{2\pi}{1} x - \frac{2\pi}{10^{-8}} t \right)$

c) $\xi(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right) = \cos \left(\frac{2\pi}{1} x - \frac{2\pi}{10^{-8}} t \right)$

$$\xi(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right) = \cos \left(\frac{2\pi}{1} x - \frac{2\pi}{10^{-8}} t \right)$$

$$\xi(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right) = \cos \left(\frac{2\pi}{1} x - \frac{2\pi}{10^{-8}} t \right)$$

Beispiel a) $\xi(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right) = \cos \left(\frac{2\pi}{1} x - \frac{2\pi}{10^{-8}} t \right)$

b) $\xi(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right) = \cos \left(\frac{2\pi}{1} x - \frac{2\pi}{10^{-8}} t \right)$

$$\xi(t) = \cos \left(\frac{2\pi}{\lambda} x - \frac{2\pi}{T} t \right) = \cos \left(\frac{2\pi}{1} x - \frac{2\pi}{10^{-8}} t \right)$$

Ex 11.11



Assume the interface to be stratified into layers having infinite thickness

$$d_1 = d_2 = d_3 = \dots = \infty$$

The corresponding equivalent permittivities of the layers are:

$$\epsilon_1 = \epsilon_2 \left(1 - \frac{d}{d}\right) \text{ with } \epsilon_2 = \frac{\epsilon_1 \epsilon_2}{\epsilon_1 + \epsilon_2}$$

$$\text{and } \epsilon_3 = \epsilon_2 > \epsilon_1 > \epsilon_2 > \epsilon_3 > \dots = \epsilon_{\infty} \left(\frac{\epsilon_1 \epsilon_2}{\epsilon_1 + \epsilon_2}\right)$$

From Snell's law of refraction

$$\sin \theta_i = \sin \theta_r \sqrt{\epsilon_2 / \epsilon_1} = \sin \theta_r \sqrt{1 - \frac{d}{d}}$$

$$\sin \theta_i = \sin \theta_r \sqrt{\epsilon_2 / \epsilon_1} = \sin \theta_r \sqrt{\epsilon_2 / \epsilon_1}$$

$$\sin \theta_i = \sin \theta_r \sqrt{\epsilon_2 / \epsilon_1} = \sin \theta_r \sqrt{\epsilon_2 / \epsilon_1}$$

For total reflection at the layer with ϵ_{∞} , the angle of refraction $\theta_{\infty} = 90^\circ$, and $\sin \theta_{\infty} = 1 = \sin \theta_i \sqrt{\epsilon_2 / \epsilon_{\infty}}$

$$\theta_{\infty} = \theta_i \left(1 - \frac{d}{\epsilon_{\infty}}\right) = \theta_i \sin \theta_i$$

$$\Rightarrow \theta = \theta_{\infty} \cos \theta = \theta \sqrt{\epsilon_{\infty} / \epsilon_2}$$

Ex 11.12 a) From Eq (11.11): $\epsilon_2 = \frac{\epsilon_1 \epsilon_2}{\epsilon_1 + \epsilon_2} = \epsilon_2 \left(\frac{\epsilon_1}{\epsilon_1 + \epsilon_2}\right) = \epsilon_2 \left(\frac{1}{1 + \frac{\epsilon_2}{\epsilon_1}}\right)$

$$\Rightarrow 1 = \frac{1}{1 + \frac{\epsilon_2}{\epsilon_1}} \Rightarrow \frac{\epsilon_2}{\epsilon_1} = 1$$

$$\epsilon_2 = \epsilon_1 \left(\frac{1}{1 + \frac{\epsilon_2}{\epsilon_1}}\right) = \epsilon_1 \left(\frac{1}{2}\right)$$

Ex 11.13 $\epsilon_{\infty} = \epsilon_1 \epsilon_2 / (\epsilon_1 + \epsilon_2)$

a) $\epsilon_1 = \sqrt{2} \epsilon_0 = 1.414 \epsilon_0$ (air), $\epsilon_2 = \epsilon_0$ (vac)

b) $\epsilon_1 = \sqrt{2} \epsilon_0 = 1.414 \epsilon_0$ (air), $\epsilon_2 = \epsilon_0$ (vac)

c) $\epsilon_1 = \epsilon_0$ (vac), $\epsilon_2 = \epsilon_0$ (vac)

d) $\epsilon_1 = \epsilon_0$ (vac), $\epsilon_2 = 2.25 \epsilon_0$ (vac)

Ex. 12 Assume a uniformly polarized plane sheet:

$$\vec{D}(x, y) = \epsilon_0 \epsilon_p \cos(\omega t - k_0 y + \phi) \hat{x} + \epsilon_0 \epsilon_p \sin(\omega t - k_0 y + \phi) \hat{y}$$

$$\vec{P}(x, y) = \epsilon_0 \frac{\epsilon_p}{\epsilon_0} \cos(\omega t - k_0 y + \phi) \hat{x} - \epsilon_0 \frac{\epsilon_p}{\epsilon_0} \sin(\omega t - k_0 y + \phi) \hat{y}$$

Applying Gauss, $\vec{D} = \vec{E} + \vec{P} = \epsilon_0 \frac{\epsilon_p}{\epsilon_0} [\cos(\omega t - k_0 y + \phi) \hat{x} + \sin(\omega t - k_0 y + \phi) \hat{y}]$
 $= \epsilon_0 \frac{\epsilon_p}{\epsilon_0} \hat{r}$; ϵ_0 not a function independent of ϵ and ω

Ex. 13 $\vec{E} = E_0 \hat{x}_p + E_0 \hat{x}_d$

$$\vec{P} = \frac{1}{2} \epsilon_0 E_0 \hat{E} = \frac{1}{2} \epsilon_0 (E_0 \hat{x}_p - E_0 \hat{x}_d)$$

$$\vec{D}_{\text{ext}} = \frac{1}{2} \epsilon_0 (\vec{E} + \vec{P}) = \epsilon_0 \frac{3}{4} (E_0 \hat{x}_p + E_0 \hat{x}_d)$$

Ex. 14 From Gauss' law, $\vec{E} = E_0 \frac{\hat{r}}{r^2}$ - where ρ is the free charge density on the inner conductor.

$$V_0 = - \int_{\infty}^R \vec{E} \cdot d\vec{s} = \frac{E_0}{2\epsilon_0} \ln\left(\frac{R}{a}\right) \implies \vec{E} = E_0 \frac{2\epsilon_0}{\ln(R/a)}$$

From Ampere's circuital law, $\vec{H} = E_0 \frac{\hat{\phi}}{2\pi r}$

Applying vector, $\vec{B} = \vec{E} \times \vec{H} = E_0 \frac{E_0 \hat{r} \times \hat{\phi}}{2\pi r \ln(R/a)}$

Power transmitted in our cross-sectional area:

$$P = \int \vec{S} \cdot d\vec{a} = \frac{E_0^2}{2\pi \ln^2(R/a)} \int_a^R \left(\frac{1}{r}\right) \cdot 2\pi r dr = \frac{1}{2} E_0^2$$

Ex. 15 a) $\vec{E} = \frac{E_0}{\sqrt{2}} e^{i(\omega t - k_0 y)} \hat{x}$; $\vec{P} = \epsilon_0^2 E_0$

b) $\vec{D}(x, y) = \epsilon_0 \epsilon_p e^{i(\omega t - k_0 y)} \cos(\omega t - k_0 y)$

$$\vec{E}_0 = (1 + \epsilon_p) \frac{E_0}{\sqrt{2}} = (1 + \epsilon_p) \frac{E_0}{\sqrt{2}} = \frac{\sqrt{2}}{\sqrt{2}} E_0 \sqrt{1 + \epsilon_p}$$

$$\vec{E}(x, y) = \epsilon_0 \frac{\sqrt{2}}{\sqrt{2}} E_0 \sqrt{1 + \epsilon_p} \cos(\omega t - k_0 y) \hat{x} = \frac{\sqrt{2}}{\sqrt{2}} E_0 \sqrt{1 + \epsilon_p}$$

c) $\vec{D}_{\text{ext}} = \frac{1}{2} \epsilon_0 (\vec{E} + \vec{P}) = \epsilon_0 \frac{1}{2} \frac{\sqrt{2}}{\sqrt{2}} E_0 \sqrt{1 + \epsilon_p} \cos \frac{\omega}{2}$
 $= \epsilon_0 \frac{1}{2} \left(\frac{\sqrt{2}}{\sqrt{2}}\right) \quad (\text{value})$

Ex 2.1 Given $\vec{r}_1 = r_1(\hat{x}_1 + \hat{y}_1) e^{i\omega t}$

a) Assume reduced $\vec{r}_2(t) = (r_2 \hat{x}_2 + r_2 \hat{y}_2) e^{i\omega t}$

Boundary condition at $t=0$: $\vec{r}_1(0) = \vec{r}_2(0) = 0$

$\implies \vec{r}_1(0) = r_1(-\hat{x}_1 + i\hat{y}_1) e^{i\omega t}$ is \hat{x}_1 -based circularly polarized wave in xy -plane.

b) $\vec{r}_2 = (r_2 - \hat{z}_2) = \vec{z} \implies r_2 = (\hat{y}_2 \cos + \hat{x}_2 \sin) = \vec{z} (\hat{x}_2 \cos + \hat{y}_2 \sin)$

$\vec{r}_1(0) = \frac{r_1}{2} \hat{x}_1 + \hat{z}_1(0) = \frac{r_1}{2} (\hat{x}_1 + \hat{z}_1)$, $\vec{r}_2(0) = \frac{r_2}{2} (\hat{x}_2 + \hat{z}_2) = \frac{r_2}{2} (\hat{x}_2 + \hat{z}_2)$

$\vec{r}_1(0) = \vec{r}_2(0) = \hat{z}_1(0) = \frac{r_1}{2} (\hat{x}_1 + \hat{z}_1)$

$\vec{z}_1 = -\hat{x}_1 + \hat{z}_1(0) = \frac{r_1}{2} (\hat{x}_1 + \hat{z}_1)$

c) $\vec{r}_1(t) = r_1 (\hat{z}_1(t) + \hat{x}_1(t)) e^{i\omega t}$

$= r_1 r_2 (\hat{x}_2 + \hat{y}_2) e^{i\omega t} + r_1 r_2 (\hat{x}_2 + \hat{y}_2) e^{i\omega t}$ (same)

$= 2r_1 r_2 (\hat{x}_2 + \hat{y}_2) e^{i\omega t}$ (same)

$= 2r_1 r_2 \sin(\omega t) (\hat{x}_2 \cos + \hat{y}_2 \sin)$

Ex 2.2 Given $\vec{r}_1(r_1, \hat{z}_1) = r_1 \hat{z}_1 e^{i\omega t + i\theta}$ ($\hat{z}_1 = \hat{z}$)

a) $\hat{x}_1 = \hat{x}$, $\hat{y}_1 = \hat{y} \implies \hat{z}_1 = \hat{z} = \sqrt{\hat{x}_1^2 + \hat{y}_1^2} = \hat{z}$ (same \hat{z}_1)

$\hat{x}_2 = \hat{x} \cos + \hat{y} \sin$ is \hat{z}_2 based only \hat{z}_1 based \hat{z}_2 based \hat{z}_1 based \hat{z}_2 based

b) $\vec{r}_1(r_1, \hat{z}_1) = r_1 \hat{z}_1 \cos(\theta) = r_1 \hat{z}_1 \cos(\theta)$ ($\hat{z}_1 = \hat{z}$)

$\vec{r}_1(r_1) = \frac{r_1}{2} (\hat{x}_1 + \hat{z}_1)$ ($\hat{x}_1 = \frac{r_1}{2} (\hat{x}_2 + \hat{z}_2) + \hat{z}_1 \hat{z}_2$)

$= \frac{r_1}{2} (\hat{x}_2 \cos + \hat{z}_2 \sin) + \hat{z}_1 \cos(\theta) = \frac{r_1}{2} (\hat{x}_2 \cos + \hat{z}_2 \sin) + \hat{z}_1 \cos(\theta)$

$\vec{r}_1(r_1, \hat{z}_1) = (\frac{r_1}{2} \hat{x}_2 \cos + \hat{z}_1 \hat{z}_2 \sin) \cos(\theta) + \hat{z}_1 \cos(\theta)$ (same)

c) $\cos \theta = \hat{z}_2 - \hat{z}_1 = (\frac{r_1}{2}) \hat{z}_2 = \hat{z}_1 \implies \hat{z}_2 = \cos(\theta) \hat{z}_1 = \hat{z}_1$

d) $\vec{r}_1(r_1) = r_1 \hat{z}_1 \cos(\theta) \implies \vec{r}_1(r_1) = r_1 \hat{z}_1 \cos(\theta)$

$\vec{r}_1(r_1, \hat{z}_1) = \frac{r_1}{2} (\hat{x}_2 + \hat{z}_2)$ ($\hat{x}_2 = \hat{x}_1 \cos + \hat{z}_1 \hat{z}_2$)

$= (\frac{r_1}{2} \hat{x}_2 + \hat{z}_1 \hat{z}_2) e^{i\omega t + i\theta}$

e) $\vec{r}_1(r_1) = \vec{r}_1(r_1) = \vec{r}_1(r_1) = r_1 \hat{z}_1 \cos(\theta) e^{i\omega t + i\theta}$

$= r_1 \hat{z}_1 \cos(\theta) e^{i\omega t + i\theta}$ (same)

$\vec{r}_1(r_1) = \vec{r}_1(r_1) = \vec{r}_1(r_1) = r_1 \hat{z}_1 \cos(\theta) e^{i\omega t + i\theta}$ (same)

Ex 10.12 Given $\mathcal{L}\{f(x)\} = \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10}$ (Given)

(i) $f_1(x) = \sin x$, $f_2(x) = \cos x$ $\implies \mathcal{L} = \sqrt{10^2 + 200^2} = 200$ (Given)

$\mathcal{L} = \sin 200x = 200 \cos 200x \implies \mathcal{L} = \sin 200x \cos 200x + \cos 200x \sin 200x$

(ii) $\mathcal{L}\{f_1(x)\} = \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10} = \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10}$ (Given)

$$\mathcal{L}\{f_1(x)\} = \frac{1}{200} \mathcal{L}_1 = \mathcal{L}_1 = \frac{1}{200} \mathcal{L}\{f_1(x) + f_2(x)\} = \mathcal{L}_1 \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10}$$

$$= \mathcal{L}_1 \left(\frac{1}{200} \right) e^{200x-10}$$

$$\mathcal{L}\{f_1(x)\} = \mathcal{L}_1 \left(\frac{1}{200} \right) e^{200x-10} = \mathcal{L}_1 \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10}$$
 (Given)

(i) $\sin 200x = \mathcal{L}_1 \cdot \mathcal{L}_2 = \frac{1}{200}$ $\implies \mathcal{L}_1 = 200 \left(\frac{1}{200} \right) = 1$

(ii) Given that $\mathcal{L}_1 = \mathcal{L}\{f_1(x)\} = 1$ and $\mathcal{L}_2 = \mathcal{L}\{f_1(x) + f_2(x)\} = 200$ (Given)

$$\mathcal{L}_1 \mathcal{L}_2 = 1 \cdot 200 = \mathcal{L}\{f_1(x)\} \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10}$$
 (Given)

$$\mathcal{L}\{f_1(x)\} = \frac{1}{200} \mathcal{L}_1 = \mathcal{L}_1 \mathcal{L}_2 = \frac{1}{200} \mathcal{L}\{f_1(x) + f_2(x)\} = \mathcal{L}_1 \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10}$$

$$= \mathcal{L}_1 \left(\frac{1}{200} \right) e^{200x-10}$$
 (Given)

(iii) $\mathcal{L}\{f_1(x)\} = \mathcal{L}\{f_1(x) + f_2(x)\} = \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10}$ (Given)

$$\mathcal{L}\{f_1(x)\} \mathcal{L}\{f_1(x) + f_2(x)\} = \mathcal{L}_1 \left(\frac{1}{200} \right) e^{200x-10} e^{200x-10}$$
 (Given)

Ex 10.13 (i) From Eqn (10-110) and (10-111):

$$\mathcal{L}\{f_1(x)\} = \mathcal{L}_1 \mathcal{L}_2 \text{ (where } \mathcal{L}_1 \text{ is the } \mathcal{L}\{f_1(x)\} \text{ and } \mathcal{L}_2 \text{ is the } \mathcal{L}\{f_1(x) + f_2(x)\} \text{)}$$

$$\mathcal{L}\{f_1(x)\} = \mathcal{L}_1 \left[\mathcal{L}_2 \left(\frac{1}{200} \right) e^{200x-10} \right]$$

$$= \mathcal{L}_1 \left(\frac{1}{200} \right) e^{200x-10}$$

(ii) $\mathcal{L}_1 = \frac{1}{200} \mathcal{L}\{f_1(x)\} = \mathcal{L}_2 \mathcal{L}_1 \left(\frac{1}{200} \right) e^{200x-10}$

Ex 10.14 (i) From Eqn (10-110) and (10-111):

$$\mathcal{L}\{f_1(x)\} = \mathcal{L}_1 \mathcal{L}_2 \left[\mathcal{L}_2 \left(\frac{1}{200} \right) e^{200x-10} \right]$$

$$= \mathcal{L}_1 \left(\frac{1}{200} \right) e^{200x-10}$$

$$\mathcal{L}\{f_1(x)\} = \mathcal{L}_1 \left(\frac{1}{200} \right) e^{200x-10}$$

(ii) $\mathcal{L}_1 = \frac{1}{200} \mathcal{L}\{f_1(x)\} = \mathcal{L}_2 \mathcal{L}_1 \left(\frac{1}{200} \right) e^{200x-10}$

Ex 10.15 For normal incidence: $1 + \Gamma = 2$, where $|\Gamma| \leq 1$.

$$\mathcal{L}\{f_1(x)\} = \mathcal{L}\{f_2(x)\} \implies \mathcal{L}_1 = \mathcal{L}_2 \implies \mathcal{L}_1 = \mathcal{L}_2 \implies |\Gamma| = \frac{1}{2}$$

$$\therefore \mathcal{L} = \frac{1}{2} \left(\frac{1}{200} \right) e^{200x-10} \implies \mathcal{L}_1 = \mathcal{L}_2 \mathcal{L}_1 \left(\frac{1}{200} \right) e^{200x-10}$$

Ex 11) In the decay reaction (reaction 11):

$$K_1 = K_2 K_3 e^{-\beta \mu_2} e^{-\beta \mu_3}$$

where from
 Maxwell-Boltzmann, $n_2 = n_2^0 e^{-\beta \mu_2} \left[\sqrt{1 - \frac{v^2}{c^2}} \right]^{-3}$, $n_3 = n_3^0 e^{-\beta \mu_3} \left[\sqrt{1 - \frac{v^2}{c^2}} \right]^{-3}$

Given: $\beta = 0$ (rest), $n_2 = n_3 = 2.0 \times 10^{23} \text{ m}^{-3}$

So $K_2 = \frac{n_2}{n_1} = 0.1 \implies n_2 = 0.1 n_1$ (initial, K_2 is constant)

$$n_2 = \sqrt{\frac{2\pi m_2 k_B T}{h^2}} e^{-\beta \mu_2} e^{-\beta \mu_2} = 2.0 \times 10^{23} \text{ m}^{-3}$$

$$K_1 = K_2 K_3 e^{-\beta \mu_2} e^{-\beta \mu_3}, \quad K_2 K_3 e^{-\beta \mu_2} = n_2 \frac{h^3}{(2\pi m_2 k_B T)^{3/2}} e^{-\beta \mu_2}$$

$$\text{So } K_1 = n_2 \frac{h^3}{(2\pi m_2 k_B T)^{3/2}} e^{-\beta \mu_2} \implies K_1 = n_2 \frac{h^3}{(2\pi m_2 k_B T)^{3/2}} e^{-\beta \mu_2}$$

Therefore, $K_1 = \left[\begin{array}{l} n_2 = 2.0 \times 10^{23} \\ n_2 = n_2^0 e^{-\beta \mu_2} \left[\sqrt{1 - \frac{v^2}{c^2}} \right]^{-3} \end{array} \right]$
 $\implies n_2^0 = 2.0 \times 10^{23}, \quad K_1 = 2.0 \times 10^{23}$

- ∴ $K_1 (2.0 \times 10^{23}) e^{-\beta \mu_2} e^{-\beta \mu_3} = 2.0 \times 10^{23}$ (initial)
 $K_1 (2.0 \times 10^{23}) e^{-\beta \mu_2} e^{-\beta \mu_3} = 2.0 \times 10^{23}$ (initial)
 $K_1 (2.0 \times 10^{23}) e^{-\beta \mu_2} e^{-\beta \mu_3} = 2.0 \times 10^{23}$ (initial)
 $K_1 (2.0 \times 10^{23}) e^{-\beta \mu_2} e^{-\beta \mu_3} = 2.0 \times 10^{23}$ (initial)

∴ $K_1 = K_2 \left(\frac{h^3}{(2\pi m_2 k_B T)^{3/2}} e^{-\beta \mu_2} \right) = 2.0 \times 10^{23}$ (initial)

$$K_1 = K_2 \frac{h^3}{(2\pi m_2 k_B T)^{3/2}} e^{-\beta \mu_2} = 2.0 \times 10^{23} \text{ (initial)}$$

Ex 12) $\Gamma = \frac{1}{\tau} = \frac{1}{\tau_0 \sqrt{1 - \frac{v^2}{c^2}}}$, $n_2 = \sqrt{\frac{2\pi m_2 k_B T}{h^2}} e^{-\beta \mu_2} e^{-\beta \mu_2}$
 $n_2 = 2.0 \times 10^{23}$

$$\text{∴ } \Gamma = \left| \frac{1}{\tau_0 \sqrt{1 - \frac{v^2}{c^2}}} \right| = \left| \frac{1}{\tau_0 \sqrt{1 - \frac{v^2}{c^2}}} \right| = \left| 1 - \frac{v^2}{c^2} \right|$$

$$= (1 - \frac{v^2}{c^2}) - 1 + \frac{v^2}{c^2} = 1 - \frac{v^2}{c^2} \frac{v^2}{c^2}$$

Fraction of power absorbed, $F = \Gamma - \Gamma_0 = \left(\frac{1}{\tau_0 \sqrt{1 - \frac{v^2}{c^2}}} \right) - \frac{1}{\tau_0}$
 $= \frac{1}{\tau_0} \left[\frac{1}{\sqrt{1 - \frac{v^2}{c^2}}} - 1 \right]$

∴ $F = \frac{1}{\tau_0} \left[\frac{1}{\sqrt{1 - \frac{v^2}{c^2}}} - 1 \right]$ For $v = 0$, $F = 0$
 $F = \frac{1}{\tau_0} \left[\frac{1}{\sqrt{1 - \frac{v^2}{c^2}}} - 1 \right]$

Ex 10.12 From Eqs. (8-104) through (8-106)

$$E_1 = E_0 \left[\frac{1}{2} (1 + \sqrt{\epsilon_r}) + \frac{1}{2} (1 - \sqrt{\epsilon_r}) \right], \quad H_1 = E_0 \sqrt{\frac{1}{2} (\epsilon_r + 1)} \left[\frac{1}{2} (1 + \sqrt{\epsilon_r}) + \frac{1}{2} (1 - \sqrt{\epsilon_r}) \right],$$

$$E_2 = E_0 \left[\frac{1}{2} (1 + \sqrt{\epsilon_r}) - \frac{1}{2} (1 - \sqrt{\epsilon_r}) \right], \quad H_2 = E_0 \sqrt{\frac{1}{2} (\epsilon_r + 1)} \left[\frac{1}{2} (1 + \sqrt{\epsilon_r}) - \frac{1}{2} (1 - \sqrt{\epsilon_r}) \right],$$

$$E_3 = E_0 e^{i\phi}, \quad H_3 = E_0 \sqrt{\epsilon_r} e^{i\phi}$$

Boundary conditions at $x=0$: $E_{1y} = E_{2y}$, $H_{1z} = H_{2z}$.

Boundary conditions at $x=d$: $E_{2y} = E_{3y}$, $H_{2z} = H_{3z}$.

Four equations to solve for unknowns E_0 , E_1 , E_2 , and E_3 in terms of E_0 :

$$\begin{aligned} \text{(1)} \quad E_0 &= \frac{E_0 \left[\frac{1}{2} (1 + \sqrt{\epsilon_r}) + \frac{1}{2} (1 - \sqrt{\epsilon_r}) \right]}{\sqrt{\epsilon_r} \left[\frac{1}{2} (1 + \sqrt{\epsilon_r}) + \frac{1}{2} (1 - \sqrt{\epsilon_r}) \right]} E_0, & \text{where} \\ E_1 &= \frac{E_0 \left[\frac{1}{2} (1 + \sqrt{\epsilon_r}) + \frac{1}{2} (1 - \sqrt{\epsilon_r}) \right]}{\sqrt{\epsilon_r} \left[\frac{1}{2} (1 + \sqrt{\epsilon_r}) + \frac{1}{2} (1 - \sqrt{\epsilon_r}) \right]} E_0, & \Gamma = \frac{\sqrt{\epsilon_r} - 1}{\sqrt{\epsilon_r} + 1} \\ E_2 &= \frac{E_0 \left[\frac{1}{2} (1 + \sqrt{\epsilon_r}) - \frac{1}{2} (1 - \sqrt{\epsilon_r}) \right]}{\sqrt{\epsilon_r} \left[\frac{1}{2} (1 + \sqrt{\epsilon_r}) - \frac{1}{2} (1 - \sqrt{\epsilon_r}) \right]} E_0, & \Gamma = \frac{1 - \sqrt{\epsilon_r}}{1 + \sqrt{\epsilon_r}} \\ E_3 &= \frac{E_0 \left[\frac{1}{2} (1 + \sqrt{\epsilon_r}) - \frac{1}{2} (1 - \sqrt{\epsilon_r}) \right]}{\sqrt{\epsilon_r} \left[\frac{1}{2} (1 + \sqrt{\epsilon_r}) - \frac{1}{2} (1 - \sqrt{\epsilon_r}) \right]} E_0, & R = \frac{1 - \sqrt{\epsilon_r}}{1 + \sqrt{\epsilon_r}} \end{aligned}$$

Ex 10.13 $\epsilon_r = 4$, $\mu_r = 1$, $E_0 = \frac{100}{\sqrt{2}} E_0$.

$\therefore \Gamma = \frac{1 - \sqrt{4}}{1 + \sqrt{4}} = \frac{1 - 2}{1 + 2} = \frac{-1}{3}$ and where $E_0 = E_0$ in brackets.

$E_1 = \frac{1}{3} E_0$, $E_2 = \frac{2}{3} E_0$, $E_3 = \frac{2}{3} E_0$.

Ex 10.14 From Example 8-11: $\Gamma_1 = \frac{1 - \sqrt{\epsilon_r}}{1 + \sqrt{\epsilon_r}}$, $\Gamma_2 = \frac{1 - \sqrt{\epsilon_r}}{1 + \sqrt{\epsilon_r}}$.

at $x=0$ (boundary of medium 1): $H_1 = \frac{1}{\sqrt{\epsilon_r}} E_1 = \frac{1}{\sqrt{\epsilon_r}} E_0 \left[\frac{1}{2} (1 + \sqrt{\epsilon_r}) + \frac{1}{2} (1 - \sqrt{\epsilon_r}) \right]$

at $x=d$ (boundary of medium 2): $H_2 = \frac{1}{\sqrt{\epsilon_r}} E_2 = \frac{1}{\sqrt{\epsilon_r}} E_0 \left[\frac{1}{2} (1 + \sqrt{\epsilon_r}) - \frac{1}{2} (1 - \sqrt{\epsilon_r}) \right]$

$H_3 = E_3 = \frac{1}{\sqrt{\epsilon_r}} E_0 \left[\frac{1}{2} (1 + \sqrt{\epsilon_r}) - \frac{1}{2} (1 - \sqrt{\epsilon_r}) \right]$

From Eq. (8-103) and using boundary conditions at $x=0$ and $x=d$:

$$\text{At } x=0: \quad E_0 = E_0 \left[\frac{1}{2} (1 + \sqrt{\epsilon_r}) + \frac{1}{2} (1 - \sqrt{\epsilon_r}) \right] = \frac{1}{\sqrt{\epsilon_r}} E_0 \left[\frac{1}{2} (1 + \sqrt{\epsilon_r}) + \frac{1}{2} (1 - \sqrt{\epsilon_r}) \right]$$

$$\Gamma = \frac{1 - \sqrt{\epsilon_r}}{1 + \sqrt{\epsilon_r}} = \frac{1 - 2}{1 + 2} = \frac{-1}{3}$$

Percentage of power reflected = $|\Gamma|^2 \times 100\%$
 $= \left(\frac{1}{3} \right)^2 \times 100\% = 11.1\%$

$$\text{E.10.11} \quad C = \frac{2\sqrt{2}a^2}{2\sqrt{2}a^2} \cdot 2a^2 = 2 \frac{2\sqrt{2}a^2 \cdot 2a^2}{2\sqrt{2}a^2 \cdot 2a^2}$$

$$C = \frac{2\sqrt{2}}{2\sqrt{2}} \longrightarrow \frac{2}{2} = \frac{2\sqrt{2}}{2\sqrt{2}}$$

$$C = \frac{2\sqrt{2}}{2\sqrt{2}} \longrightarrow \frac{2}{2} = \frac{2\sqrt{2}}{2\sqrt{2}}$$

$$C = \frac{2\sqrt{2} \cdot \frac{2\sqrt{2} \cdot 2a^2}{2\sqrt{2} \cdot 2a^2} = \frac{2}{2} \left(\frac{2\sqrt{2}}{2\sqrt{2}} = \frac{2\sqrt{2} \cdot 2a^2}{2\sqrt{2} \cdot 2a^2} \right)}$$

$$= \frac{2 + 2 \cdot \frac{2\sqrt{2}}{2\sqrt{2}} \cdot 2a^2 = \frac{2\sqrt{2}}{2\sqrt{2}} \left(\frac{2\sqrt{2}}{2\sqrt{2}} = \frac{2\sqrt{2} \cdot 2a^2}{2\sqrt{2} \cdot 2a^2} \right)}$$

$$= \frac{2\sqrt{2} + 2\sqrt{2} \cdot \frac{2\sqrt{2} \cdot 2a^2}{2\sqrt{2} \cdot 2a^2}}$$

$$\text{E.10.12} \quad E_1 = a_1 (a_2 e^{i\omega t} + a_3 e^{i\omega t})$$

$$E_2 = a_2 \frac{1}{2} (a_2 e^{i\omega t} + a_3 e^{i\omega t})$$

$$E_3 = a_3 (a_1' e^{i\omega t} + a_2' e^{i\omega t})$$

$$E_4 = a_4 \frac{1}{2} (a_1' e^{i\omega t} + a_2' e^{i\omega t})$$

$$\text{At } t = 0, \quad E_1 = 0 \longrightarrow E_2 = -a_2' e^{i\omega t}$$

$$E_3 = a_3 a_1' (e^{i\omega t} - e^{i\omega t + \pi})$$

$$E_4 = a_4 \frac{1}{2} (e^{i\omega t} + e^{i\omega t + \pi})$$

$$\text{Secondary interaction: } E_1 = a_1 a_2 \longrightarrow E_2 = a_2 a_3 = a_1' (e^{i\omega t} - e^{i\omega t + \pi})$$

$$E_3 = a_3 a_4 \longrightarrow E_4 = a_4 a_1 = a_2' \frac{1}{2} (e^{i\omega t} + e^{i\omega t + \pi})$$

$$E_1' = \frac{2a_1 a_2}{(a_1 + a_2) e^{i\omega t}}$$

$$E_2' = \left(\frac{2a_3 - 2a_4 \frac{2a_1 a_2}{2a_1 + 2a_2}}{2a_3 + 2a_4} \right) a_1$$

$$\Rightarrow \mathcal{L}\{x_1(t)\} = \mathcal{L}\{x_2(t) + \sin(\omega t) - \frac{\omega}{2}(1 + \cos 2\omega t)\} = \frac{1}{s} - \frac{\omega}{2} \frac{1 + \cos 2\omega t}{s}$$

$$\Rightarrow \mathcal{L}\{x_2(t)\} = \mathcal{L}\{x_2(t) + \sin(\omega t) - \frac{\omega}{2}(1 + \cos 2\omega t) + \frac{\omega}{2}(1 + \cos 2\omega t) - \frac{\omega}{2}(1 + \cos 2\omega t)\}$$

$$\Rightarrow \mathcal{L}\{x_2(t)\} = \mathcal{L}\left\{\frac{\sin 2\omega t}{2\omega} + \frac{\sin 2\omega t}{2\omega} - \frac{\sin 2\omega t}{2\omega} + \frac{\sin 2\omega t}{2\omega} + \sin(\omega t) - \frac{\omega}{2}(1 + \cos 2\omega t)\right\}$$

$$P = \mathcal{L}\left\{\frac{\sin 2\omega t}{2\omega} + \frac{\sin 2\omega t}{2\omega} + \sin(\omega t) - \frac{\omega}{2}(1 + \cos 2\omega t)\right\}$$

$$\Rightarrow \mathcal{L}\{x_2(t)\} = \frac{1}{2\omega} \mathcal{L}\{2\sin 2\omega t + 2\omega \sin(\omega t) - \omega(1 + \cos 2\omega t)\} = 0$$

$$\Rightarrow \mathcal{L}\{x_2(t)\} = 0$$

$$\Rightarrow \mathcal{L}\{x_2(t)\} = \mathcal{L}\{x_2(t)\} = 0 \Rightarrow x_2(t) = 0 \Rightarrow x_2(t) = 0 \Rightarrow x_2(t) = 0$$

$$\underline{\text{Ex 1.11}} \quad x_1 = x_2 + x_3 = (1 + \cos t) \frac{1}{2}, \quad x_2 = x_3 = \frac{1}{2} \sin t \sqrt{2} \sin t$$

$$x_3 = (1 + \cos t) \frac{1}{2} + x_2 \quad \text{or } x_2 = x_3 - x_1$$

a) From Problem 1.10

$$\mathcal{L}\{x_1\} = \mathcal{L}\{x_2 + x_3\} = \frac{1}{2} \left(\frac{1}{s} + \frac{1}{s^2 + 1} \right) = \frac{1}{2} \frac{s + 1 + s^2 + 1}{(s^2 + 1)s} = \frac{1}{2} \frac{s^2 + s + 2}{(s^2 + 1)s}$$

$$\Rightarrow \mathcal{L}\{x_2\} = -\mathcal{L}\{x_3\} = -\frac{1}{2} \left(\frac{1}{s} + \frac{1}{s^2 + 1} \right) = -\frac{1}{2} \frac{s + 1 + s^2 + 1}{(s^2 + 1)s} = -\frac{1}{2} \frac{s^2 + s + 2}{(s^2 + 1)s}$$

$$\Rightarrow \mathcal{L}\{x_3\} = \mathcal{L}\{x_2 + x_1\} = \frac{1}{2} \left(\frac{1}{s} + \frac{1}{s^2 + 1} \right) + \frac{1}{2} \left(\frac{1}{s} + \frac{1}{s^2 + 1} \right) = \frac{1}{s} + \frac{1}{s^2 + 1}$$

$$\Rightarrow \mathcal{L}\{x_2\} = -\frac{1}{2} \frac{s^2 + s + 2}{(s^2 + 1)s}$$

$$= -\frac{1}{2} \left(\frac{1}{s} + \frac{1}{s^2 + 1} \right) = -\frac{1}{2} \frac{s + 1 + s^2 + 1}{(s^2 + 1)s} = -\frac{1}{2} \frac{s^2 + s + 2}{(s^2 + 1)s}$$

$$\mathcal{L}\{x_2\} = \frac{1}{2} \mathcal{L}\{x_3 - x_1\} = \frac{1}{2} (\mathcal{L}\{x_3\} - \mathcal{L}\{x_1\})$$

$$= \frac{1}{2} \left(\frac{1}{s} + \frac{1}{s^2 + 1} \right) - \frac{1}{2} \left(\frac{1}{s} + \frac{1}{s^2 + 1} \right) = 0$$

$$\Rightarrow \mathcal{L}\{x_2\} = \frac{1}{2} \frac{s + 1 + s^2 + 1}{(s^2 + 1)s} - \frac{1}{2} \frac{s + 1 + s^2 + 1}{(s^2 + 1)s} = 0$$

$$\mathcal{L}\{x_2\} = \frac{1}{2} \frac{s + 1 + s^2 + 1}{(s^2 + 1)s} - \frac{1}{2} \frac{s + 1 + s^2 + 1}{(s^2 + 1)s} = 0$$

$$\mathcal{L}\{x_2\} = \frac{1}{2} \frac{s + 1 + s^2 + 1}{(s^2 + 1)s} - \frac{1}{2} \frac{s + 1 + s^2 + 1}{(s^2 + 1)s} = 0$$

$$\therefore \frac{\partial \mathcal{L}}{\partial \mathcal{E}_1} = \frac{1}{2} \left(\frac{\partial}{\partial \mathcal{E}_1} \right) \frac{1}{\sin^2 \theta_1 \cos^2 \theta_1 \mu_1 \mu_2 \cos^2 \theta_2 \cos^2 \theta_3}$$

$$\frac{\partial \mathcal{L}}{\partial \mathcal{E}_1} = \frac{1}{2} \frac{1}{\mathcal{E}_1} \frac{1}{\sin^2 \theta_1 \cos^2 \theta_1 \mu_1 \mu_2 \cos^2 \theta_2 \cos^2 \theta_3}$$

At $\theta = 0^\circ$ and $\theta_2 = \theta_3 = 90^\circ$ and $\theta_1 = \theta_2 = \theta_3 = 0^\circ$, $\mu_1 = \mu_2 = \mu_3 = \mu_0$

$$\frac{\partial \mathcal{L}}{\partial \mathcal{E}_1} = \frac{1}{2 \mu_0^3 \mathcal{E}_1}$$

Ex. 21 Given $\beta = \mu_2/\mu_1$ and $\theta_1 = 0^\circ$

$$\text{From Eq. (2-102)} \quad \mathcal{R}_1 = \frac{1 - \beta \cos^2 \theta_1}{1 + \beta \cos^2 \theta_1} = \frac{1 - \beta}{1 + \beta}, \quad \mathcal{T}_1/\mathcal{E}_1 = \frac{2\beta}{1 + \beta}$$

From Eq. (2-103) $\mathcal{R}_2 = \frac{1 - \beta \cos^2 \theta_2}{1 + \beta \cos^2 \theta_2}$, $\mathcal{R}_2/\mathcal{E}_2 = \frac{1 - \beta}{1 + \beta}$

$$\text{From Eq. (2-104)} \quad \mathcal{L}_1 = \frac{2\beta \cos^2 \theta_1}{(1 + \beta) \cos^2 \theta_1 + 1} = \frac{2\beta}{1 + \beta}$$

$$\text{From Eq. (2-105)} \quad \mathcal{T}_2 = \frac{2\beta \cos^2 \theta_2}{(1 + \beta) \cos^2 \theta_2 + 1} = \frac{2\beta}{1 + \beta}$$

$$\text{From Eq. (2-106)} \quad \mathcal{R}_3 = \frac{1 - \beta \cos^2 \theta_3}{(1 + \beta) \cos^2 \theta_3 + 1} = \frac{1 - \beta}{1 + \beta}$$

$$\text{From Eq. (2-107)} \quad \mathcal{T}_3 = \frac{2\beta \cos^2 \theta_3}{(1 + \beta) \cos^2 \theta_3 + 1} = \frac{2\beta}{1 + \beta}$$

$\mathcal{R}_1 = \mathcal{R}_2 = \mathcal{R}_3$, but the phase shift of the reflected wave depends on the polarization of the incident wave. There are standing waves in the air and exponentially decaying transmitted waves in the hemisphere.

Ex. 22 $\mathcal{R}_1^2 + \mathcal{R}_2^2 + \mathcal{R}_3^2 = \mathcal{R}_1^2 + \mathcal{R}_2^2 + \mathcal{R}_3^2 = \frac{1 - \beta^2}{1 + \beta^2} + \frac{1 - \beta^2}{1 + \beta^2} + \frac{1 - \beta^2}{1 + \beta^2}$ (1)

Continuity conditions at $z=0$ for all θ imply further

$$\mathcal{R}_1 = \mathcal{R}_2 = \mathcal{R}_3 = \frac{1 - \beta^2}{1 + \beta^2} \quad \text{for all } \theta \quad \text{(2)}$$

$$\mathcal{R}_1 = \mathcal{R}_2 = \mathcal{R}_3 \quad \text{(3)}$$

Combining (1), (2) and (3), we determine for \mathcal{R}_1 and \mathcal{R}_2 in terms of ω , μ_1 , μ_2 , ϵ_1 , ϵ_2 and β . And, since

$$A_0^2 = \frac{1}{2} \rho_0 v_0^2$$

we have $v_0 = \frac{1}{\rho_0} \frac{dA_0}{dx} = \frac{1}{\rho_0} \frac{d}{dx} \left(\frac{1}{2} \rho_0 v_0^2 \right) = v_0 \frac{dv_0}{dx}$

$$a) \quad v_0 = \frac{1}{\rho_0} \frac{dA_0}{dx} = \frac{1}{\rho_0} \frac{d}{dx} \left(\frac{1}{2} \rho_0 v_0^2 \right) = v_0 \frac{dv_0}{dx} \implies dx = \frac{1}{2} \frac{dv_0}{v_0}$$

$$b) \quad v_0 = \frac{1}{\rho_0} \frac{dA_0}{dx} = \frac{1}{\rho_0} \frac{d}{dx} \left(\frac{1}{2} \rho_0 v_0^2 \right) = v_0 \frac{dv_0}{dx} \implies dx = \frac{1}{2} \frac{dv_0}{v_0}$$

$$c) \quad (v_0) = \frac{1}{\rho_0} \frac{dA_0}{dx}$$

$$v_0 = \frac{1}{\rho_0} \frac{dA_0}{dx} \implies v_0 = \frac{1}{\rho_0} \frac{d}{dx} \left(\frac{1}{2} \rho_0 v_0^2 \right) = v_0 \frac{dv_0}{dx}$$

$$\implies \frac{dv_0}{v_0} = \frac{1}{2} \frac{dv_0}{v_0} \implies dx = \frac{1}{2} \frac{dv_0}{v_0}$$

$$d) \quad \text{for } v_0 = v_0^* \implies dx = \frac{1}{2} \frac{dv_0}{v_0} = \frac{1}{2} \frac{dv_0^*}{v_0^*}$$

Ex 10



a) Small flow

$$\frac{v_1 A_1}{v_2 A_2} = \frac{v_3 A_3}{v_2 A_2}$$

$$v_2 = v_1 \left(\frac{A_1}{A_2} \right)$$

$$b) \quad v_2 = \sqrt{v_1^2 \left(\frac{A_1}{A_2} \right)^2}$$

$$v_2 = \sqrt{10^2 \left(\frac{20}{10} \right)^2} = \frac{10 \cdot 20}{10} = 20 \text{ m/s}$$

$$c) \quad v_2 = \sqrt{10^2 \left(\frac{20}{10} \right)^2} = \frac{10 \cdot 20}{10} = 20 \text{ m/s}$$

Ex 11

$$a) \quad v_1 = \sqrt{10} \implies v_2 = \sqrt{10} \cdot \frac{v_1}{v_2} \implies v_2 = \sqrt{10} \cdot \frac{10}{10} = 10 \text{ m/s}$$

From Eqs. (1-28) and (1-29):

$$L_1 \cos \alpha = L_2 L_3 e^{-i\alpha} e^{i\theta} e^{i\phi}$$

$$L_2 \cos \alpha = \frac{L_1}{L_3} (L_3 \cos \alpha + L_3 \sqrt{\frac{L_3^2}{L_1^2} - \cos^2 \alpha}) e^{-i\alpha} e^{i\theta} e^{i\phi}$$

where $L_3 = L_1 \sin \alpha = L_3 \sqrt{\frac{L_3^2}{L_1^2} - \cos^2 \alpha}$,

$$\alpha = \sin^{-1} \left(\frac{L_3}{L_1} \right) \cos \theta = \alpha$$

$$L_3 = \frac{L_1 L_2 \sin \alpha}{L_1 \cos \alpha + L_3 \left(\frac{L_3}{L_1} \right) \cos \theta} \quad \text{from Eq. (1-28)}$$

(1) $(L_3)_{\alpha=0} = \frac{1}{2} (L_3)_{\alpha=0} = 0$

Ex-11 Given $\theta = \alpha$ then $\alpha = 0$ or $\alpha = \pi$, and $\alpha = \pi$

(a) From Eq. (1-28): $(L_3)_{\alpha=0} = L_3$

(b) From Eq. (1-29): $(L_3)_{\alpha=0} = L_2 L_3$

(c) $L_1 \cos \alpha = L_2 L_3 \cos \alpha \left[1 + \frac{L_3}{L_1} (\cos \alpha + \cos \theta) \right]$

$$L_1 \cos \alpha = L_2 L_3 e^{-i\alpha} \cos \alpha \left[1 + \frac{L_3}{L_1} \cos \alpha \right]$$

$$= L_2 L_3 e^{-i\alpha} \cos \alpha \left[1 + \frac{L_3}{L_1} \cos \alpha \right]$$

where $\alpha = \sin^{-1} \left(\frac{L_3}{L_1} \right) \cos \theta = \alpha$ when $\theta = \alpha$.

Ex-12 (a) $\alpha = \sin^{-1} \sqrt{\frac{L_3}{L_1}} = \sin^{-1} \sqrt{\frac{L_3}{L_1}} \cos \theta$

(b) $\alpha = \sin^{-1} \frac{L_3}{L_1}$, $\sin \alpha = \frac{L_3}{L_1} \cos \theta$ where $\cos \theta = \frac{L_1}{L_3}$

$$L_1 = \frac{L_1 L_2 \sin \alpha \cos \theta}{L_3 \cos \alpha \cos \theta} = e^{i\theta} = e^{i\theta}$$

(c) $L_1 = \frac{L_1 L_2 \sin \alpha}{L_3 \cos \alpha \cos \theta} = L_2 e^{i\theta} \cos \theta$

(d) The transmitted wave is in phase as $e^{-i\alpha} e^{i\theta} e^{i\phi}$

where $\alpha = \sin^{-1} \left(\frac{L_3}{L_1} \right) \cos \theta = \frac{L_3}{L_1} \cos \theta$

Attenuation is in the direction of

$$= 2 \sin \alpha e^{-i\alpha} = 2 \sin \alpha \cos \theta$$

Ex. 20 When the incident light first strikes the AgBr-coated surface, $\theta_1 = \theta_2 = 0$, $\tau = \sqrt{\frac{2\mu_0 \mu_1}{\mu_0 + \mu_1}}$.

$$\frac{dR_{\text{AgBr}}}{dR_{\text{AgCl}}} = \frac{\mu_1}{\mu_0} \tau^2 = \frac{\mu_1^2 \mu_0}{\mu_0^2 + \mu_1^2}$$

Total reflection never occurs the point at which grazing incidence occurs

$$\theta_1 = \theta_2^* = \theta_2 = \sin^{-1}\left(\frac{\mu_0}{\mu_1}\right) = 90^\circ$$

On exit from the prism, $\theta_2 = \frac{\mu_0}{\mu_1} \theta_1$.

$$\frac{dR_{\text{AgBr}}}{dR_{\text{AgCl}}} = \frac{\mu_1}{\mu_0} \tau^2 = \frac{\mu_1^2 \mu_0}{\mu_0^2 + \mu_1^2}$$

$$\therefore \frac{dR_{\text{AgBr}}}{dR_{\text{AgCl}}} = \left[\frac{\mu_1^2 \mu_0}{\mu_0^2 + \mu_1^2} \right]^2 = \left[\frac{\mu_1^2}{\mu_0 + \mu_1^2} \right]^2 = 0.25$$

Ex. 21 (a) $n_2 \sin \theta_2 = n_1 \sin(\theta_2' + \theta_1) = n_1 \sin \theta_1$

$$= n_1 \sqrt{1 - \cos^2 \theta_1} = n_1 \sqrt{1 - n_2^2 \sin^2 \theta_2} = n_1 \sqrt{1 - n_2^2}$$

$$\sin \theta_2 = \frac{n_1}{n_2} \sqrt{1 - n_2^2} = \sqrt{1 - n_2^2}, \quad (\theta_1 = 0)$$

$$(b) \quad n_1 \sin \theta_1 = n_2 \sin \theta_2 = \sqrt{1 - n_2^2} = 0.7546$$

$$\theta_1 = \sin^{-1}(0.7546) = 49.04^\circ$$

Ex. 22 $R_p(\theta = 0) = \frac{n_2}{n_1} = \frac{1.50}{1.00}$

$$(a) \quad R_p(\theta = 90^\circ) \tau_1 = \frac{(n_1 \mu_1 \cos \theta_1 + n_2 \mu_2) (n_1 \mu_1 \cos \theta_1 - n_2 \mu_2)}{(n_1 \mu_1 \cos \theta_1 + n_2 \mu_2) (n_1 \mu_1 \cos \theta_1 + n_2 \mu_2)} = \frac{n_1^2 \mu_1^2 \cos^2 \theta_1 - n_2^2 \mu_2^2}{n_1^2 \mu_1^2 \cos^2 \theta_1 + n_2^2 \mu_2^2}$$

$$= \frac{1.50^2 (1.00)^2 \cos^2 \theta_1 - 1.00^2 (1.50)^2}{1.50^2 (1.00)^2 \cos^2 \theta_1 + 1.00^2 (1.50)^2}$$

$$R_p(\theta = 90^\circ) \tau_1 = \frac{1.50^2 (1.00)^2 \cos^2 \theta_1 - 2.25}{1.50^2 (1.00)^2 \cos^2 \theta_1 + 2.25}$$

$$(b) \quad R_p(\theta = 90^\circ) \tau_1 = \frac{(n_1 \mu_1 \cos \theta_1 - n_2 \mu_2) (n_1 \mu_1 \cos \theta_1 + n_2 \mu_2)}{(n_1 \mu_1 \cos \theta_1 + n_2 \mu_2) (n_1 \mu_1 \cos \theta_1 + n_2 \mu_2)} = \frac{n_1^2 \mu_1^2 \cos^2 \theta_1 - n_2^2 \mu_2^2}{n_1^2 \mu_1^2 \cos^2 \theta_1 + n_2^2 \mu_2^2}$$

$$= \frac{1.50^2 (1.00)^2 \cos^2 \theta_1 - 2.25}{1.50^2 (1.00)^2 \cos^2 \theta_1 + 2.25}$$

$$R_p(\theta = 90^\circ) \tau_1 = \frac{n_1^2 \mu_1^2 \cos^2 \theta_1 - n_2^2 \mu_2^2}{n_1^2 \mu_1^2 \cos^2 \theta_1 + n_2^2 \mu_2^2} = \frac{1.50^2 (1.00)^2 \cos^2 \theta_1 - 2.25}{1.50^2 (1.00)^2 \cos^2 \theta_1 + 2.25}$$

Ex-11) a) For perpendicular polarization and $\mu_1 = \mu_2 = \mu_0$

$$\sin \theta_{cp} = \frac{1}{\sqrt{1 + \frac{\epsilon_2}{\epsilon_1}}}$$

Under condition of no-reflection:

$$\begin{aligned} \cos \theta &= \sqrt{1 - \frac{\epsilon_2}{\epsilon_1} \sin^2 \theta_{cp}} = \frac{1}{\sqrt{1 + \frac{\epsilon_2}{\epsilon_1}}} \\ &= \sin \theta_{cp} \implies \theta_i = \theta_{cp} = \theta_{tr} \end{aligned}$$

b) For parallel polarization and $\mu_1 = \mu_2 = \mu_0$

$$\sin \theta_{cp} = \frac{1}{\sqrt{1 + \frac{\epsilon_2}{\epsilon_1}}}$$

$$\begin{aligned} \cos \theta &= \sqrt{1 - \frac{\epsilon_2}{\epsilon_1} \sin^2 \theta_{cp}} = \frac{1}{\sqrt{1 + \frac{\epsilon_2}{\epsilon_1}}} \\ &= \sin \theta_{cp} \implies \theta_i = \theta_{cp} = \theta_{tr} \end{aligned}$$

Ex-12) a) $\sin \theta_i = \sqrt{\frac{\epsilon_2}{\epsilon_1}}$; $\sin \theta_{tr} = \frac{1}{\sqrt{1 + \frac{\epsilon_2}{\epsilon_1}}}$



$$\implies \sin \theta_{cp} = \sqrt{\frac{\epsilon_2}{\epsilon_1}}$$

$$\therefore \sin \theta_i = \sin \theta_{cp} \quad (\theta_i > \theta_{cp})$$

b) Let $n_1/n_2 = n$.



Ex-13) a) For perpendicular polarization:

$$r_{\perp} = \frac{E_{1\perp} - E_{2\perp}}{E_{1\perp} + E_{2\perp}} = \frac{E_{1\perp} \cos \theta_i - E_{2\perp} \cos \theta_r}{E_{1\perp} \cos \theta_i + E_{2\perp} \cos \theta_r}$$

$$\sin \theta_i = \sqrt{\frac{\epsilon_2}{\epsilon_1}} \sin \theta_r \implies \cos \theta_r = \sqrt{1 - \frac{\epsilon_2}{\epsilon_1}} \cos \theta_i$$

$$r_{\perp} = \frac{E_{1\perp} \cos \theta_i - E_{2\perp} \sqrt{1 - \frac{\epsilon_2}{\epsilon_1}} \cos \theta_i}{E_{1\perp} \cos \theta_i + E_{2\perp} \sqrt{1 - \frac{\epsilon_2}{\epsilon_1}} \cos \theta_i}$$

$$r_{\perp} = \frac{E_{1\perp} \cos \theta_i}{E_{1\perp} \cos \theta_i + E_{2\perp} \sqrt{1 - \frac{\epsilon_2}{\epsilon_1}} \cos \theta_i} = \frac{E_{1\perp} \cos \theta_i}{E_{1\perp} \cos \theta_i + E_{2\perp} \sqrt{1 - \frac{\epsilon_2}{\epsilon_1}} \cos \theta_i}$$

For parallel polarization:

$$G = \frac{\frac{E_0 \cos \theta_i \cos \theta_t}{\mu_0} - \frac{E_0 \cos \theta_i \cos \theta_r}{\mu_0}}{\frac{E_0 \cos \theta_i \sin \theta_t}{\mu_0} + \frac{E_0 \cos \theta_i \sin \theta_r}{\mu_0}}$$

$$G = \frac{1 - \frac{\mu_2 \cos \theta_t}{\mu_1 \cos \theta_r}}{1 + \frac{\mu_2 \sin \theta_t}{\mu_1 \sin \theta_r}}$$

② $n_2/n_1 = 2.10$, $\theta_i = \theta_t = 45^\circ \rightarrow G = 0.17 \frac{\mu_2}{\mu_1} = 0.35$



Ex-21



$$\vec{E}_i \cdot \hat{n} = E_0 \cos \theta_i \hat{n} = E_0 \cos \theta_i \hat{z}$$

$$\vec{E}_r \cdot \hat{n} = E_r \cos \theta_r = E_r \cos \theta_r \hat{z}$$

$$\vec{E}_t \cdot \hat{n} = \frac{1}{2} E_t \cos \theta_t$$

$$= \frac{1}{2} E_t \cos \theta_t \hat{z} \quad (\text{since } \hat{n} = \cos \theta_t \hat{z} + \sin \theta_t \hat{x})$$

$$E_0 \cos \theta_i = E_r \cos \theta_r + \frac{1}{2} E_t \cos \theta_t$$

a) From the continuity:

$$E_0 = E_r + \frac{1}{2} E_t$$

$$\text{where } E_r/E_0 = \frac{\mu_2 \cos \theta_i}{\mu_1 \cos \theta_r} = 0.667 \frac{\mu_2}{\mu_1} =$$

$$\vec{E}_i \cdot \hat{n} = E_0 \cos \theta_i \hat{n} = E_0 \cos \theta_i \hat{z}$$

$$\vec{E}_r \cdot \hat{n} = \frac{1}{2} E_r \cos \theta_r = \frac{1}{2} E_r \cos \theta_r \hat{z} \quad (\text{since } \hat{n} = \cos \theta_r \hat{z} + \sin \theta_r \hat{x})$$

b) From Eq. (8-110) $\sin \theta_1 = \frac{\beta \sin \theta_2}{\beta_1 - \beta_2}$ (Complex).

$$\cos \theta_1 = \sqrt{1 - \sin^2 \theta_1} \quad (\text{Complex}).$$

The x - and y -components of \vec{E}_2 in Eq. (8) above are different amplitudes and are out of phase, indicating that it is elliptically polarized.

8-112

$$a) \Gamma_{\parallel} = \frac{Z_2 \cos \theta_1}{Z_1 \cos \theta_2} \Big|_{\text{av}} = \frac{Z_2 \cos \theta_1}{Z_2 \cos \theta_2} = \frac{Z_2}{Z_2} = \Gamma_{\parallel} = \frac{Z_2 \cos \theta_1 - Z_1 \cos \theta_2}{Z_2 \cos \theta_1 + Z_1 \cos \theta_2}.$$

$$\tau_{\parallel} = \frac{2Z_2 \cos \theta_1}{Z_1 \cos \theta_2} \Big|_{\text{av}} = \frac{2Z_2 \cos \theta_1}{Z_2 \cos \theta_2} = \tau_{\parallel}(\cos \theta_1) = \frac{2Z_2 \cos \theta_1}{Z_2 \cos \theta_1 + Z_1 \cos \theta_2}.$$

b) From part a) we have

$$\Gamma + \Gamma_{\parallel}^* = \tau_{\parallel}$$

This compares with

$$\Gamma + \Gamma_{\parallel} = \tau_{\parallel} \left(\frac{Z_1 \cos \theta_1}{Z_2 \cos \theta_2} \right) \text{ in Eq. (8-110).}$$

Chapter 9

Theory and Application of Transmission Lines

Ex.1

$$P \cdot Z = \begin{vmatrix} Z_1 & Z_2 & Z_3 \\ Z_1 & Z_2 & Z_3 \\ Z_3 & * & * \end{vmatrix} = Z_1 Z_2 Z_3 \implies \frac{P}{Z_1} = Z_2$$

$$P \cdot Z = \begin{vmatrix} Z_1 & Z_2 & Z_3 \\ Z_1 & Z_2 & Z_3 \\ * & Z_3 & * \end{vmatrix} = Z_1 Z_2 Z_3 \implies \frac{P}{Z_2} = Z_1$$

Ex.2

a) $P = (Z_1 Z_2 + Z_2 Z_3) = \rho \sin(\theta_1 \theta_2 + \theta_2 \theta_3)$

$$\implies \begin{cases} \frac{P}{Z_2} = \rho \sin \theta_1 \\ \frac{P}{Z_1} = \rho \sin \theta_3 \\ \frac{P}{Z_3} = \rho \sin \theta_2 \end{cases}$$

b) $P = (Z_1 Z_2 + Z_2 Z_3) = \rho \sin(\theta_1 \theta_2 + \theta_2 \theta_3)$

$$\implies \begin{cases} \frac{P}{Z_2} = \rho \sin \theta_1 \\ \frac{P}{Z_1} = \rho \sin \theta_3 \\ \frac{P}{Z_3} = \rho \sin \theta_2 \end{cases}$$

From (1) and (2) $\frac{P}{Z_1 Z_2} = \rho \sin \theta_3$ (3)

From (1) or (3) $\frac{P}{Z_2} = \rho \sin \theta_1$ (4)

From (2) or (3) $\frac{P}{Z_1} = \rho \sin \theta_3$ (5)

(3) From (4) $\frac{P}{Z_1 Z_2} = \frac{P}{Z_2} \sin \theta_1$ (6)

From (4), (5), and (6) $\frac{P}{Z_1} = \frac{P}{Z_2} \sin \theta_1 \implies \frac{P}{Z_1} = \frac{P}{Z_2} \sin \theta_1$ (7)

Combining (4) and (7), we have $\frac{P}{Z_2} = \frac{P}{Z_1} \sin \theta_1$

Similarly, $\frac{P}{Z_3} = \frac{P}{Z_2} \sin \theta_2$

Ex.3

$Z_1 = \frac{Z_0}{\cos \theta}$

a) $Z_1 = \frac{Z_0}{\cos \theta} = \frac{Z_0}{\cos \theta} \implies \theta = \cos^{-1} \frac{Z_0}{Z_1}$

b) $Z_1 = \frac{Z_0}{\cos \theta} = \frac{Z_0}{\cos \theta} \implies \theta = \cos^{-1} \frac{Z_0}{Z_1}$

$$d) \quad \xi_2 = \frac{1-i}{2}\sqrt{2} = \frac{1}{2}\sqrt{2} \longrightarrow \text{w/o } i \text{ in }.$$

$$e) \quad \eta_1 = \frac{1}{\sqrt{2}} \longrightarrow \begin{array}{l} \eta_{21} = \eta_1 \cdot i \text{ for } \cos \frac{\pi}{4}, \\ \eta_{22} = \eta_1 \cdot (-i) \text{ for } \sin \frac{\pi}{4}, \\ \eta_{23} = \eta_1 \text{ for } \cos \frac{3\pi}{4}. \end{array}$$

Prob 10 Given: $\xi_1 = \cos \frac{\pi}{4} + i \sin \frac{\pi}{4}$, $\xi_2 = \cos \frac{3\pi}{4} + i \sin \frac{3\pi}{4}$, $\xi_3 = \cos \frac{5\pi}{4} + i \sin \frac{5\pi}{4}$
 Using de Moivre's rule: $\xi_1^2 = \xi_2$, $\xi_2^2 = \xi_3$, $\xi_3^2 = \xi_1$
 $f = 2 + \cos^2 \frac{\pi}{4}$

$$a) \quad \xi_1 = \frac{1}{2} \sqrt{\frac{2+2i}{2}} = \frac{1+i}{2} \quad (\text{valid})$$

$$\xi_2 = \frac{1}{2} \sqrt{\frac{2-2i}{2}} = \frac{1-i}{2} \quad (\text{valid})$$

$$\xi_3 = \frac{1}{2} \sqrt{\frac{2+2i}{2}} = \frac{1+i}{2} \quad (\text{valid})$$

$$\xi_4 = \frac{1}{2} \sqrt{\frac{2-2i}{2}} = \frac{1-i}{2} \quad (\text{valid}).$$

$$b) \quad \frac{\xi_1 \xi_2}{\xi_3 \xi_4} = \frac{\sqrt{2+2i}}{\sqrt{2-2i}} = \frac{2+2i}{2} = 1+i$$

$$c) \quad \cos \frac{\pi}{4} = \frac{1}{\sqrt{2}} \text{ and } \sin \frac{\pi}{4} = \frac{1}{\sqrt{2}} \text{ and } \cos \frac{3\pi}{4} = \frac{-1}{\sqrt{2}}$$

$$\Rightarrow \xi_1 = \frac{1}{\sqrt{2}} \left[1 + \frac{1}{\sqrt{2}} \left(\frac{1}{\sqrt{2}} + \frac{1}{\sqrt{2}} i \right) \right] = \frac{1}{2} (1 + 1 + i + i) = \frac{1}{2} (2 + 2i) = 1+i$$

$$\xi_2 = \frac{1}{\sqrt{2}} \left[1 + \frac{1}{\sqrt{2}} \left(\frac{-1}{\sqrt{2}} + \frac{1}{\sqrt{2}} i \right) \right] = \frac{1}{2} (1 - 1 + i + i) = i$$

Prob 11 Solving for ξ_1, ξ_2 and ξ_3 and

$$\xi_1 \xi_2 = \xi_1^2 + \xi_2^2$$

$$\xi_1 \xi_3 = \xi_1^2 + \xi_3^2$$

into Eq. (1) and (2) we have

$$\xi_1 (\xi_2 - \xi_3) = (\xi_1^2 + \xi_2^2) - (\xi_1^2 + \xi_3^2) = \xi_2^2 - \xi_3^2$$

which implies

$$\xi_1 \xi_2 = (\xi_2 + \xi_3) \xi_2 = \xi_2^2 + \xi_2 \xi_3$$

$$\xi_1 \xi_3 = (\xi_2 + \xi_3) \xi_3 = \xi_2 \xi_3 + \xi_3^2$$

$$\therefore \frac{\xi_1}{\xi_2} = -\frac{\xi_3}{\xi_2} = \frac{\xi_3 + \xi_2}{\xi_2}$$

$$\text{Subst. } \xi_1 = \xi_2 \cdot \frac{\xi_3 + \xi_2}{\xi_2} = \xi_2 (1 + \frac{\xi_3}{\xi_2}) = \xi_2 (1 + \frac{\xi_3}{\xi_2})$$

$$\text{From Eq. (1) we have } \xi_2 = \xi_2 + \xi_3 = \xi_2 \sqrt{2} \left(1 + \frac{\xi_3}{\xi_2} \right)^2$$

Squaring both sides, we obtain two equations (one
 for real and imaginary parts):

$$\begin{aligned} x^2 + y^2 &= -\frac{a^2}{2} \\ 2xy &= \frac{a^2}{2} \end{aligned}$$

From which Eqs. (17-214) and (17-215) follow.

$$\begin{aligned} \text{E21.1} \quad x &= j \sqrt{2} (1-j \frac{a}{2})^n (1+j \frac{a}{2})^n \\ &= j \sqrt{2} \left[(1-j \frac{a}{2})^n + j \left(\frac{a}{2} \right)^n + \frac{a^n}{2} \right] \\ &\quad + (1+j \frac{a}{2})^n + j \left(\frac{a}{2} \right)^n + \frac{a^n}{2} = a^n j \end{aligned}$$

$$\begin{aligned} \text{---} \quad a &= \frac{1}{\sqrt{2}} \left(\frac{a}{2} + \frac{a}{2} \right)^n + \frac{a}{2} \left(\frac{a}{2} - \frac{a}{2} \right)^n \\ &\quad + \frac{a}{2} \sqrt{2} \left[1 + \frac{a}{2} \left(\frac{a}{2} - \frac{a}{2} \right)^n \right] \end{aligned}$$

$$\begin{aligned} \text{E21.2} \quad x &= \sqrt{2} (1-j \frac{a}{2})^n (1+j \frac{a}{2})^n \\ &= \sqrt{2} \left[1 + \frac{a}{2} \left(\frac{a}{2} - \frac{a}{2} \right)^n + \frac{a}{2} \left(\frac{a}{2} + \frac{a}{2} \right)^n \right] = a \sqrt{2} \end{aligned}$$

$$\text{---} \quad a = \sqrt{2} \left[1 + \frac{a}{2} \left(\frac{a}{2} - \frac{a}{2} \right)^n + \frac{a}{2} \right]$$

$$a = \frac{1}{\sqrt{2}} \sqrt{2} \left(\frac{a}{2} \right)^n$$

$$a = \frac{1}{\sqrt{2}} + \frac{a}{2} \left[1 + \left(\frac{a}{2} - \frac{a}{2} \right)^n \right]$$

$$\text{E21.3} \quad x = \sqrt{2} (1-j \frac{a}{2})^n (1+j \frac{a}{2})^n = a \sqrt{2}$$

$$\text{---} \quad a = \sqrt{2} \left[1 + \frac{a}{2} \left(\frac{a}{2} - \frac{a}{2} \right)^n \right], \quad a = \frac{1}{\sqrt{2}} \sqrt{2} \left(\frac{a}{2} \right)^n$$

$$a = \sqrt{2} \sqrt{2} = \sqrt{2} (1-j \frac{a}{2})^n (1+j \frac{a}{2})^n = a \sqrt{2}$$

$$\text{---} \quad a = \sqrt{2} \left[1 + \frac{a}{2} \left(\frac{a}{2} - \frac{a}{2} \right)^n \right]$$

$$a = \frac{1}{\sqrt{2}} \sqrt{2} \left(\frac{a}{2} \right)^n$$

$$\text{E21.4} \quad a = \sqrt{2} \sqrt{2} = \sqrt{2} (1-j \frac{a}{2})^n (1+j \frac{a}{2})^n = a \sqrt{2}$$

From Eqs. (17-214), (17-215) and (17-216) $a = \frac{1}{\sqrt{2}} \sqrt{2} = \frac{1}{\sqrt{2}} \sqrt{2} = 1$, $a = \sqrt{2} \sqrt{2} = 2$, $a = \sqrt{2} \sqrt{2} = 2$

Given $Z_0 = 20 + j10 \text{ } \Omega$,
 Z is a real constant impedance (resistor),
 $\beta = 0.2 \text{ rad/m}$ (assumed),
 $l = 10^3 \text{ m}$.

$$Z_1 = Z_0 = 20 + j10 \text{ } \Omega \text{ (at } l \text{)}, \quad I_1 = \frac{V_1}{Z_0} = 0.20 \text{ (assumed)}$$

$$Z_2 = Z_0^* = 20 - j10 \text{ } \Omega \text{ (at } 0 \text{)}, \quad I_2 = \frac{V_2}{Z_0} = 0.20 \text{ (assumed)}$$

Ex. 10.10 (a) For lossless transmission line:

$$Z_1 = \sqrt{\frac{L}{C}} = \frac{1}{\beta} \sqrt{\frac{L}{C}} \text{ (at } l \text{)} = \frac{1}{0.2} \sqrt{\frac{10^{-7}}{10^{-10}}} = 25 \text{ } \Omega$$

$$\frac{V_1}{Z_1} = 0.20 \text{ } \longrightarrow V_1 = 0.20 \times 25 = 5 \text{ V}$$

(b) For matched transmission line:

$$Z_1 = \sqrt{\frac{L}{C}} = \frac{1}{\beta} \sqrt{\frac{L}{C}} = \frac{1}{0.2} \sqrt{\frac{10^{-7}}{10^{-10}}} = 25 \text{ } \Omega$$

$$\frac{V_1}{Z_1} = 0.20 \text{ } \longrightarrow V_1 = 0.20 \times 25 = 5 \text{ V}$$

$$\text{Ex. 10.11 } (P_{\text{in}})_1 = (P_{\text{out}})_1 = \int_0^l P_{\text{av}}(z) dz \quad V_1 = \sqrt{\frac{2}{\rho}} \frac{V_0}{\beta}$$

$$= \frac{1}{\rho} \frac{V_0^2}{\beta} \int_0^l dz \quad I_1 = \sqrt{\frac{2}{\rho}} \frac{I_0}{\beta}$$

$$\text{To maximize } (P_{\text{in}})_1, \text{ set } \frac{d(P_{\text{in}})_1}{dV_0} = 0, \quad \left. \begin{array}{l} \text{and } \frac{d(P_{\text{in}})_1}{dI_0} = 0 \end{array} \right\} \begin{array}{l} V_1 = V_0 \\ I_1 = I_0 \end{array}$$

$$\text{Max. } (P_{\text{in}})_1 = \frac{V_0^2}{\rho} = (P_{\text{in}})_0$$

\longrightarrow Max. power-transfer efficiency = 100%.

$$\text{Ex. 10.12 } \text{RTN} = V_1^* V_2^* = V_1^* V_1^*$$

$$\text{IITN} = I_1^* I_2^* = I_1^* I_1^*$$

$$\text{At } z=0: \text{RTN} = V_1^* = V_1^* + V_1^*, \quad \text{IITN} = I_1^* = I_1^* + I_1^* = \frac{1}{2} I_1^* (2)$$

$$\longrightarrow V_1^* = \frac{1}{2} V_1 = 2.5 \text{ V}, \quad I_1^* = \frac{1}{2} I_1 = 2.5 \text{ A}$$

$$\text{(a) RTN} = \frac{1}{2} V_1 = 2.5 \text{ V} \times 2.5 \text{ A} = 6.25 \text{ W}$$

$$\text{IITN} = \frac{1}{2} I_1 = 2.5 \text{ A} \times 2.5 \text{ A} = 6.25 \text{ W}$$

$$\text{(b) RTN} = V_1 \text{ and } I_1 = 2.5 \text{ A and } V_1 = 2.5 \text{ V}$$

$$\text{IITN} = I_1 \text{ and } I_1 = 2.5 \text{ A and } I_1 = 2.5 \text{ A}$$

Ex. 10 From Eq. (1) and (2) $x = \frac{1}{2}z + \frac{1}{2}(y-z)$
 $= (\frac{1}{2} + \frac{1}{2})z = \frac{1}{2}z$. (3)

Also $y = z + (y-z) = \frac{1}{2}z + y$
 $= \frac{1}{2}z + (1 + \frac{1}{2}z)z$. (4)

Substituting (3) in (2):
 $y = (1 + \frac{1}{2}z)z + z(1 + \frac{1}{2}z)z$. (5)

(3) Letting $z=1$, $x_1 = \frac{1}{2}$, and $x_2 = y = \frac{3}{2}$ in Eq. (1) and (2) and (3) and (4) and (5):

$$y = 3 = (1 + \frac{1}{2})z_1 + \frac{1}{2}z_1^2 = \frac{3}{2}z_1 + \frac{1}{2}z_1^2$$
 (6)

$$z = 3 = (\frac{1}{2} + \frac{1}{2}z_1)z_1 = (\frac{1}{2}z_1 + \frac{1}{2}z_1^2)z_1$$
 (7)

Both Eqs. (6) & (7) and Eqs. (3) & (4) are of the following form:
$$\begin{bmatrix} y \\ z \end{bmatrix} = \begin{bmatrix} a & b \\ c & d \end{bmatrix} \begin{bmatrix} z \\ z^2 \end{bmatrix}$$
 (8)

where $a = \frac{1}{2}$, $b = \frac{1}{2}$, $c = \frac{3}{2}$, $d = \frac{1}{2}$. (9)

and $b = \frac{1}{2}$, $d = \frac{1}{2}$, $c = \frac{3}{2}$. (10)

and $c = \frac{3}{2}$, $d = \frac{1}{2}$. (11)

———— $AD - BC = \frac{1}{2} \cdot \frac{1}{2} - \frac{3}{2} \cdot \frac{1}{2} = \frac{1}{4} - \frac{3}{4} = -\frac{2}{4} = -\frac{1}{2}$

(1) (8) is the required result in Eq. (1) and (2).
 Eq. (2) and (3) can be obtained by using (8) in (2):
 $x = \frac{1}{2}(z + (1 - \frac{1}{2}z)z) = \frac{1}{2}z$

Ex. 11 $x = \frac{1}{2}z + \frac{1}{2}z = z$, $y = \frac{1}{2}z + \frac{1}{2}z = z$.

$$\begin{cases} \frac{1}{2}z + \frac{1}{2}z = 2z \\ \frac{1}{2}z + \frac{1}{2}z = 2z \end{cases}$$

(1) $x = \frac{1}{2}z + \frac{1}{2}z = z$

$y = \frac{1}{2}z + \frac{1}{2}z = z$, $z = \sqrt{2}$

$$\frac{x}{\sqrt{2}} = \frac{y}{\sqrt{2}} = z = \sqrt{2}$$

$$\begin{aligned} \text{We have } V(t) &= \int (v_1 + v_2) e^{at} dt = \int (v_1 + v_2) e^{at} dt \\ Z(t) &= \int \left(\frac{v_1}{\lambda_1} + v_2 \right) e^{at} dt = \int \left(\frac{v_1}{\lambda_1} + v_2 \right) e^{at} dt \end{aligned}$$

$$\text{where } v_1 = \frac{b_1}{\lambda_1 - \lambda_2} v_2 \quad \text{and } v_2 = \frac{b_2}{\lambda_2 - \lambda_1}$$

a) For an infinite line, $\lambda_1 = \lambda_2$:

$$V(t) = \frac{b_1}{\lambda_1 - \lambda_2} v_2 e^{at}, \quad Z(t) = \frac{v_2}{\lambda_1 - \lambda_2} e^{at}$$

b) For a finite line of length L terminated in Z_L :

$$Z_L = Z_0 \frac{Z_L + Z_0 \tanh \beta L}{Z_0 + Z_L \tanh \beta L}$$

Ex. 11.11 Distributed line: $Z_0 = \sqrt{\frac{L}{C}} = 50 \Omega$, $\beta = 0.02 \text{ rad/m}$

$$\cos \left(\frac{\beta L}{2} \right) = \cos \left(\frac{0.02}{2} \right) = 0.9999$$

$$\Rightarrow \frac{Z_L + Z_0 \tanh \beta L}{Z_0 + Z_L \tanh \beta L} = 0.9999 \Rightarrow Z_L = \frac{Z_0}{0.9999}$$

$$Z_L = \frac{50 \Omega}{0.9999} = 50.005 \Omega, \quad C = \frac{L}{Z_0^2} = 0.04 \text{ F/m}$$

$$v = \frac{1}{\sqrt{LC}} = 0.5 \text{ m/s}, \quad \beta = \omega \sqrt{LC} = 0.02 \text{ rad/m}$$

$$\Rightarrow V(t) = \frac{V_0}{\sqrt{LC}} e^{-\alpha t} e^{i\omega t} = \frac{100 \text{ V}}{\sqrt{0.04}} e^{-0.02 t} e^{i 200 \pi t}, \quad Z(t) = \frac{V_0}{Z_0}$$

$$\therefore \text{Re}\{Z\} = 50 e^{-0.02 t} \cos(200 \pi t - 0.02 t) \quad (1)$$

$$\text{Im}\{Z\} = 50 e^{-0.02 t} \sin(200 \pi t - 0.02 t) \quad (2)$$

$$\text{b) } Z_L = 50.005 \Omega \Rightarrow \text{Re}\{Z_L\} = 50.005 \cos(0.02) \quad (3)$$

$$\text{Im}\{Z_L\} = 50.005 \sin(0.02) \quad (4)$$

$$\text{c) } \text{Re}\{Z_L\} = \frac{1}{2} (Z_L + Z_L^*) = \frac{1}{2} (50.005 \cos(0.02) + 50.005 \cos(0.02)) \quad (5)$$

Ex. 11.12 From Eq. (11-113) $Z_L = Z_0 \tanh \beta L = Z_0 \tanh \beta L$

From Eqs. (11-113) and (11-112) $\Gamma = \frac{Z_L - Z_0}{Z_L + Z_0} = \frac{Z_0 \tanh \beta L - Z_0}{Z_0 \tanh \beta L + Z_0}$

$$\therefore Z_L = (1 + \Gamma) Z_0$$

$$\text{d) From Eq. (11-113) } Z_L = Z_0 \cosh \beta L = \frac{Z_0}{2} \left(e^{\beta L} + e^{-\beta L} \right)$$

Ex 2.11 a) From Eq. (9-100) $Z_{in} = Z_0$, then $V_{in} = Z_0 \frac{I_{in} e^{-\beta z}}{1 + \frac{Z_0}{Z_L} e^{-2\beta z}}$.

For $Z_L = Z_0$, $\beta L = \pi$, $z = 0$,

$$Z_{in} = Z_0 \frac{I_{in} e^{-\beta z}}{1 + \frac{Z_0}{Z_L} e^{-2\beta z}} = Z_0 \frac{I_{in} e^{-\beta(0)}}{1 + \frac{Z_0}{Z_0} e^{-2\beta(0)}} \\ = 4 Z_0 I_{in}.$$

b) From Eq. (9-100) $Z_{in} = Z_0$, then $V_{in} = Z_0 \frac{I_{in} e^{-\beta z}}{1 + \frac{Z_0}{Z_L} e^{-2\beta z}}$.

For $Z_L = jZ_0$, $Z_{in} = Z_0 \frac{I_{in} e^{-\beta z}}{1 + \frac{Z_0}{Z_L} e^{-2\beta z}} = Z_0 \frac{I_{in} e^{-\beta(0)}}{1 + \frac{Z_0}{jZ_0} e^{-2\beta(0)}}$
 $= Z_0 I_{in} j.$

Ex 2.12 $\beta L = \frac{2\pi}{\lambda} L = \frac{\pi}{2} = 1.57$

then $\beta L = \tan^{-1} \beta L = -1.107$.

$$Z_{in} = Z_0 \frac{Z_L + jZ_0 \tan \beta L}{Z_0 + jZ_L \tan \beta L} = Z_0 \frac{(1 + j \tan \beta L)}{1 - \tan^2 \beta L} \\ = 2.61 - j1.887 \text{ } \Omega.$$

Ex 2.13 a) From $Z_{in} = Z_0$, then $V_{in} = 200 \cos \beta z$ (V).

$I_{in} = \frac{V_{in}}{Z_0} \cos \beta z = 1 \cos \beta z$ (A).

b) $Z_{in} = \sqrt{Z_0 Z_L} = 100 \cos \beta L = 100 \cos 1.57$ (A).

then $V_{in} = \sqrt{\frac{Z_0}{Z_L}} = 40 \cos \beta z = 40 \cos \beta(0) = 40 \cos(0) = 40$ (V).

$L = 0.5 \lambda$ then $\omega = 2\pi f$ (Hz),

$\beta = 2.517$ (rad/m).

c) $Z_{in} = \sqrt{\frac{Z_0 Z_L}{1 + \frac{Z_0}{Z_L} e^{-2\beta L}}} = 7 = \sqrt{\frac{100 \times 100}{1 + \frac{100}{Z_L} e^{-2\beta L}}}$.

then $49 Z_L = 100 Z_0$, $49 = \frac{100}{Z_L}$.

$\omega = \beta L = 2.517 \times 0.5 \lambda = 1.2585 \lambda$ (rad/m).

we obtain $L = 0.25 \lambda$ (m), $f = 1.2585 \lambda$ (Hz),

$\omega = 2.517$ (rad/s), $C = 3 \times 10^8$ (m/s).

P. 2.22 (a) Since the line is very short compared to λ , we can neglect β , and thus use $\beta \approx \beta_0 = 1$ and $\beta \approx 1 - \beta_0^2$

$$\left. \begin{aligned} \beta &= \frac{\beta_0 \sqrt{1 - \beta_0^2}}{\beta_0} = \beta_0 \sqrt{1 - \beta_0^2} \\ \beta &= \frac{\beta_0 \sqrt{1 - \beta_0^2}}{\beta_0} = \beta_0 \sqrt{1 - \beta_0^2} \end{aligned} \right\} \beta_0 = \sqrt{\frac{1}{2}} = 70.71\% \\ \beta \approx 1 - \beta_0^2 \implies \beta_0 = \frac{1}{\sqrt{2}} = 0.707$$

(b) $\beta = \frac{\beta_0}{\gamma} = \beta_0 \sqrt{1 - \beta_0^2} = 0.41$ (table); $\beta \approx \frac{1 - \beta_0^2}{1 + \beta_0^2} = \frac{1 - 0.5}{1 + 0.5} = 0.33$

$$\therefore \beta_{\text{table}} = \beta_0 \sqrt{1 - \beta_0^2} = 0.41$$

$$\beta_0 = \beta_0 \sqrt{1 - \beta_0^2} = 0.41 \implies \beta_0 = 0.51$$

P. 2.23 From Eq. (2.11) $\beta_{\text{table}} = \beta_0 \sqrt{1 - \beta_0^2} = \beta_0 \frac{\beta_0 \sqrt{1 - \beta_0^2}}{\beta_0}$

$$= \beta_0 \frac{\beta_0 \sqrt{1 - \beta_0^2}}{\beta_0 \sqrt{1 - \beta_0^2}} \quad \text{--- (1)}$$

For a function like, $\beta_{\text{table}} = \beta_0 \frac{\beta_0 \sqrt{1 - \beta_0^2}}{\beta_0 \sqrt{1 - \beta_0^2}}$ --- (2)

At $\beta_{\text{table}} = \beta_0$, $\beta_{\text{table}} = \beta_0 \frac{\beta_0 \sqrt{1 - \beta_0^2}}{\beta_0 \sqrt{1 - \beta_0^2}}$ (continuous) --- (3)

When the frequency is slightly off resonance:

$$f = f_0 + \Delta f \quad (\Delta f > 0), \quad \beta_{\text{table}} \approx \beta_0 \frac{\beta_0 \sqrt{1 - \beta_0^2}}{\beta_0 \sqrt{1 - \beta_0^2}}$$

$$\text{and } \beta_{\text{table}} \approx \beta_0 \frac{\beta_0 \sqrt{1 - \beta_0^2}}{\beta_0 \sqrt{1 - \beta_0^2}} = \beta_0 \frac{\beta_0 \sqrt{1 - \beta_0^2}}{\beta_0 \sqrt{1 - \beta_0^2}}$$

--- (4) Indeed, after having set several order small terms:

$$\beta_{\text{table}} \approx \frac{\beta_0^2}{1 + \beta_0^2} \quad \text{--- (5)}$$

Combining (2) & (5) $\frac{\beta_0^2}{1 + \beta_0^2} = \frac{\beta_0 \sqrt{1 - \beta_0^2}}{\beta_0 \sqrt{1 - \beta_0^2}}$ --- (6)

Half-power point at: $\cos(\frac{\pi}{2}) = 0$; $\cos(\frac{\pi}{2}) = \frac{\beta_0^2}{1 + \beta_0^2}$ --- (7)

For $\Delta = 0$, $\beta_{\text{table}} = \beta_0 \frac{\beta_0 \sqrt{1 - \beta_0^2}}{\beta_0 \sqrt{1 - \beta_0^2}}$, and (7) is $\frac{\beta_0^2}{1 + \beta_0^2}$

which, for a low-loss transmission line, becomes (7)

$$\Delta f > 0 \implies \Delta f = \frac{\beta_0^2}{1 + \beta_0^2} \left(\frac{1}{\beta_0} - \frac{1}{\beta_0} \right) \implies \frac{\Delta f}{f_0} \approx \frac{\beta_0^2}{1 + \beta_0^2} \left(\frac{1}{\beta_0} - \frac{1}{\beta_0} \right)$$

$$\Delta = \frac{\beta_0^2}{1 + \beta_0^2} = \frac{\beta_0^2}{1 + \beta_0^2}$$

Ex. 11 (i) For a loaded quarter-wave line (see Fig. 11.11)

$$Z_L = \frac{Z_0}{j} = -\frac{Z_0}{j} = -\frac{Z_0^2}{jZ_0} = -\frac{Z_0^2}{jZ_0} = -jZ_0 \quad (1)$$

$$\rightarrow Z_L = \frac{jZ_0^2}{Z_0} \quad (2) \quad Z_L = -\frac{jZ_0^2}{Z_0} \quad (3)$$

(Resistive Z_L and capacitive reactance Z_L in series.)

Input impedance Z_i can also be expressed in terms of a resistance R_i and a capacitive reactance X_i in parallel:

$$Z_i = \frac{R_i X_i}{R_i + jX_i} = \frac{R_i X_i}{R_i + jX_i} = R_i - jX_i \quad (4)$$

Combining Eqs. (1), (2), and (3), we find

$$R_i = \frac{Z_0^2}{Z_0} \quad \text{and} \quad X_i = -\frac{Z_0^2}{Z_0}$$

both of which are reciprocal of Eq. (11-106).

(ii) From Eq. (11-106): $\cos \beta l = \frac{Z_0 \cos \beta l + jZ_L \sin \beta l}{Z_0}$

At the input, $Z = Z_0$, $\beta l = \pi/2$, we have

$$Z_0 = Z_0 \cos \beta l = Z_0 \cos \pi/2 = 0$$

At the load, $Z = Z_L$, $\beta l = 0$, and $Z_L = Z_0 \cos \beta l = Z_0 \cos 0 = Z_0$

$$\therefore \frac{Z_0}{Z_0} = \frac{Z_0}{Z_0} = \frac{Z_0 \cos \beta l}{Z_0}$$

Ex. 12 (i) $Z = \frac{Z_0}{j} = \left| \frac{Z_0}{j} \right| = \frac{Z_0 \cos \beta l + jZ_L \sin \beta l}{j \cos \beta l + \sin \beta l}$

where $Z_L = Z_0 \cos \beta l$ and $Z_0 \cos \beta l$

$$\rightarrow Z = \frac{Z_0 \cos \beta l + jZ_0 \cos \beta l \sin \beta l}{j \cos \beta l + \sin \beta l}$$

When $\beta l = \pi/2$, $Z = \frac{Z_0 \cos \pi/2 + jZ_0 \cos \pi/2 \sin \pi/2}{j \cos \pi/2 + \sin \pi/2} = \frac{0 + jZ_0 \cdot 0 \cdot 1}{j \cdot 0 + 1} = 0$

(ii) $\beta l = \pi$ and $Z = Z_0 \cos \pi = -Z_0 \rightarrow Z = -Z_0 \cos \pi$

$$Z = -Z_0 \cos \pi = -Z_0 \cos \pi$$

Q) From Eq. (b) above, $v = \beta c = \frac{c\sqrt{1-\beta^2}}{\beta}$

where $\beta = v/c = \beta_0$, and $L = L_0 \sqrt{1-\beta^2}$

$$\implies \beta_0 = \frac{(20 \times 10^8) \sqrt{1-\beta_0^2}}{1.5 \times 10^8}$$

$\implies \beta = \frac{1}{3}$ or $\frac{1}{3}$ for $v = c$ and $L_0 = 20$.

Also, $L_0 = \frac{L_0 \sqrt{1-\beta_0^2}}{1-\beta_0^2} \implies L = \frac{L_0}{\beta_0} \left[1 - \frac{1-\beta_0^2}{1-\beta_0^2} \right]$

$L_0 = 2$ yields negative + (discarded).

For $L_0 = \frac{1}{3}$, $L = \left[\frac{1/3}{1/3} \right] \implies L_0 = 2/3 \text{ (OK)}$
 $L = \frac{1}{3} \implies L_0 = 2/3 \text{ (OK)}$

Use $L_0 = 2/3$ to obtain L_0 correct to the last
 of significant figures.

Ex. 21 a) $|r|^2 = \left| \frac{(x_2 - x_1) + j(z_2 - z_1)}{(x_2 + x_1) + j(z_2 + z_1)} \right|^2 = \frac{(x_2 - x_1)^2 + z_2^2}{(x_2 + x_1)^2 + z_2^2}$

$$\frac{z_2^2}{z_1^2} = 3 \implies z_2 = \sqrt{3} z_1$$

If $z_1 = 40 \angle 130^\circ$, $z_2 = 70 \angle 130^\circ$

b) $\text{Min. } |r| = \sqrt{\frac{z_2^2 - z_1^2}{z_1^2 + z_2^2}} = \sqrt{\frac{3z_1^2 - z_1^2}{z_1^2 + 3z_1^2}} = \frac{1}{2}$

$\text{Min. } \beta = \frac{1/2}{1/2} = 1$

c) From Eq. (b) above, $v = \beta c = \frac{c\sqrt{1-\beta^2}}{\beta} = 2.0 \times 10^8$

$$\implies \beta = \frac{1}{\sqrt{3}} \left[(1-\beta^2) \sqrt{1-\beta^2} \right] \text{ (Discarding } \beta = 1 \text{)}$$

At voltage minimum, $L = \frac{1}{2} = \frac{1}{2}$

$L = L$ (the negative sign)

Use $L_0 = \frac{L_0 \sqrt{1-\beta^2}}{1-\beta^2} = L \implies L_0 = \frac{1}{2}$

\therefore Voltage minimum occurs to the load at $\left(\frac{1}{2} - \frac{1}{2}\right)$
 or $0 \lambda/8$ from the load.

Ex 11.12 (i) From Eqs. (9-11a) and (9-11b):

$$v(t) = \frac{1}{2}(V_1 + V_2)e^{j\omega t} [1 + \cos(2\omega t + \phi)]$$

$$\text{where } V = \frac{\sqrt{2}V_m}{2} = V_m e^{j\phi}, \quad \phi = \alpha_2 - \alpha_1$$

$$\text{Max } |v(t)| = \frac{1}{2}(V_1 + V_2)e^{j\omega t} [1 + \cos(2\omega t)] \text{ for } \phi = 0$$

$$\text{min } |v(t)| = \frac{1}{2}(V_1 + V_2)e^{j\omega t} [1 - \cos(2\omega t)] \text{ for } \phi = \pi$$

$$\text{So } V = \frac{\text{Max } |v(t)|}{\text{min } |v(t)|} = \frac{1 + \cos(2\omega t)}{1 - \cos(2\omega t)} \quad \left\{ \begin{array}{l} \text{Upper envelope is } V \\ \text{Lower envelope is } 1/V \end{array} \right.$$

(ii) From Eqs. (9-11a): $V_1 \cos \omega t = \frac{1}{2}(V_1 + V_2) \cos \omega t + \frac{1}{2}(V_1 - V_2) \cos 3\omega t$

$$\text{At a voltage max, } \phi = 0, \quad V_1 \cos \omega t = \frac{1}{2}(V_1 + V_2)$$

$$\text{(iii) At a voltage min, } \phi = \pi, \quad V_1 \cos \omega t = \frac{1}{2}(V_1 - V_2)$$

Ex 11.13 From Eqs. (9-11a): $V_1 = V \left(\frac{1 + \cos 2\omega t}{2} \right) \rightarrow V_1 \cos \left(\frac{1 + \cos 2\omega t}{2} \right) = \cos \omega t$

$$V_2 = V \left(\frac{1 - \cos 2\omega t}{2} \right) \rightarrow V_2 = V \left(\frac{1 - \cos 2\omega t}{2} \right)$$

Now $V_1 = 2 \cos \omega t$ and $V_2 = 2 \cos 3\omega t$, we have

$$2 \cos 3\omega t = V \left(\frac{1 - \cos 2\omega t}{2} \right) \rightarrow \begin{cases} 4 \cos^2 \omega t \cos \omega t = 1 - \cos^2 \omega t \\ 2 \cos^2 \omega t \cos \omega t = 1 - \cos^2 \omega t \end{cases}$$

$$\therefore V_1 = 2 \cos \omega t, \quad \phi = \cos 3\omega t = \cos \omega t \rightarrow V = 2 \cos \omega t$$

Ex 11.14 (a) $|v| = \frac{1}{2}|v| = \frac{1}{2}|v| = \frac{1}{2}$

$$\text{Eq. (9-11a): } v(t) = \frac{1}{2}(V_1 + V_2)e^{j\omega t} [1 + \cos(2\omega t)]$$

$$\text{Eq. (9-11b): } V = \frac{\sqrt{2}V_m}{2} = V_m e^{j\phi}, \quad \phi = \alpha_2 - \alpha_1$$

$$\text{Perhaps it is a minimum when } \phi = \pi \rightarrow V_1 = V_2 \frac{1 + \cos 2\omega t}{2} = 1$$

$$\therefore V = \frac{1}{2} e^{j\omega t}$$

$$\text{(d) } V_1 = V_2 \left(\frac{1 + \cos 2\omega t}{2} \right) = 4 \cos \omega t \cos 3\omega t \text{ (Eq. 9-11a)}$$

$$\text{(e) Terminating resistance } R_1 = \frac{1}{2} = \frac{1}{2} \Omega = 100 \text{ m}\Omega$$

$$R_2 = \frac{1}{2} \cos^2 \omega t = \cos^2 \omega t = 1 \Omega$$

$$\text{Another set of solutions: } R_1 = 1 \Omega, \text{ source } \& R_2 = 1 \Omega$$

Ex 2.21 $\lim_{x \rightarrow \infty} (x^2 - 2x) = \infty = \frac{\infty}{1}$ (L'Hôpital's Rule).

Let $a = \frac{1}{2}$, $b = \frac{1}{2}$, $c = \frac{1}{2}$, and substitute

$$x^2 - 2x = \frac{x^2 - 2x + \frac{1}{4}}{\frac{1}{x^2}} \rightarrow \begin{cases} \text{if } x \rightarrow \infty, \\ \text{if } x \rightarrow \infty, \frac{1}{x^2} = 0. \end{cases}$$

We have

$$a = \frac{1}{2} \left[(x^2 - 2x + \frac{1}{4}) \sqrt{x^2 - 2x + \frac{1}{4}} \right]$$

$$b = \frac{1}{2} \left[-(x^2 - 2x + \frac{1}{4}) \sqrt{x^2 - 2x + \frac{1}{4}} \right],$$

$$c = \frac{1}{2} (2x)^2.$$

Ex 2.22 $a_1 = a_2 \frac{1}{1+x^2}$

$$f = (1+x^2)^{-1}, \quad f' = \frac{-2x}{(1+x^2)^2}, \quad a_1 = \frac{1}{2} a_2 \text{ (a.e.)}$$

$$\therefore a_1 = \frac{1}{2} \frac{(-2x)(1+x^2)^{-2}}{(1+x^2)^{-2}}$$

$$= \frac{1}{2} \frac{(-2x)(1+x^2)^{-2} \cdot (1+x^2)^2}{(1+x^2)^{-2} \cdot (1+x^2)^2}$$

$$= \frac{1}{2} \frac{(-2x)(1+x^2)^0}{(1+x^2)^0}$$

Ex 2.23 (i) Given: $a_1 = 2x^2 \cos x$, $a_2 = 2x \sin x$, $a_3 = 2x^2 \cos x$

$$a_1 = \frac{2x^2}{2x^2} a_2, \quad a_2 = \frac{2x^2}{2x^2} a_3$$

$$\text{where } a_1 = \frac{1}{2} \frac{(2x^2)(2x \cos x)}{(2x^2)(2x \cos x)} = \frac{1}{2} \frac{(2x^2)(2x \cos x)}{(2x^2)(2x \cos x)}$$

$$\therefore a_1 = \frac{1}{2} \frac{(2x^2)(2x \cos x)}{(2x^2)(2x \cos x)} = \frac{1}{2} \left(\frac{2x^2}{2x^2} \frac{2x \cos x}{2x \cos x} \right) \text{ (a.e.)}$$

$$a_2 = \frac{1}{2} \frac{(2x^2)(2x \cos x)}{(2x^2)(2x \cos x)} = \frac{1}{2} \left(\frac{2x^2}{2x^2} \frac{2x \cos x}{2x \cos x} \right) \text{ (a.e.)}$$

Setting $a_1 = a_2$ and $a_2 = a_3$ (a.e.) and hence

$$\text{we have } a_1 = a_2 = a_3 = \frac{1}{2} \frac{2x^2}{2x^2} \frac{2x \cos x}{2x \cos x} \left(\frac{2x^2}{2x^2} \frac{2x \cos x}{2x \cos x} \right)$$

$$= \frac{1}{2} (1) (1) \text{ (a.e.)}$$

$$a_1 = a_2 = a_3 = \frac{1}{2} \frac{2x^2}{2x^2} \frac{2x \cos x}{2x \cos x} = \frac{1}{2} (1) (1) \text{ (a.e.)}$$

$$\text{d) } \beta = \frac{1+2\beta^2}{1-\beta^2} = 2.$$

$$\begin{aligned} \text{e) } (R_{2n})_1 &= \int_0^1 R_n(x) dx = \int_0^1 \left(\frac{1}{2} \left(\frac{1}{2} + x \right)^n + \frac{1}{2} \left(\frac{1}{2} - x \right)^n \right) dx \\ &= \frac{1}{2} \int_0^1 \left(\frac{1}{2} + x \right)^n dx + \frac{1}{2} \int_0^1 \left(\frac{1}{2} - x \right)^n dx \\ &= \frac{1}{2} \left[\frac{1}{n+1} \left(\frac{1}{2} + x \right)^{n+1} \right]_0^1 + \frac{1}{2} \left[-\frac{1}{n+1} \left(\frac{1}{2} - x \right)^{n+1} \right]_0^1 \\ &= \frac{1}{2(n+1)} \left(\left(\frac{3}{2} \right)^{n+1} - \left(\frac{1}{2} \right)^{n+1} \right) - \frac{1}{2(n+1)} \left(\left(\frac{1}{2} \right)^{n+1} - \left(\frac{3}{2} \right)^{n+1} \right) \\ &= \frac{1}{2(n+1)} \left(\left(\frac{3}{2} \right)^{n+1} - \left(\frac{1}{2} \right)^{n+1} + \left(\frac{3}{2} \right)^{n+1} - \left(\frac{1}{2} \right)^{n+1} \right) \\ &= \frac{1}{n+1} \left(\left(\frac{3}{2} \right)^{n+1} - \left(\frac{1}{2} \right)^{n+1} \right) \end{aligned}$$

$$\begin{aligned} \text{f) } R_{2n}(0) &= \frac{1}{2} \left(\left(\frac{1}{2} \right)^n + \left(\frac{1}{2} \right)^n \right) = \frac{1}{2} \left(2 \left(\frac{1}{2} \right)^n \right) = \left(\frac{1}{2} \right)^n \\ R_{2n}(1) &= \frac{1}{2} \left(\left(\frac{3}{2} \right)^n + \left(\frac{1}{2} \right)^n \right) = \frac{1}{2} \left(\left(\frac{3}{2} \right)^n + \left(\frac{1}{2} \right)^n \right) \end{aligned}$$

$$\text{g) } R_{2n} = \int_0^1 R_n(x) dx = \int_0^1 \left(\frac{1}{2} \left(\frac{1}{2} + x \right)^n + \frac{1}{2} \left(\frac{1}{2} - x \right)^n \right) dx$$

$$\begin{aligned} \text{h) } R_n &= \int_0^1 R_{2n}(x) dx = \int_0^1 \left(\frac{1}{2} \left(\frac{1}{2} + x \right)^{2n} + \frac{1}{2} \left(\frac{1}{2} - x \right)^{2n} \right) dx \\ &= \frac{1}{2} \int_0^1 \left(\frac{1}{2} + x \right)^{2n} dx + \frac{1}{2} \int_0^1 \left(\frac{1}{2} - x \right)^{2n} dx \\ &= \frac{1}{2} \left[\frac{1}{2n+1} \left(\frac{1}{2} + x \right)^{2n+1} \right]_0^1 + \frac{1}{2} \left[-\frac{1}{2n+1} \left(\frac{1}{2} - x \right)^{2n+1} \right]_0^1 \\ &= \frac{1}{2(2n+1)} \left(\left(\frac{3}{2} \right)^{2n+1} - \left(\frac{1}{2} \right)^{2n+1} \right) - \frac{1}{2(2n+1)} \left(\left(\frac{1}{2} \right)^{2n+1} - \left(\frac{3}{2} \right)^{2n+1} \right) \\ &= \frac{1}{2n+1} \left(\left(\frac{3}{2} \right)^{2n+1} - \left(\frac{1}{2} \right)^{2n+1} \right) \end{aligned}$$

$$\text{i) } \frac{R_n}{R_{2n}} = \frac{1 - \left(\frac{1}{2} \right)^{2n} = 1 - \left(\frac{1}{4} \right)^n}{\frac{1}{2n+1} \left(\left(\frac{3}{2} \right)^{2n+1} - \left(\frac{1}{2} \right)^{2n+1} \right)}$$

$$\begin{aligned} \text{j) } R_{2n} &= \frac{1}{2} \left(\left(\frac{3}{2} \right)^n + \left(\frac{1}{2} \right)^n \right) = \frac{1}{2} \left(\left(\frac{3}{2} \right)^n + \left(\frac{1}{2} \right)^n \right) \\ R_n &= \frac{1}{2} \left(\left(\frac{3}{2} \right)^n + \left(\frac{1}{2} \right)^n \right) = \frac{1}{2} \left(\left(\frac{3}{2} \right)^n + \left(\frac{1}{2} \right)^n \right) \\ R_n &= \frac{1}{2} \left(\left(\frac{3}{2} \right)^n + \left(\frac{1}{2} \right)^n \right) = \frac{1}{2} \left(\left(\frac{3}{2} \right)^n + \left(\frac{1}{2} \right)^n \right) \end{aligned}$$

$$\begin{aligned} \text{k) } R_n &= \frac{1}{2} \left(\left(\frac{3}{2} \right)^n + \left(\frac{1}{2} \right)^n \right) = \frac{1}{2} \left(\left(\frac{3}{2} \right)^n + \left(\frac{1}{2} \right)^n \right) \\ R_{2n} &= \frac{1}{2} \left(\left(\frac{3}{2} \right)^{2n} + \left(\frac{1}{2} \right)^{2n} \right) = \frac{1}{2} \left(\left(\frac{3}{2} \right)^{2n} + \left(\frac{1}{2} \right)^{2n} \right) \end{aligned}$$

$$\begin{aligned} \text{l) } R_{2n} &= \frac{1}{2} \left(\left(\frac{3}{2} \right)^{2n} + \left(\frac{1}{2} \right)^{2n} \right) = \frac{1}{2} \left(\left(\frac{3}{2} \right)^{2n} + \left(\frac{1}{2} \right)^{2n} \right) \\ R_n &= \frac{1}{2} \left(\left(\frac{3}{2} \right)^n + \left(\frac{1}{2} \right)^n \right) = \frac{1}{2} \left(\left(\frac{3}{2} \right)^n + \left(\frac{1}{2} \right)^n \right) \end{aligned}$$

$$\begin{aligned} \text{m) } R_{2n} &= \frac{1}{2} \left(\left(\frac{3}{2} \right)^{2n} + \left(\frac{1}{2} \right)^{2n} \right) = \frac{1}{2} \left(\left(\frac{3}{2} \right)^{2n} + \left(\frac{1}{2} \right)^{2n} \right) \\ R_n &= \frac{1}{2} \left(\left(\frac{3}{2} \right)^n + \left(\frac{1}{2} \right)^n \right) = \frac{1}{2} \left(\left(\frac{3}{2} \right)^n + \left(\frac{1}{2} \right)^n \right) \end{aligned}$$

n) At the limit, $n \rightarrow \infty$.

$$\begin{aligned} R_{2n} &= \frac{1}{2} \left(\left(\frac{3}{2} \right)^{2n} + \left(\frac{1}{2} \right)^{2n} \right) \\ &= \frac{1}{2} \left(\left(\frac{3}{2} \right)^{2n} + \left(\frac{1}{2} \right)^{2n} \right) = \frac{1}{2} \left(\left(\frac{3}{2} \right)^{2n} + \left(\frac{1}{2} \right)^{2n} \right) \end{aligned}$$

$$R_n = \frac{1}{2} \left(\left(\frac{3}{2} \right)^n + \left(\frac{1}{2} \right)^n \right) = \frac{1}{2} \left(\left(\frac{3}{2} \right)^n + \left(\frac{1}{2} \right)^n \right)$$

$$(R_{2n})_1 = \int_0^1 R_n(x) dx = \frac{1}{2(n+1)} \left(\left(\frac{3}{2} \right)^{n+1} - \left(\frac{1}{2} \right)^{n+1} \right)$$

2.2-22 $E_1 = 0$, $E_2 = 1$

(a) $0 < t < T/2$



(b) $T/2 < t < T$



(c) $T < t < 3T/2$



(d) $3T/2 < t < 2T$



$$V_1^+ = V_1, \quad V_1^- = V_1^+ = V_1$$

$$V_2^+ = V_2, \quad V_2^- = V_2^+ = V_2$$

$$V_3^+ = V_2, \quad V_3^- = V_2$$

$$V_1^+ = V_1, \quad V_1^- = -V_1^+ = -V_1$$

$$V_2^+ = V_1^-, \quad V_2^- = -V_1^- = V_1$$

$$V_3^+ = V_2^-, \quad V_3^- = V_2^- = V_2$$

At $t=2T$, both V_1 and V_2 revert back to the initial value of 0, and the cycle repeats itself with a period $2T$.



At the connecting points of two transmission lines with different characteristic impedances Z_0 and Z_0'

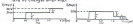
$$\begin{array}{l} \frac{V_1}{Z_0} = \frac{V_2}{Z_0} + \frac{V_3}{Z_0} \\ Z_0 = Z_0' + \frac{Z_0 Z_0'}{Z_0'} \end{array} \quad \begin{array}{l} V_1 = V_2 + V_3 \\ Z_0 = Z_0' \left(1 + \frac{Z_0}{Z_0'} \right) = Z_0' \frac{Z_0 + Z_0'}{Z_0'} \end{array}$$

$$\begin{array}{l} \text{Solving } V_2 = \frac{Z_0 Z_0'}{Z_0 + Z_0'} V_1, \quad V_3 = \frac{Z_0}{Z_0 + Z_0'} V_1 \\ Z_0' = \frac{Z_0 Z_0'}{Z_0 + Z_0'} Z_0', \quad Z_0' = \frac{Z_0 Z_0'}{Z_0 + Z_0'} Z_0' \end{array}$$

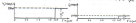
$$\begin{array}{l} a) \quad V_2 = \frac{50 \cdot 100}{50 + 100} V_1 = 33.33 V, \quad V_3 = \frac{50}{50 + 100} V_1 = 16.67 V \\ V_2 = \frac{100 \cdot 50}{100 + 50} V_1 = 33.33 V, \quad Z_0' = \frac{100 \cdot 50}{100 + 50} Z_0' = 33.33 \Omega \\ V_2 = \frac{100 \cdot 50}{100 + 50} V_1 = 33.33 V, \quad Z_0' = \frac{100 \cdot 50}{100 + 50} Z_0' = 33.33 \Omega \end{array}$$

At transient wave on the parallel cable after V_2 and V_3 reach the input terminated at $Z_0 = 50 \Omega$, $V_2 = 33.33 V$ and $V_3 = 16.67 V$, and no transient waves on the series line after V_2 and V_3 reach the load Z_0' at $t_2 = 1.5 \mu s$, $V_2 = 33.33 V$ and $V_3 = 16.67 V$, $V_2 = 33.33 V$ and $V_3 = 16.67 V$.

b) On the parallel cable of length $L/2 = 0.1$ (just for V_2 and V_3 to reach the output $Z_0 = 100 \Omega$). The reflected waves V_2 and V_3 arrive at the output at $t = 1.5 \mu s$. There are no changes after that.



On the series line, steady state is reached at $t = 1.5 \mu s$.



c)



Ex. 2.12 The current reflection diagram for Example 9-17 is



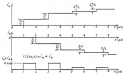
$$\Gamma_L = \frac{1}{2} = \Gamma_H = 1$$

$$\Gamma = 2 \text{ (vol.)}$$

Indices on the directed lines are normalized with respect to

$$\Gamma^2 = \frac{I^2}{I_0^2} = \frac{16}{16}$$

$$= 1 \text{ (vol.)}$$



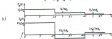
Ex 2.11 Use the equivalent circuit in Fig. 2.11(b) to study transient voltage and currents:



(a) Amplitude of first current wave (starting from $t=0$ to $t=0.1$) $I_1^* = \frac{V_0}{\sqrt{R_0^2 + \omega_0^2 L_0^2}} = \frac{100}{\sqrt{100}}$

Refer to Fig. 2.11(a) $L_0 = -L_1$, $V_0 = 100 \sin \omega_0 t$

$L = 100 \mu\text{H} \rightarrow C_2 = \frac{100 \times 10^{-6}}{100} = 1 \mu\text{F}$, $C_1 = 1 \mu\text{F}$, $R = 100 \Omega$



$R = 100 \rightarrow C_2 = \frac{100 \times 10^{-6}}{100} = 1 \mu\text{F}$, $C_1 = 1 \mu\text{F}$, $R = 100 \Omega$

Ex 2.12 (a) Governing equation of the load for $t > T$

$$L_0 \frac{di(t)}{dt} + (R_0 + R_1)i(t) = 0 \quad t > T$$

$$\text{Solution: } i(t) = \frac{V_0}{R_0 + R_1} \left[1 - e^{-\frac{R_0 + R_1}{L_0}(t-T)} \right], \quad t > T$$

For the present problem, $V_0 = 100 \text{ V}$, $R_0 = 100 \Omega$, $R_1 = 100 \Omega$

$$L_0 = 100 \mu\text{H} = 100 \times 10^{-6} \text{ H}, \quad T = 0.1 \text{ s} = 100 \times 10^{-3} \text{ s} = 100 \times 10^{-6} \text{ s}$$

$$i(t) = \frac{100}{200} \left[1 - e^{-\frac{200}{100 \times 10^{-6}}(t-100 \times 10^{-6})} \right], \quad t > 100 \times 10^{-6} \text{ s}$$

$$i(t) = 0.5 \left[1 - e^{-2000(t-100 \times 10^{-6})} \right], \quad t > 100 \times 10^{-6} \text{ s}$$



$$\text{At } t = 0.1 \text{ s},$$

$$i(0.1) = \frac{100}{200} \left[1 - e^{-2000(0.1-0.1)} \right]$$

$$= 0.5 \text{ A}$$



At $x = \pi/2$ and $3\pi/2$

$$f(\pi/2) = f(3\pi/2) = 1 - e^{-2(\pi/2)} = 1 - e^{-\pi}$$

Ex. 10.11 From Eq. (10-100) $\psi(x) = 2Ae^{-\alpha x} - A_0 \delta(x)$ (1)

At the lead $\psi(0) = \frac{2A}{\alpha} + A_0 = \frac{2A}{\alpha}$ (2)

Substituting (2) in (1) $\psi(x) = 2Ae^{-\alpha x} - \left(\frac{2A}{\alpha} - \frac{2A}{\alpha}\right)\delta(x) = 2Ae^{-\alpha x}$ (3)

(a) Solution of (3): $\psi(x) = 2Ae^{-\alpha x} \left[1 - e^{-\alpha x} \delta(x)\right]$ (4)

For this problem $\alpha = \frac{1}{2} = 0.5 \text{ rad}^{-1}$ (5)

$T = 2\pi \text{ sec}$, $A_0 = 0.5 \text{ sec}^{-1}$, $A_0 A = 0.111 \text{ sec}^{-2}$

$\psi(x) = 0.222 \left[1 - e^{-0.5x} \delta(x)\right]$ (6)

From (3): $\psi(0) = 0.444 = 2 \text{ sec}^{-2} \delta(x)$ (7)

(b) At $x = \pi/2$, $\psi(x) = 0.222 e^{-0.5(\pi/2)}$ (8)



Ex. 10.12 (a) $E_1 = E_2 = \frac{1}{2} E_0 = \frac{1}{2} |E_0| e^{i\omega t}$ (1)

Between $x = 0, \ell$

$E_0 = \sqrt{\frac{2\mu_0 \omega^2}{\epsilon_0} P} = |E_0| e^{i\omega t}$ (2)

Between $x = \ell, 2\ell$

$\therefore E_2 = \frac{1}{2} E_0 = \frac{1}{2} |E_0| e^{i\omega t}$ (3)

and q must be determined

(b) $\Gamma = \left| \frac{E_2 - E_1}{E_2 + E_1} \right|^2 = \frac{(\frac{1}{2} |E_0| e^{i\omega t} - |E_0| e^{i\omega t})^2}{(\frac{1}{2} |E_0| e^{i\omega t} + |E_0| e^{i\omega t})^2} = \frac{(\frac{1}{2} |E_0| e^{i\omega t})^2}{(\frac{3}{2} |E_0| e^{i\omega t})^2} = \frac{1}{9}$ (4)

between $x = \ell, 2\ell$

(c) $\frac{dE}{dt} = 0$ (5)

At $x = \ell$, $\psi = 0.5 \text{ sec}^{-1}$

At $x = 2\ell$, $\psi = 0.111$

$$P(2) = \frac{1}{2} \cos^2(180^\circ) = \frac{1}{2} \cos^2(0^\circ)$$

a) Quarter-circle line, $d = 100$ in, $d/4 = 25$ in.

Start at the extreme-left point P_1 on the extreme circle, rotate clockwise one complete revolution (360° or 2π rad) and continue on P_2 an additional 180° or π rad on the "wavy-line" toward generator "back". Read $d = 100$ in $\rightarrow P_2 = 75 + (1/4)(100) = 75 + 25 = 100$ in.

Draw a straight line from the (0, 100) point through the center and intercept at (100, 100) on the opposite side of the chart $\rightarrow Q = \frac{1}{2} \pi + (1/2)(\pi) = \pi = 180^\circ$ (2).

b) Semi-circular line, $d = 100$ in, $d/2 = 50$ in.

Start from the extreme-left point P_1 , rotate clockwise one complete revolution and continue on the arc another 180° to read $d = 100$ in $\rightarrow P_2 = 75 + (1/2)(100) = 75 + 50 = 125$ in.

Draw a straight line from the (0, 125) point through the center and intercept at (100, 125) on the opposite side of the chart $\rightarrow Q = \frac{1}{2} \pi + (1/2)(\pi) = \pi = 180^\circ$ (2).

Example



$$Q = \frac{1}{2} \pi + (1/2)(\pi) = \pi = 180^\circ$$

a) 1. Locate $d/4 = 25$ in on

1. Read scale of P above a 100-inch stroke through P_1 , intercepting P_2 at 125 in $\rightarrow d = 125$.

$$P = \frac{1}{4} \cos^2(180^\circ) = \frac{1}{4} \cos^2(0^\circ)$$

a) 1. Draw line OP_1 , intercepting the periphery of ϕ' .
Read 100 in "wavy-line" toward generator "back".

2. Move clockwise by 180° to start (Point Q').

3. Draw OQ' and Q' , intercepting the (0) circle of ϕ' .

4. Read $d = 125$ in on P .

$$Q = 180^\circ = \pi = 180^\circ$$

4) External line $Q_1'Q_2'$ to Q_2 . Read $Q_1' = 0.00$ (Point Q_1').

$$Q_1' = \frac{1}{2} Q_2 = 0.000 - 0.000 \text{ (2)}$$

Q There is no voltage minimum on this line, but $Q_1 = Q_2$.

Example



$$Q_2 = \frac{1}{2} (1 + j1) = 0.5 + j0.5$$

4) Draw $Q_2 = 0.5 + j0.5$ on Smith chart (Point Q_2). Move counter-clockwise through Γ circles through Q_2 , intersecting line R_{NL} at 1.00 . ——— $d = 0.25\lambda$.

$$d) P = 0.25\lambda^{0.25}$$

4) Draw line Q_1' , intersecting the periphery at Q_1' . Read 0.00 as "normalized forward power" scale.

5. Move clockwise by 0.25λ to 0.50λ (Point Q_1').

6. Draw Q_1' and Q_2' , intersecting the Γ circle at Q_1 .

Read $Q_1 = 0.01 - j0.01$ at Q_1 .

$$Q_1 = 0.01 - j0.01 \text{ at } P = 0.25$$

4) External line $Q_1'Q_2'$ to Q_2 . Read $Q_1' = 0.00$ (Point Q_1').

$$Q_1' = \frac{1}{2} Q_2 = 0.000 - 0.000 \text{ (2)}$$

Q There is a voltage minimum at $Q_1 = 0.000$.

Example $\lambda/4 = 0.25$, $\lambda = 0.25 \text{ (100)}$

First voltage minimum occurs at $Q_1 = \frac{1}{2} Q_2 = 0.10$.



4) Start from Q_2 and rotate counter-clockwise 0.25λ toward the center to Q_1' .

5. Draw the Γ circle, intersecting line Q_1' at d (Point).

6. Draw Q_1' , intersecting the Γ circle at Q_1 .

4. Find $\eta_1 = \cos \theta - j \sin \theta$.

$$\eta_1 = \cos \theta - j \sin \theta = e^{-j\theta} \quad \text{or}$$

5) $F = \frac{1}{s^2} \Rightarrow F^{-1} = t e^{j\omega t}$

6) If $\eta_2 = 0$, the first voltage minimum would be at $\eta_2 = \lambda/4 = 25 \text{ cm}$ from the short-circuit.

Example



a) $\eta_1 = \frac{1}{\sqrt{2}} (\cos \theta - j \sin \theta)$
 $= \cos \theta - j \sin \theta$

1. Define η_1 as field along z-axis (Point P).

2. Take θ and r , and extend to K .

3. Find an isosceles triangle formed parallel to axis z .

$$r \theta = \pi \sin \theta \quad \text{and} \quad r = \frac{\pi}{\theta} \sin \theta \quad \text{or} \quad \theta = \frac{\pi}{r} \sin \theta$$

$$\frac{\pi}{r} = \theta \sin \theta \quad \text{or} \quad \theta = \frac{\pi}{r} \sin \theta \quad \text{or} \quad \theta = \frac{\pi}{r} \sin \theta \quad \text{or} \quad \theta = \frac{\pi}{r} \sin \theta$$



b) 1. Define η_2 as field along z-axis (Point Q).

2. Draw line from θ through P to Q . Find an isosceles triangle parallel to axis z .

3. Define η_2 as field along z-axis (Point P).

4. Take η_2 as field along z-axis (Point Q).

5. Mark point P on the z -axis such that $\frac{\pi}{r} = \theta \sin \theta$.

6. Find η_1 at P : $\eta_1 = \cos \theta - j \sin \theta \quad \text{or} \quad \eta_1 = \cos \theta - j \sin \theta$

Q) 1. Show clockwise from \mathbb{R}^2 an "orthogonal" vector generator" leads to $\alpha \mathbb{R}^2$, say \mathcal{P} .

2. Give \mathcal{P} .

3. Show point β on line \mathcal{P} ' such that

$$\vec{\beta\mathcal{P}} = e^{i\pi/4} \vec{\beta\mathcal{O}} = \alpha \mathcal{P} \vec{\mathcal{O}}$$

4. Read off \mathcal{P} : $\mathcal{P} = \alpha \cos(\pi/4) \vec{i} + \alpha \sin(\pi/4) \vec{j}$.

Ex: $\mathcal{P} = \beta = 2 + 2i \text{ rad}$, $\alpha = 1.2 \text{ rad} \implies \beta = \frac{\alpha}{\beta} \alpha \cos(\pi/4)$

$$\mathbb{R}^2 = 73 + 100 = 100 \text{ rad}$$

For $\beta = \alpha \cos(\pi/4)$ line: $\mathbb{R}^2 = \alpha \cos(\pi/4) (\frac{\beta}{\alpha})$

$$\beta = 1.2 \text{ rad} \implies \alpha = 100 \text{ rad}$$

Ex: \mathcal{P}



$$\mathbb{R}^2 = \alpha \cos(\pi/4)$$

$$\mathbb{R}^2 = 1.2$$

Q) For $\mathcal{P} = \alpha \cos(\pi/4)$

$$\mathbb{R}^2 = \alpha \cos(\pi/4)$$

$$\mathbb{R}^2 = \alpha + 1.2 \cos(\pi/4) = \alpha \cos(\pi/4)$$

$$\mathbb{R}^2 = \alpha + 1.2$$

$$\implies \alpha = \frac{1.2}{\cos(\pi/4)}$$

$$\mathbb{R}^2 = \frac{1.2}{\cos(\pi/4)} + 1.2$$

$$\mathbb{R}^2 = \frac{1.2}{\cos(\pi/4)} + 1.2$$

Q) For $\mathbb{R}^2 = \alpha \cos(\pi/4)$, $\mathbb{R}^2 = \alpha \cos(\pi/4)$

The required values of \mathbb{R}^2 are $\mathbb{R}^2 = \alpha \cos(\pi/4)$

	$\mathbb{R}^2 = \alpha \cos(\pi/4)$	$\mathbb{R}^2 = \alpha \cos(\pi/4)$
$\mathbb{R}^2 = \alpha \cos(\pi/4)$	$\alpha = 1.2$, $\mathbb{R}^2 = 1.2 \cos(\pi/4)$	$\alpha = 1.2$, $\mathbb{R}^2 = 1.2 \cos(\pi/4)$
$\mathbb{R}^2 = 1.2$	$\alpha = 1.2 \cos(\pi/4)$, $\mathbb{R}^2 = 1.2$	$\alpha = 1.2 \cos(\pi/4)$, $\mathbb{R}^2 = 1.2$

Ex. 10.11 $\alpha = 0, \beta = j\omega$

Use Smith chart as an impedance chart. Some restrictions as that in problem 10-10 except R_{in} would be in the positive half (bounded by a vertical part) of the impedance chart.

$$\hat{z}_1 = \hat{z}_2 = 1 + j\omega L, \quad \hat{z}_2 = \hat{z}_3 = j\omega L \quad \text{with } \hat{z}_2 = \hat{z}_1 \hat{z}_3 = \text{const.} = 1 + j\omega L.$$

$$\hat{z}_2 = \hat{z}_3 = 1 - j\omega L \quad \text{with } \hat{z}_2 = \hat{z}_3 = \text{const.} = 1 - j\omega L.$$

To obtain a match with a series stub having $\hat{z}_2 = \frac{1 + j\omega L}{1 - j\omega L}$, we need a normalized stub impedance of $j\omega L = j\omega L$ for solution corresponding to \hat{z}_1 . From Smith chart we obtain the required stub length $L_2 = 0.25\lambda$.

Similarly, for solution corresponding to \hat{z}_3 , a stub with a normalized impedance of $-j\omega L$ is needed, which requires a stub length $L_2 = 0.25\lambda$.

Ex. 10.12



$$\hat{z}_1 = 1 + j1 = j\omega L.$$

$$\hat{z}_2 = \hat{z}_1 = 1 + j1 = \text{const.} = j\omega L.$$

$$\hat{z}_2 = \hat{z}_3 = 1 + j1 = \text{const.} = j\omega L.$$

$$\hat{z}_2 = \hat{z}_3 = 1 + j1 = \text{const.} = j\omega L.$$

$$\hat{z}_2 = \hat{z}_3 = 1 + j1 = \text{const.} = j\omega L.$$

$$\hat{z}_2 = \hat{z}_3 = 1 + j1 = \text{const.} = j\omega L.$$

of Normalized stub of Quarter-wavelength stub

$\Gamma_{\text{max}} = \hat{z}_1 = 1 + j1$	$L_2 = 0.25\lambda$	$L_2 = 0.25\lambda$
$\Gamma_{\text{min}} = \hat{z}_2 = 1 - j1$	$L_2 = 0.25\lambda$	$L_2 = 0.25\lambda$
$\Gamma_{\text{max}} = j1$	$L_2 = 0.25\lambda$	$L_2 = 0.25\lambda$
$\Gamma_{\text{min}} = -j1$	$L_2 = 0.25\lambda$	$L_2 = 0.25\lambda$

Ex-20



$$x_1 = \frac{Rr}{\sin \alpha} = R \sin \alpha$$

Point Q_1 on circle about
(Center at O')

Since the rotated gear
circle is tangent to the
gear circle, an added
line length d_1 is needed
to connect Q_1 (radius r),

moving from Q_1 along the (r)-circle to Q_2 (radius R)
on the gear circle (Center at O). Note that Q_2
is different from Q_1 , the point of tangency between
the gear and rotated gear circles.

$$d_1 = R_1 \alpha = R \sin \alpha = R \sin \alpha$$

$$d_2 = r_2 \alpha = r \alpha \quad (\text{Center at } O')$$

$$d_3 = r_3 \alpha = r \alpha \quad (\text{Center at } O')$$

$$x_2 = x_1 - x_3 = (1 - \beta)R = (1 - \beta)R \sin \alpha = R \sin \alpha$$

$$x_2 = r \alpha = R \sin \alpha$$

Ex-21 Let $\beta = \beta_0 = \frac{R}{r} \alpha_0$

Report: $x_1 = \frac{Rr}{\sin \beta}$ (Analytical solution)

α_0	β	no. solutions
$\pi/2$	$2\pi/3$	0
$\pi/3$	$\pi/2$	1
$\pi/4$	$3\pi/4$	2
$\pi/6$	$\pi/3$	3
$\pi/8$	$\pi/8$	4

¹ See B. P. Chang and C. H. Liang, "Computer Solution of Double-Link Inverse-Positioning Problems," *IEEE Transactions on Education*, vol. E-11, pp. 107-111, November 1968.

Chapter 11

Waveguides and Cavity Resonators

Ex 11-1 $\vec{H} \times \vec{E} = -j\omega\mu\vec{H}_t$ (1)
 $\vec{H} \times \vec{E} = j\omega\vec{E}_t$ (2)

From (1) $(\vec{E}_t \times \vec{E}_t) + (\vec{E}_t \times \vec{E}_z) = -j\omega\mu(\vec{H}_t + \vec{H}_z\hat{z})$

$\vec{E}_t \times \vec{E}_z = -j\omega\mu\vec{H}_t$

$\vec{E}_t \times \vec{E}_z = \vec{H}_t \times \vec{E}_z = -j\omega\mu\vec{H}_t$ (3)
 $(\because \vec{E}_t \times \vec{E}_z) = \vec{E}_t \times \vec{E}_z$

Similarly from (2) we obtain

$\vec{H}_t \times \vec{E}_z = j\omega\vec{H}_t \times \vec{E}_z = j\omega\mu\vec{H}_t$ (4)

Combining (3) and (4), we have

$j\omega\mu\vec{H}_t = (\vec{H}_t \times \vec{E}_z) = \vec{H}_t \times \vec{E}_z = j\omega\mu\vec{H}_t$

$\vec{H}_t = -\frac{1}{j\omega\mu} (\vec{H}_t \times \vec{E}_z - \vec{H}_t \times j\omega\mu\vec{H}_t)$ (5)

Similarly, $\vec{E}_t = -\frac{1}{j\omega\mu} (\vec{H}_t \times \vec{E}_z + \vec{H}_t \times j\omega\mu\vec{H}_t)$

Ex 11-2



From Eq. (11-28a):

$$\left(\frac{\beta}{k_0}\right)^2 = \left(\frac{\omega}{\omega_c}\right)^2 - 1$$

From Eq. (11-28b):

$$\left(\frac{\alpha}{k_0}\right)^2 = \left(\frac{\omega}{\omega_c}\right)^2 - 1$$

Both are equivalent of a unit circle.

(b)



From Eq. (11-28a):

$$\left(\frac{\beta}{k_0}\right)^2 = 1 - \frac{1}{\left(\frac{\omega}{\omega_c}\right)^2}$$

From Eq. (11-28b):

$$\left(\frac{\alpha}{k_0}\right)^2 = 1 - \frac{1}{\left(\frac{\omega}{\omega_c}\right)^2}$$



From Eq. (10-11):

$$\left(\frac{d\phi}{dr}\right) = \frac{(\partial \Delta F}{\partial r})_{T=0}$$

$$\text{At } r_0, \phi_0 = -1.07$$

$$\phi_0/a = -0.07$$

$$r_0/a = 0.34$$

$$\beta U_0 = 0.46$$

$$U_0/a^3 = 0.07$$

Ex. 11. a) For parallel-plate approximation:



$$\phi(r) = \phi_0 \left(\frac{r}{r_0}\right)^2$$

$$\phi_0 = \frac{2\pi\epsilon_0 q^2}{\epsilon_0}$$

$$\phi_1 = \frac{2\pi\epsilon_0 q^2}{\epsilon_1}$$

$$\phi_2 = \frac{2\pi\epsilon_0 q^2}{\epsilon_2}$$

Disintegrative parameter β and α adjust both ϕ_0 and the slope of the $\phi(r)$ curves, it affects α

but not the slope of high-temperature βU_0 .

Ex. 12. Field expansion for U_0 mode, from Eqs. (10-11) and (10-12):

$$E_x^0(r) = A_0 \cos(\alpha_0 r)$$

$$E_x^1(r) = \frac{A_1}{2} A_0 \cos(\alpha_0 r)$$

$$E_x^2(r) = -\frac{A_2}{2} A_0 \cos(\alpha_0 r)$$

Surface charge densities:

$$\sigma_0 = \epsilon_0 \cdot E_x^0|_{r=a} = \epsilon_0 A_0 \sin(\alpha_0 a) = -\frac{A_0}{2} A_0$$

$$\sigma_1 = \epsilon_0 \cdot E_x^1|_{r=a} = -\epsilon_0 A_1 \sin(\alpha_0 a) = \epsilon_0 \frac{A_1}{2} A_0$$

Surface current densities:

$$j_0 = \epsilon_0 \cdot \dot{E}_x^0|_{r=a} = \epsilon_0 \cdot A_0 \alpha_0 \sin(\alpha_0 a) = \epsilon_0 \frac{A_0}{2} \alpha_0 A_0$$

$$j_1 = \epsilon_0 \cdot \dot{E}_x^1|_{r=a} = -\epsilon_0 \cdot A_1 \alpha_0 \sin(\alpha_0 a) = \epsilon_0 \alpha_0 \frac{A_1}{2} A_0 \sin(\alpha_0 a) \cos(\alpha_0 a)$$

Ex. 10.1 Field expressions for \vec{H}_0 modes, from Eqs. (10.104)–(10.106):

$$E_z^{(0)}(y) = E_0 \cos(k_y y),$$

$$E_y^{(0)}(y) = \frac{1}{k_y} E_0 \sin(k_y y),$$

$$E_x^{(0)}(y) = -\frac{1}{k_y^2} E_0 \cos(k_y y),$$

$$\vec{H}_0 = \vec{H}_y + H_z \hat{z} = H_0 \hat{z},$$

$$\vec{H}_0 = -\vec{H}_y + H_z \hat{z} = H_0 \cos(k_y y) \hat{z}, \quad \begin{cases} \vec{H}_0 \text{ for } n \text{ odd,} \\ \vec{H}_0 \text{ for } n \text{ even.} \end{cases}$$

Ex. 10.2 Plot \vec{H}_0 and its field expressions in problem 10.1.



Ex. 10.3 Plot \vec{H}_0 and its field expressions in problem 10.1.



Ex. 10.4 Using the field expressions in problem 10.1, show:

$$\vec{E}_0 \cdot \vec{H}_0 = \int_{-a}^a \int_{-a}^a (\vec{E}_0 \cdot \vec{H}_0) dy dz = \int_{-a}^a \int_{-a}^a (E_0 H_0 \cos^2(k_y y)) dy dz.$$

$$\vec{E}_0 \cdot \vec{E}_0 = \int_{-a}^a \int_{-a}^a (E_0^2 \cos^2(k_y y)) dy dz = \frac{1}{2} E_0^2 \int_{-a}^a \int_{-a}^a \cos^2(k_y y) dy dz.$$

$$\int_{-a}^a \int_{-a}^a \vec{E}_0 \cdot \vec{H}_0 dy dz = \frac{1}{2} E_0 H_0 \int_{-a}^a \int_{-a}^a \cos^2(k_y y) dy dz \quad (\text{per unit guide length})$$

$$\int_{-a}^a \int_{-a}^a \vec{E}_0 \cdot \vec{E}_0 dy dz = \frac{1}{2} E_0^2 \int_{-a}^a \int_{-a}^a \cos^2(k_y y) dy dz.$$

$$\int_{-a}^a \int_{-a}^a \vec{E}_0 \cdot \vec{E}_0 dy dz = \frac{1}{2} E_0^2 \int_{-a}^a \int_{-a}^a \cos^2(k_y y) dy dz = \int_{-a}^a \int_{-a}^a \vec{E}_0 \cdot \vec{H}_0 dy dz \quad (\text{per unit guide length})$$

$$\text{From Ex. 10.1:} \quad \int_{-a}^a \int_{-a}^a \vec{E}_0 \cdot \vec{H}_0 dy dz = \frac{1}{2} E_0 H_0 \int_{-a}^a \int_{-a}^a \cos^2(k_y y) dy dz = \frac{1}{2} E_0 H_0 \int_{-a}^a \int_{-a}^a \cos^2(k_y y) dy dz.$$

which is the same as Eq. (10.108).

Ex. 10 Given: $\beta = 2.00 \times 10^3$ (radial), $\alpha = 1.00$, $\mu = 1$,
 $r = 10^3$ (radial), $\lambda = 2.00 \times 10^3$ (rad), $\rho = 10^3$ (rad).

(a) 1st mode

$$\begin{aligned} \beta &= \alpha \sqrt{\lambda} \cdot \rho = 2.00 \times 10^3 \text{ (radial)}, \\ \alpha_1 &= \frac{\beta}{\rho} \sqrt{\frac{\lambda}{\rho}} = 2.00 \times 10^3 \text{ (radial)}, \\ \alpha_2 &= \frac{\beta}{\rho} \sqrt{\frac{\lambda}{\rho}} = 2.00 \times 10^3 \text{ (radial)}, \\ \alpha_3 &= \alpha_1 = \alpha_2 = \frac{\beta}{\rho} = 2.00 \times 10^3 \text{ (radial)}, \\ \alpha_4 &= \alpha_1 = \frac{\beta}{\rho} = 2.00 \times 10^3 \text{ (radial)}. \end{aligned}$$

(b) 2nd mode — $C_1 \lambda_{02} = \frac{\beta^2}{\lambda_{02}^2} = 2.00 \times 10^3$ (rad) $\neq \rho$.

$$C_1 = \sqrt{1 - \lambda_{02}^2 / \rho^2} = 0.9999.$$

$$\begin{aligned} \beta &= \alpha \sqrt{\lambda} \cdot \rho = 2.00 \times 10^3 \text{ (radial)}, \\ \alpha_1 &= \frac{\beta}{\rho} \sqrt{\frac{\lambda}{\rho}} = 2.00 \times 10^3 \text{ (radial)}, \\ \alpha_2 &= \frac{\beta}{\rho} \sqrt{\frac{\lambda}{\rho}} = \frac{\beta}{\rho} \sqrt{\frac{\lambda}{\rho}} = 2.00 \times 10^3 \text{ (radial)}, \\ \alpha_3 &= \alpha_1 = 2.00 \times 10^3 \text{ (radial)}, \\ \alpha_4 &= \alpha_1 = 2.00 \times 10^3 \text{ (radial)}, \\ \alpha_5 &= \alpha_1 = 2.00 \times 10^3 \text{ (radial)}, \\ \alpha_6 &= \alpha_1 = 2.00 \times 10^3 \text{ (radial)}. \end{aligned}$$

(c) 3rd mode — $C_1 \lambda_{03} = \frac{\beta^2}{\lambda_{03}^2} = 2.00 \times 10^3$ (rad) $\neq \rho$.

$$C_1 = \sqrt{1 - \lambda_{03}^2 / \rho^2} = 0.9999.$$

$$\begin{aligned} \beta &= \alpha \sqrt{\lambda} \cdot \rho = 2.00 \times 10^3 \text{ (radial)}, \\ \alpha_1 &= \frac{\beta}{\rho} \sqrt{\frac{\lambda}{\rho}} = 2.00 \times 10^3 \text{ (radial)}, \\ \alpha_2 &= \frac{\beta}{\rho} \sqrt{\frac{\lambda}{\rho}} = 2.00 \times 10^3 \text{ (radial)}, \\ \alpha_3 &= \alpha_1 = 2.00 \times 10^3 \text{ (radial)}, \\ \alpha_4 &= \alpha_1 = 2.00 \times 10^3 \text{ (radial)}, \\ \alpha_5 &= \alpha_1 = 2.00 \times 10^3 \text{ (radial)}, \\ \alpha_6 &= \alpha_1 = 2.00 \times 10^3 \text{ (radial)}. \end{aligned}$$

Ex. 11 (a) **2nd mode** — $C_1 \lambda_{02} = C_2 \lambda_{02} = 2.00 \times 10^3$ (radial).

All required parameters are the same as above for the 2nd mode in problem 10. If λ_{02} is equal to ρ , using $C_1 = C_2 = 1$, we have

$$\alpha_1 = \frac{\beta}{\rho} \sqrt{\frac{\lambda}{\rho}} = \frac{\beta}{\rho} \sqrt{\frac{\lambda}{\rho}} = 2.00 \times 10^3 \text{ (radial)}$$

6) EM mode ——— $(\mathcal{L})_{EM} = (\mathcal{L})_{EM} = \frac{1}{2} \epsilon_0 \dot{\phi}^2 - \epsilon_0 \phi \Delta \phi$ & $\rho = 0$.

All required quantities are the same as those for the EM mode in problem 2 but \mathcal{L} , except ρ_0 .

$$\rho_0 = \frac{1}{4\pi} \int \frac{\partial \mathcal{L}}{\partial \phi} = \left(\frac{1}{4\pi} \right) \epsilon_0 \Delta \phi = \epsilon_0 \Delta \phi \quad (\text{Eqn. 6})$$

Ex 20.10 For TM_z mode in a parallel-plate waveguide,

$$\begin{aligned} \rho_0 &= \frac{1}{4\pi} \int \frac{\partial \mathcal{L}}{\partial \phi} = \frac{1}{4\pi} \frac{\partial}{\partial \phi} \left(\frac{1}{2} \epsilon_0 \dot{\phi}^2 - \epsilon_0 \phi \Delta \phi \right) \\ &= \frac{1}{4\pi} \int \frac{\partial \mathcal{L}}{\partial \phi} = \frac{1}{4\pi} \epsilon_0 \Delta \phi \end{aligned}$$

where $\Delta \phi = -k^2 \phi$, $k = \sqrt{k_x^2 + k_z^2}$.

6) To find minimum ρ_0 , set

$$\frac{\partial \rho_0}{\partial k} = 0 = \frac{1}{4\pi} \epsilon_0 k \quad \rightarrow \quad k = \frac{1}{2\pi}$$

$$\therefore \quad f = \sqrt{2} \frac{1}{2\pi}$$

8) At $k_x/f = 0.5$, $\frac{1}{\sqrt{2}} = \frac{1}{2\pi}$ and

$$\text{and} \quad \text{min } \rho_0 = \frac{1}{4\pi} \epsilon_0 \int \frac{\partial \mathcal{L}}{\partial \phi}$$

9) For $\rho_0 = \epsilon_0 \Delta \phi$ (Eqn. 6) $k = \pi a^{-1}$ (Eqn. 9), $\rho_0 = \epsilon_0 \Delta \phi$, and $\mu_0 \rho_0 = \epsilon_0 \Delta \phi$.

$$(\mathcal{L})_{EM} = \frac{1}{4\pi} \epsilon_0 \dot{\phi}^2 - \epsilon_0 \phi \Delta \phi$$

$$\text{min } \rho_0 = \epsilon_0 \Delta \phi \quad (\text{Eqn. 6})$$

Ex 20.11 Parallel-plate waveguide: incident ρ_0 for $\omega^2 < \omega_c^2$.

6) EM mode

From Eqs. (1) and (2):

$$\begin{cases} \mathcal{L}_T^2 = \mathcal{L}_z \\ \mathcal{L}_T^2 = \frac{1}{2} \epsilon_0 \dot{\phi}^2 \end{cases}$$

$$\rho_0 = \frac{1}{4\pi} \int \mathcal{L}_T^2 \mathcal{L}_z^{-1} \rho_0 = \frac{1}{4\pi} \frac{1}{2} \epsilon_0 \dot{\phi}^2$$

Substituting straight up into $\Delta \phi = -k^2 \phi$ (Eqn. 9)

$$\Delta \phi = \left(\frac{\rho_0}{\epsilon_0} \right) = \frac{1}{8\pi} \epsilon_0 \dot{\phi}^2 = -k^2 \phi \quad (\text{Eqn. 9}) = -k^2 \phi \quad (\text{Eqn. 9})$$

ii) TM₁₀ mode

From Eq. (20-10) and (20-11)

$$\begin{cases} E_z^0(x,y) = E_0 \cos\left(\frac{\pi x}{a}\right) \\ H_z^0(x,y) = -\frac{E_0}{\eta_0} \frac{a}{\sqrt{a^2 - \lambda^2}} \sin\left(\frac{\pi x}{a}\right) \end{cases}$$

$$k_z = \frac{2\pi}{\lambda} \sqrt{1 - \left(\frac{\lambda}{a}\right)^2} \text{ rad.}$$

$$R_{TM} = \frac{E_z^0}{H_z^0} = E_0 \cos\left(\frac{\pi x}{a}\right) \frac{\eta_0 \sqrt{a^2 - \lambda^2}}{a} = \frac{\eta_0 \sqrt{a^2 - \lambda^2}}{a}$$

$$\text{Max. } \left(\frac{R_{TM}}{Z_0}\right) = \frac{a \sqrt{a^2 - \lambda^2}}{a \sqrt{a^2 - \lambda^2}} = 1 \text{ at } \lambda = a \text{ (cutoff wavelength)}$$

iii) TE₁₀ mode

From Eq. (20-19) and (20-21)

$$\begin{cases} E_z^0(x,y) = E_0 \sin\left(\frac{\pi x}{a}\right) \\ H_z^0(x,y) = \frac{E_0}{\eta_0} \sqrt{1 - \left(\frac{\lambda}{a}\right)^2} \cos\left(\frac{\pi x}{a}\right) \end{cases}$$

$$R_{TE} = \frac{E_z^0}{H_z^0} = E_0 \sin\left(\frac{\pi x}{a}\right) \frac{\eta_0}{E_0 \sqrt{1 - \left(\frac{\lambda}{a}\right)^2}} = \frac{\eta_0}{\sqrt{1 - \left(\frac{\lambda}{a}\right)^2}}$$

$$\text{Max. } \left(\frac{R_{TE}}{Z_0}\right) = \frac{a \sqrt{a^2 - \lambda^2}}{a \sqrt{a^2 - \lambda^2}} = 1 \text{ at } \lambda = a \text{ (cutoff wavelength)}$$

Ex. 20-12 a) TM₁₀ mode



b) TE₁₀ mode



— Electric Field Lines
- - - Magnetic Field Lines

Ex. 20-13 If you take (20-19) through (20-21) for TM₁₀ mode

$$E_z^0(x,y) = \frac{2E_0}{\sqrt{2}} \left(\frac{a}{2}\right) E_0 \cos\left(\frac{\pi x}{a}\right) \sin\left(\frac{\pi y}{b}\right)$$

$$E_y^0(x,y) = \frac{2E_0}{\sqrt{2}} \left(\frac{a}{2}\right) E_0 \sin\left(\frac{\pi x}{a}\right) \cos\left(\frac{\pi y}{b}\right)$$

$$I_y^c(x,y) = I_y^c(x,y) + A \left(\frac{b^2}{12} + y^2 \right)$$

$$I_y^c(x,y) = \frac{A b^3}{12} + A y^2 \left(\frac{b^2}{12} + y^2 \right)$$

$$I_y^c(x,y) = \frac{A b^3}{12} + A y^2 \left(\frac{b^2}{12} + y^2 \right)$$

a) *Parallel axis theorem:*

$$\begin{aligned} I_y^c(x,y) &= I_y^c(x,y) + A d^2 = I_y^c(x,y) + A d^2 \\ &= I_y^c(x,y) + A d^2 = I_y^c(x,y) + A d^2 \\ &= I_y^c(x,y) \end{aligned}$$

$$\begin{aligned} I_x^c(x,y) &= I_x^c(x,y) + A d^2 = I_x^c(x,y) + A d^2 \\ &= I_x^c(x,y) + A d^2 = I_x^c(x,y) + A d^2 \\ &= I_x^c(x,y) \end{aligned}$$



Ex-14 Rectangular cross-section: $a = 4$ cm, $b = 2.5$ cm.

$$I_p = \frac{a^4 + b^4}{12} = \frac{4^4 + 2.5^4}{12}$$

Moments with the centroid I_{x_c} and I_{y_c} are:

Moments	I_{x_c}	I_{y_c}	I_{x_c}	I_{y_c}
in cm ⁴	14.4	3.125	4.32	4.125

a) For $A = 10$ cm², the only propagating mode is TE_{10} .

b) For $A = 1$ cm², the propagating modes are:

$$TE_{10}, TE_{01}, TE_{20}, TE_{11}, \text{ and } TE_{02}$$

Solution: $\lambda_{\text{min}} = \alpha_1 \sqrt{1 - \frac{1}{\text{CIR}}}$.

For the TC_{20} mode, $\lambda_{\text{min}} = \frac{20}{25}$.

$\therefore \alpha_1 = \frac{20}{25} \sqrt{1 - \frac{1}{\text{CIR}}} = \frac{4}{5} \sqrt{1 - \frac{1}{\text{CIR}}}$.

Q.10.11 $\text{CIR}_{\text{min}} = \frac{1}{\alpha_1^2} \sqrt{\frac{1}{\text{CIR}} - \frac{1}{\text{CIR}^2}} = \frac{1}{\alpha_1^2} \sqrt{1 - \frac{1}{\text{CIR}}}$.

a) $\alpha_1 = 1$, $\text{CIR}_{\text{min}} = \sqrt{1 - \frac{1}{\text{CIR}}}$

b) $\alpha_1 = 2$, $\text{CIR}_{\text{min}} = \sqrt{1 - \frac{1}{\text{CIR}}}$

Mode	CIR _{min}
TC_{10}	1
$\text{TC}_{10}, \text{TC}_{20}$	2
$\text{TC}_{10}, \text{TC}_{30}$	3
TC_{10}	4
TC_{20}	5
TC_{30}	6

Mode	CIR _{min}
$\text{TC}_{10}, \text{TC}_{20}$	1
$\text{TC}_{10}, \text{TC}_{30}$	2
$\text{TC}_{20}, \text{TC}_{30}$	3
TC_{10}	4
TC_{20}	5
TC_{30}	6

Q.10.12 $\mu = 10 \times 10^3 \text{ rad/s}$, $\lambda = 0.1 \text{ m} = 10 \text{ cm}$.

Let $\alpha = kx$, where $\text{CIR}_{\text{min}} = \frac{1}{\alpha^2} \sqrt{1 - \frac{1}{\text{CIR}}}$.

a) $\text{CIR}_{\text{min}} = \frac{1}{\alpha^2} \sqrt{1 - \frac{1}{\text{CIR}}}$ for the desired TC_{20} mode.

For $\mu > \omega$ ($\lambda < \lambda_{\text{min}}$) is assumed.

The next higher-order mode is TC_{30} with $\text{CIR}_{\text{min}} = \frac{1}{\alpha^2} \sqrt{1 - \frac{1}{\text{CIR}}}$.

For $\mu < \omega$ ($\lambda > \lambda_{\text{min}}$) is assumed.

We choose $\alpha = 4.7 \text{ rad/m}$ and $k = 2.35 \text{ rad/m}$.

b) $\alpha_1 = \frac{1}{\sqrt{1 - \frac{1}{\text{CIR}}}} = 1.02 \text{ rad/m}$.

$\lambda_{\text{min}} = \frac{1}{\alpha_1} = 0.98 \text{ m} = 98 \text{ cm}$.

$\mu = \frac{\omega}{2\pi} = 1.59 \text{ MHz}$.

$\text{CIR}_{\text{min}} = \frac{1}{\alpha^2} \sqrt{1 - \frac{1}{\text{CIR}}} = 0.98 \text{ dB}$.

Ex 10.19 Given: $a = 1.2 \times 10^2$ (cm), $b = 2.5 \times 10^2$ (cm), $p = 3 \times 10^2$ (cm).

a) $A = \frac{1}{2} ab = \frac{1}{2} (1.2 \times 10^2)(2.5 \times 10^2)$

$A = \sqrt{1.2 \times 2.5} \times 10^4 = 1.5 \times 10^4$

$A_1 = A \sin \theta_1 = 1.5 \times 10^4 \sin 30^\circ = 0.75 \times 10^4$

$A_2 = A \sin \theta_2 = 1.5 \times 10^4 \sin 60^\circ$

$A_3 = A \sin \theta_3 = 1.5 \times 10^4 \sin 90^\circ$

$A_4 = A \sin \theta_4 = 1.5 \times 10^4 \sin 120^\circ$

$(A_1 A_2 A_3 A_4) = 0.75 \times 10^4 = 7500 = 7.5 \times 10^3$

b) $A = \frac{1}{2} ab = \frac{1}{2} (1.2 \times 10^2)(2.5 \times 10^2)$

$A = \sqrt{1.2 \times 2.5} \times 10^4 = 1.5 \times 10^4$

$A_1 = A \sin \theta_1 = 1.5 \times 10^4 \sin 30^\circ = 0.75 \times 10^4$

$A_2 = A \sin \theta_2 = 1.5 \times 10^4 \sin 60^\circ$

$A_3 = A \sin \theta_3 = 1.5 \times 10^4 \sin 90^\circ$

$A_4 = A \sin \theta_4 = 1.5 \times 10^4 \sin 120^\circ$

$(A_1 A_2 A_3 A_4) = \frac{0.75 \times 10^4}{2} = 3.75 \times 10^3$

Ex 10.20 Given: $a = 3 \times 10^2$ (cm), $b = 4 \times 10^2$ (cm), $p = 2 \times 10^2$ (cm)

a) $A_1 = 3 \times 10^2 \times 4 \times 10^2 = 12 \times 10^4$ (cm²)

$A_2 = \frac{1}{2} ab = 6 \times 10^4$ (cm²)

b) $A = \frac{1}{2} ab = 6 \times 10^4$ (cm²) $\sqrt{1 - \left(\frac{2}{4}\right)^2} = 0.7071$

$A_1 = \frac{6 \times 10^4}{0.7071} = 8.485 \times 10^4$ (cm²)

c) $A_2 = \frac{6 \times 10^4}{\sqrt{1 - \left(\frac{2}{3}\right)^2}} = 7.5 \times 10^4$ (cm²) $A_3 = \frac{6 \times 10^4}{\sqrt{1 - \left(\frac{2}{3}\right)^2}} = 7.5 \times 10^4$ (cm²)

d) $A_4 = \frac{6 \times 10^4}{\sqrt{1 - \left(\frac{2}{3}\right)^2}} = 7.5 \times 10^4$ (cm²)

Ex 10.21 Given: $a = 3 \times 10^2$ (cm), $b = 4 \times 10^2$ (cm), $p = 10^2$ (cm)

a) $A = \frac{1}{2} ab = 6 \times 10^4$ (cm²) $A_1 = 3 \times 10^2 \times 4 \times 10^2 = 12 \times 10^4$ (cm²)

$\sqrt{1 - \left(\frac{10^2}{4 \times 10^2}\right)^2} = \sqrt{1 - \left(\frac{1}{4}\right)^2} = 0.96875$

$A_2 = \frac{6 \times 10^4}{0.96875} = 6.193 \times 10^4$ (cm²) $\left[1 - \frac{10^4}{(4 \times 10^2)^2}\right]$

$\times 10^4 = 6.193 \times 10^4$ (cm²)

b) From Eqs. (20-42), (20-43) and (20-44):

$$E_0 = E_0 \sin\left(\frac{\pi z}{L}\right)$$

$$E_z = -\frac{E_0}{L} \sqrt{L^2 - z^2} \cos\left(\frac{\pi z}{L}\right)$$

$$E_r = \frac{E_0}{L} \frac{z}{r} \cos\left(\frac{\pi z}{L}\right)$$

$$E_{\text{av}} = \left(\frac{E_0}{L}\right)^2 \int_0^L (L^2 - z^2) dz = \frac{E_0^2}{L} \left(1 - \frac{1}{3}\right) L$$

For E_{av} a bit of the last sentence, assuming average squared conditions:

$$\langle E_z^2 \rangle = E_z = 0.250 \text{ (vol)}, \quad \langle E_r^2 \rangle = 0.250 \text{ (vol)}, \quad \langle E_\theta^2 \rangle = 0.250 \text{ (vol)}$$

The two greatest field eq. — The field intensities are higher at the center and by a factor of $2^{1/2} = 1.414$

$$\therefore \text{Max. } \langle E_z^2 \rangle = 0.250 \text{ (vol)}$$

$$\text{Max. } \langle E_r^2 \rangle = 0.250 \text{ (vol)}$$

$$\text{Max. } \langle E_\theta^2 \rangle = 0.250 \text{ (vol)}$$

$$\text{d) } P_{\text{total}} = E_0 = (E_z^2 + E_r^2 + E_\theta^2)_{\text{av}} = 0.250 \text{ (vol)} = 0.250 \left(\frac{\text{vol}}{\text{cm}^3}\right) \frac{E_0^2}{L^2}$$

$$(P_{\text{total}}) = 0.250 \text{ (vol)}$$

$$P_{\text{total}} = E_0 = (E_z^2 + E_r^2 + E_\theta^2)_{\text{av}} = 0.250 \text{ (vol)} = 0.250 \text{ (vol)}$$

$$(P_{\text{total}}) = (0.250 + 0.250) \frac{E_0^2}{L^2} = \frac{E_0^2}{L^2} \left(\frac{1}{L^2} + \frac{1}{L^2} \right) = \frac{E_0^2}{L^2} \left(\frac{2}{L^2} \right) = \frac{E_0^2}{L^2} \left(\frac{2}{L^2} \right)$$

$$\text{At the center and } \text{Max. } (P) = \frac{E_0^2}{L^2} \left(\frac{2}{L^2} \right) = 0.250 \text{ (vol)}$$

e) Total amount of average power dissipated in 1 cm of average pipe:

$$E_0 = 0.250 (1^2 - 0) = 0.250 (1^2 - 0) = 0.250 \text{ (vol)}$$

Solve From problem 2 (c) we have

$$E_{\text{av}} = \frac{E_0^2}{L} \left(1 - \frac{1}{3}\right) = \frac{E_0^2}{L} \left(\frac{2}{3}\right) = 0.250$$

$$\therefore \text{Max. } E_0 = \frac{(0.250)^2 \left(\frac{3}{2}\right)}{1 - \frac{1}{3}} = 0.250 = 0.250 \text{ (vol)}$$

Ex 11.1 Let $\alpha = \frac{1}{\sqrt{2}} \sqrt{\frac{2\pi\hbar^2}{m}}$ and $\psi = \frac{1}{\sqrt{2}} \psi_0(x)$ (11-111)

We write $\langle \psi, \hat{H}_0 \psi \rangle = E(\alpha)$, where $E(\alpha) = \frac{1 + \frac{3}{2}\alpha^2}{1 + \frac{3}{2}\alpha^2}$

For min. $\langle \psi, \hat{H}_0 \psi \rangle$, set $\frac{dE(\alpha)}{d\alpha} = 0$.

$$\implies \alpha = \frac{1}{\sqrt{2}} = \sqrt{\frac{1}{2}} \left[\sqrt{\frac{2\pi\hbar^2}{m}} \right]^{-1/2}$$

Ex 11.2 Trial expansion for \hat{H}_0 made from Eqs (11-111) and (11-112) is given by:

$$\psi_0^+(x, \alpha) = \frac{1}{\sqrt{2}} \left[\frac{1}{\sqrt{2}} \psi_0(x) \cos\left(\frac{\pi x}{\alpha}\right) \right],$$

$$\psi_1^+(x, \alpha) = \frac{1}{\sqrt{2}} \left[\frac{1}{\sqrt{2}} \psi_0(x) \sin\left(\frac{\pi x}{\alpha}\right) \right],$$

$$\psi_2^+(x, \alpha) = \frac{1}{\sqrt{2}} \left[\frac{1}{\sqrt{2}} \psi_0(x) \cos\left(\frac{2\pi x}{\alpha}\right) \right],$$

$$\psi_3^+(x, \alpha) = \frac{1}{\sqrt{2}} \left[\frac{1}{\sqrt{2}} \psi_0(x) \sin\left(\frac{2\pi x}{\alpha}\right) \right],$$

$$\psi_4^+(x, \alpha) = \frac{1}{\sqrt{2}} \left[\frac{1}{\sqrt{2}} \psi_0(x) \cos\left(\frac{3\pi x}{\alpha}\right) \right].$$

Calculate ψ_0 from Eq. (11-111): $\psi_0 = \frac{1}{\sqrt{2}} \psi_0$

$$E(0) = \frac{1}{2} \int_{-\alpha}^{\alpha} \left[\psi_0^+ \hat{H}_0 \psi_0 - \psi_0^+ \psi_0 \right] dx = \frac{m \frac{1}{2} \frac{1}{2} \alpha^2}{2 \int_{-\alpha}^{\alpha} \left[\frac{1}{2} - \frac{1}{2} \right] dx}$$

From problem 11-11:

$$\int_{-\alpha}^{\alpha} \psi_0^+ \psi_0 dx = \int_{-\alpha}^{\alpha} \frac{1}{\sqrt{2}} \left[\frac{1}{\sqrt{2}} \psi_0(x) \right] dx = \alpha$$

$$\int_{-\alpha}^{\alpha} \psi_0^+ \hat{H}_0 \psi_0 dx = \int_{-\alpha}^{\alpha} \frac{1}{\sqrt{2}} \left[\frac{1}{\sqrt{2}} \psi_0(x) \right] dx = \alpha$$

$$\int_{-\alpha}^{\alpha} \psi_0^+ \psi_0 dx = \int_{-\alpha}^{\alpha} \frac{1}{\sqrt{2}} \left[\frac{1}{\sqrt{2}} \psi_0(x) \right] dx = \alpha$$

$$\int_{-\alpha}^{\alpha} \psi_0^+ \hat{H}_0 \psi_0 dx = \int_{-\alpha}^{\alpha} \frac{1}{\sqrt{2}} \left[\frac{1}{\sqrt{2}} \psi_0(x) \right] dx = \alpha$$

$$\int_{-\alpha}^{\alpha} \psi_0^+ \hat{H}_0 \psi_0 dx = \int_{-\alpha}^{\alpha} \frac{1}{\sqrt{2}} \left[\frac{1}{\sqrt{2}} \psi_0(x) \right] dx = \alpha$$

$$E(0) = \frac{m \frac{1}{2} \frac{1}{2} \alpha^2}{2 \left[\frac{1}{2} - \frac{1}{2} \right]} \left[\left(\frac{1}{2} \right) \alpha - \left(\frac{1}{2} \right) \alpha \right]$$

$$\therefore \langle \psi, \hat{H}_0 \psi \rangle = \frac{m \frac{1}{2} \frac{1}{2} \alpha^2}{2 \left[\frac{1}{2} - \frac{1}{2} \right]} \left[\left(\frac{1}{2} \right) \alpha - \left(\frac{1}{2} \right) \alpha \right] \quad \text{Min. } \langle \psi, \hat{H}_0 \psi \rangle \text{ occurs at } \langle \psi, \hat{H}_0 \psi \rangle = 0.$$

Ex. 24 $f = \cos^2(\theta)$, $f_0 = \frac{1}{2}$, $\frac{df}{d\theta} = \frac{d(\cos^2 \theta)}{d\theta} = 2 \cos \theta (-\sin \theta) = -2 \cos \theta \sin \theta = -\sin(2\theta)$
 $\theta_2 = \frac{2\pi}{2} = \pi$ (at $\theta = \pi$, $f = \cos^2(\pi) = 1$)



At $\theta = \pi$, $f = \cos^2(\pi) = 1$,
 from the lead,
 which is represented by the
 point θ_2 at $\theta = \pi$.

Use Smith's relationship about
 how close a curve is to a straight
 line, indicating the position
 of θ_2 . Lead at $\theta_2 = \pi$ is 0.

Draw a straight line from θ through θ_2 , intersecting
 the perimeter of θ_2 . Lead at $\theta = \pi$ is the "maximum"
 forward lead. $\therefore \theta_2$ is at $\theta = \pi$ exactly, it is
 from θ_2 , the position of a min. lead. In other words,
 θ_2 the desired location of the lead, should be at
 $\theta = \pi$ or $\theta = 2\pi$, or $\theta = 0$, from the lead.

From Eq. (20) the relationship is to say $\theta = \pi$, $\theta = 2\pi$,

$$\theta = \pi = \frac{2\pi}{2} \text{ is } [2\pi] \text{ or } \theta = 2\pi \text{ or } \theta = 0$$

Ex. 25 $f = \cos^2(\theta)$, $\theta_2 = \pi$

<u>Substituted</u>	<u>Full form</u>
$\frac{d^2f}{d\theta^2} = -2\cos^2 \theta = -2\cos^2 \theta$ (1)	$2\cos^2 \theta = -2\cos^2 \theta$ (2)
$-2\cos^2 \theta = -2\cos^2 \theta = -2\cos^2 \theta$ (3)	$-2\cos^2 \theta = -2\cos^2 \theta$ (4)
$\frac{1}{2} [2\cos^2 \theta - 2\cos^2 \theta] = 0$ (5)	$\frac{1}{2} [2\cos^2 \theta - 2\cos^2 \theta] = 0$ (6)

Combining θ_2 from (1) and (2): $\theta_2 = \frac{2\pi}{2} = \pi$ (7)

Combining θ_2 from (3) and (4): $\theta_2 = \frac{2\pi}{2} = \pi$ (8)

Combining (5) and (6): $\theta_2 = \frac{2\pi}{2} = \pi$ (9)

Ex. 11.11 a)

TM₁



b) TE₁₁



———— Electric Field Lines

..... Magnetic Field Lines

$$\text{Ex. 11.11 c) } E_p \text{ (TM)}: E_z = \frac{h}{\gamma_{z0}^2} = \frac{h \cos^2 \theta}{\gamma^2}$$

$$\text{For TM}_{10} \text{ mode, } \gamma_{z0} = \frac{h}{a} \rightarrow \gamma_{z0}^2 = \frac{h^2}{a^2} \rightarrow E_z = \frac{h \cos^2 \theta}{\frac{h^2}{a^2}} = \frac{a \cos^2 \theta}{h}$$

$$\text{For TE}_{10} \text{ mode, } \gamma_{z0} = \frac{h}{a} \rightarrow \gamma_{z0}^2 = \frac{h^2}{a^2} \rightarrow E_z = \frac{h \cos^2 \theta}{\frac{h^2}{a^2}} = \frac{a \cos^2 \theta}{h}$$

Ex. 11.12 $\beta^2 = \beta^2 - \beta_c^2 = \omega^2 \mu \epsilon - \beta_c^2$

$$\text{For TE}_{10} \text{ mode, } \beta_c = \frac{\pi}{a} \rightarrow \beta_c^2 = \frac{\pi^2}{a^2} \rightarrow \beta^2 = \omega^2 \mu \epsilon - \frac{\pi^2}{a^2} \rightarrow \omega_c = \frac{\pi}{a \sqrt{\mu \epsilon}}$$

$$\text{For TM}_{10} \text{ mode, } \beta_c = \frac{\pi}{a} \rightarrow \beta_c^2 = \frac{\pi^2}{a^2} \rightarrow \beta^2 = \omega^2 \mu \epsilon - \frac{\pi^2}{a^2} \rightarrow \omega_c = \frac{\pi}{a \sqrt{\mu \epsilon}}$$



a) If a is doubled, ω_c and β_c in diagram (a) are halved but diagram (b) will remain the same.

b) If the dielectric medium is changed then ω_c and β_c (i.e., $\omega_c \sqrt{\mu \epsilon}$ and β_c) are reduced by a factor of $\sqrt{\epsilon}$ and the slope of the asymptote line is changed from $\frac{1}{a} \sqrt{\mu \epsilon}$, diagram (b) remains unchanged.

Ex 21 of Problems: $L_2^2 = C_2 L_2(\text{Re}) \neq 0$.

Boundary conditions $L_2^2 = 0$ at both ends and $\mu \in \mathbb{R}$ are satisfied when $n \in \mathbb{N}$ is integer.

There are no TL_2 modes.

(i) TL modes: $L_2^2 = C_2 L_2(\text{Re}) \cos \mu x$, where $\mu \in \mathbb{N}$.

TL_2 modes do exist.

(ii) For TL modes, S.E. at $x=0$ requires that $L_2(\text{Re})=0$
 \implies Eigenvalues $(TL)_{2, \mu} = \mu_{2, \mu}^2/a_2$, $\mu \in \mathbb{N}, \mu > 0$.

For TL modes, S.E. at $x=l$ requires that $L_2(\text{Re})=0$.

\implies Eigenvalues $(TL)_{2, \mu} = \mu_{2, \mu}^2/a_2$, $\mu \in \mathbb{N}, \mu > 0$.

Ex 22 From Eqs. (10-100) and (10-101)

Inside the slab: $\beta^2 = \mu^2/a_1^2 + k_1^2 = \mu^2/a_1^2$.

Outside the slab: $\beta^2 = \mu^2/a_2^2 + k_2^2 = \mu^2/a_2^2$.

$$\therefore \mu \sqrt{a_1^2} + \beta = \mu \sqrt{a_2^2},$$

$$\text{and } \frac{\mu}{\sqrt{a_1^2}} + \mu \sqrt{\beta} = \frac{\mu}{\sqrt{a_2^2}}.$$

Ex 23 From Eqs. (10-120) and (10-121)

$$\left(\frac{X^2}{a_1^2}\right) + \left(\frac{Y^2}{a_2^2}\right) = \left(\frac{\mu^2}{a_1^2}\right) \left(\frac{a_2^2}{a_1^2} - 1\right). \quad \text{--- (1)}$$

$$\frac{X^2}{a_1^2} + \left(\frac{Y^2}{a_2^2}\right) \cos^2 \theta = \left(\frac{\mu^2}{a_1^2}\right) \cos^2 \theta. \quad \text{--- (2)}$$

Let $X = a_1 \alpha$, $Y = a_2 \beta$, $\alpha = \mu \sin \theta$, and $\beta = \frac{\mu}{\sqrt{a_1^2}} \frac{a_2^2}{a_1^2} \cos^2 \theta$.

Eqs. (1) and (2) become $\begin{cases} \alpha^2 + \beta^2 = \mu^2, & \text{--- (3)} \\ Y = a_2 \beta \cos \theta. & \text{--- (4)} \end{cases}$

(i) $\beta = \mu \cos^2 \theta$, $\alpha = \mu \sin \theta = \mu \sqrt{1 - \cos^2 \theta}$.

$a_2 \beta \cos \theta = \mu \cos^3 \theta = \mu \cos \theta$, $\alpha = \mu \sqrt{1 - \cos^2 \theta} = \mu \sin \theta$.



Graphical solution:
 $x_2^* = 0.2000$, $x_1^* = 1.0000 \text{ m}^3$
 $w = 0.2000 = 0.2000 \text{ (kg/m}^3\text{)}$
 $\beta = 0.2000 = 0.2000 \text{ (m}^3\text{/kg)}$

From Eq. (2-112) $\beta = \beta_0 \sqrt{1 - x_2^*} = 0.2000 \text{ (m}^3\text{/kg)}$

4) $f = 0.2000^2 \text{ (kg)}$, $w = 0.2000 \text{ (kg/m}^3\text{)}$, $\beta = 0.2000 \text{ (m}^3\text{/kg)}$.

$$x_1 = 0.2000, \quad x_2 = 0.2000$$

$$x_2^* = 0.2000, \quad x_1^* = 1.0000 \text{ m}^3$$

We obtain $w = 0.2000 \text{ (kg/m}^3\text{)}$.

$$\beta = 0.2000 \text{ (m}^3\text{/kg)}$$

Ex. 11. From Eq. (2-111):

$$\left(\frac{dx_1}{dt}\right) = -\frac{1}{\beta} \left(\frac{dx_2}{dt}\right) \text{ and } \left(\frac{dx_2}{dt}\right) \quad \text{--- (1)}$$

Using the relations in problem 2.10(1), we obtain two equations from (1) in 2.10-1) and (2) above:

$$\begin{cases} x_1^* + x_2^* = 0^* & \text{--- (2)} \\ x_2^* = 0.2000 \text{ and } x_1^* & \text{--- (3)} \end{cases}$$

4) $f = 0.2000^2 \text{ (kg)}$, $w = 0.2000 \text{ (kg/m}^3\text{)}$, $\beta = 0.2000 \text{ (m}^3\text{/kg)}$.

$$x_1 = 0.2000,$$

$$x_2 = 0.2000,$$

$$x_1 = 0.2000,$$

$$x_2 = 0.2000.$$

There are no intersections for curves representing Eqs. (2) and (3); hence, even the roots do not exist at the given composition.

Ex. 12. Use Eqs. (2-111) and (2-112)

$$x_1^* = -\frac{dx_2}{dt} \frac{dx_1}{dx_2}, \quad x_2^* = \frac{dx_1}{dt} \frac{dx_2}{dx_1}$$

$$f(x_1, x_2) = 0.20 \left[\frac{dx_1}{dt} \frac{dx_2}{dx_1} - \frac{dx_2}{dt} \frac{dx_1}{dx_2} \right]$$

$$f(x_1, x_2) = 0.20 \left[\frac{dx_1}{dt} \frac{dx_2}{dx_1} - \frac{dx_2}{dt} \frac{dx_1}{dx_2} \right]$$

Table 1:

$$\begin{aligned}
\mathcal{L}_1^2(\alpha) &= \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \longrightarrow \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \\
\mathcal{L}_1^3(\alpha) &= \mathcal{L}_1^2(\alpha) \otimes \mathcal{L}_1(\alpha) \longrightarrow \mathcal{L}_1^2(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \\
\mathcal{L}_1^4(\alpha) &= \mathcal{L}_1^3(\alpha) \otimes \mathcal{L}_1(\alpha) \longrightarrow \mathcal{L}_1^3(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha)
\end{aligned}$$

Table 2:

$$\begin{aligned}
\mathcal{L}_1^2(\alpha) &= \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \\
\mathcal{L}_1^3(\alpha) &= \mathcal{L}_1^2(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \\
\mathcal{L}_1^4(\alpha) &= \mathcal{L}_1^3(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha)
\end{aligned}$$

Table 3:

$$\begin{aligned}
\mathcal{L}_1^2(\alpha) &= \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \\
\mathcal{L}_1^3(\alpha) &= \mathcal{L}_1^2(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \\
\mathcal{L}_1^4(\alpha) &= \mathcal{L}_1^3(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha)
\end{aligned}$$

Table 4) From Table 1 and Table 2 it is seen that $\mathcal{L}_1^2(\alpha)$ for \mathbb{Z}_2 mode, which is the dominant mode.

From Eq. (10-11):

$$w = \frac{\mathcal{L}_1^2(\alpha)}{\mathcal{L}_1^2(\alpha)} \otimes \frac{\mathcal{L}_1^2(\alpha)}{\mathcal{L}_1^2(\alpha)} = \frac{\mathcal{L}_1^4(\alpha)}{\mathcal{L}_1^4(\alpha)}, \text{ for } \mathbb{Z}_2 \text{ mode}$$

Multiplying the w^2 term in Eq. (10-11):

$$w^2 = \mathcal{L}_1^4(\alpha) \otimes \mathcal{L}_1^4(\alpha) = w \otimes w = \mathcal{L}_1^8(\alpha) \otimes \mathcal{L}_1^8(\alpha)$$

From Eq. (10-11): $\mathcal{L}_1^2(\alpha) \otimes \mathcal{L}_1^2(\alpha) = \mathcal{L}_1^4(\alpha) = \mathcal{L}_1^2(\alpha)$

$$\therefore w = \frac{\mathcal{L}_1^4(\alpha)}{\mathcal{L}_1^4(\alpha)} = \mathcal{L}_1^2(\alpha)$$

2) when $\mathbb{Z}_2 \otimes \mathbb{Z}_2$, $\mathbb{Z}_2 \otimes \mathbb{Z}_2$, $\mathbb{Z}_2 \otimes \mathbb{Z}_2$, $\mathbb{Z}_2 \otimes \mathbb{Z}_2$, $\mathbb{Z}_2 \otimes \mathbb{Z}_2$:

$$w = \frac{\mathcal{L}_1^2(\alpha)}{\mathcal{L}_1^2(\alpha)} \otimes \mathcal{L}_1^2(\alpha) = \mathcal{L}_1^2(\alpha) \otimes \mathcal{L}_1^2(\alpha)$$

$$w^2 = \mathcal{L}_1^4(\alpha) \otimes \mathcal{L}_1^4(\alpha), \quad w \otimes w = \mathcal{L}_1^4(\alpha) \otimes \mathcal{L}_1^4(\alpha)$$

$$\longrightarrow (w \otimes w) = \mathcal{L}_1^4(\alpha) \otimes \mathcal{L}_1^4(\alpha)$$

Table 1 See Eqs. (19-21) and (22-23)

$$\begin{aligned} \epsilon_1^* &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} & \epsilon^* &= \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} \\ \epsilon_{1, \text{odd}}^* &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} & \epsilon^* &= \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} \\ \epsilon_{1, \text{even}}^* &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} & \epsilon^* &= \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} \end{aligned}$$

(19-21)

$$\begin{aligned} \epsilon_1^* \text{ odd } \epsilon_{1, \text{odd}}^* &\longrightarrow \epsilon_{1, \text{odd}}^* = \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} \\ \epsilon_1^* \text{ even } \epsilon_{1, \text{even}}^* &\longrightarrow \epsilon_{1, \text{even}}^* = \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} \\ \epsilon_1^* \text{ odd } \epsilon_{1, \text{odd}}^* &\longrightarrow \epsilon_{1, \text{odd}}^* = \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} \end{aligned}$$

(22-23)

$$\begin{aligned} \epsilon_1^* \text{ odd } \epsilon_{1, \text{odd}}^* &\longrightarrow \epsilon_{1, \text{odd}}^* = \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} \\ \epsilon_1^* \text{ even } \epsilon_{1, \text{even}}^* &\longrightarrow \epsilon_{1, \text{even}}^* = \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} \\ \epsilon_1^* \text{ odd } \epsilon_{1, \text{odd}}^* &\longrightarrow \epsilon_{1, \text{odd}}^* = \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} \end{aligned}$$

(24-25)

$$\begin{aligned} \epsilon_1^* \text{ odd } \epsilon_{1, \text{odd}}^* &\longrightarrow \epsilon_{1, \text{odd}}^* = \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} \\ \epsilon_1^* \text{ even } \epsilon_{1, \text{even}}^* &\longrightarrow \epsilon_{1, \text{even}}^* = \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} \\ \epsilon_1^* \text{ odd } \epsilon_{1, \text{odd}}^* &\longrightarrow \epsilon_{1, \text{odd}}^* = \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} \end{aligned}$$

Setting $\omega = \beta$ in $\epsilon_1^* = \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta}$ and
in $\epsilon^* = \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta}$

and expanding, we obtain

$$\begin{aligned} \epsilon_1^* &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} \Big|_{\omega=\beta} = \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} \Big|_{\omega=\beta} \\ &\longrightarrow \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} = \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \end{aligned}$$

Table 2 a) Odd TE and even TE modes are also propagating modes. Using the Eq. (1) in the formulas in Table 1, we have

$$\epsilon_1^* = \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} \quad \text{for odd TE modes,}$$

$$\epsilon_1^* = \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \beta} \quad \text{for even TE modes,}$$

4) Dielectric Dielectric — From Eq. (2) and (3):

$$\text{for } r < a: \quad \epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0$$

$$\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0$$

Find charge density in conductor $\rho = \epsilon_0 \cdot \nabla \cdot \mathbf{E}$

$$\rho = -\epsilon_0 \nabla^2 \psi = -\epsilon_0 \frac{\rho}{\epsilon_1 \epsilon_0} \epsilon_1$$

Find charge density in conductor $\rho = \epsilon_0 \cdot \nabla \cdot \mathbf{E}$

$$\rho = \epsilon_0 \nabla^2 \psi = -\frac{\rho}{\epsilon_1} \epsilon_1$$

Dielectric Dielectric — From problem 4 (a):

$$\text{for } r < a: \quad \epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0$$

$$\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0$$

$$\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0$$

$$\therefore \rho = \epsilon_0 \nabla \cdot \mathbf{E} = \epsilon_0 \nabla \cdot \left(-\frac{\rho}{\epsilon_1 \epsilon_0} \mathbf{r} \right) = -\frac{\rho}{\epsilon_1} \epsilon_1$$

$$\rho = \epsilon_0 \nabla \cdot \mathbf{E} = 0$$

4.11.11 Dielectric Dielectric — From problem 4 (a):

for $r < a$: $\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0}$ and $\psi = 0$ at $r = a$

$$\left[\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \right] \Rightarrow \left[\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \right] \Rightarrow \left[\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \right] \quad (1)$$

$$\left[\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \right] \Rightarrow \left[\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \right] \Rightarrow \left[\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \right] \quad (2)$$

$$\text{From (1) & (2): } \left[\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \right] \Rightarrow \left[\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \right] \Rightarrow \left[\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \right]$$

$$\left[\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \right] \Rightarrow \left[\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \right] \Rightarrow \left[\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \right]$$

Similarly, for $r > a$:

$$\epsilon_2 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0 \text{ at } r = a$$

$$\epsilon_2 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0 \text{ at } r = a$$

Boundary conditions: $\left. \begin{array}{l} \text{for } r = a \\ \text{at } r = a \end{array} \right\} \begin{array}{l} \epsilon_1 \nabla \psi = \epsilon_2 \nabla \psi \\ \psi = \psi \end{array} \quad (3)$

$$\frac{\rho}{\epsilon_0} \rightarrow \text{Characteristic equation } \frac{\rho}{\epsilon_0} = -\frac{\rho}{\epsilon_1} \frac{\rho}{\epsilon_0}$$

Solve From Eq. (20-103): $R_{\text{avg}} = \frac{1}{2} \sqrt{\frac{200}{100} + \frac{100}{100} - \frac{200}{100}}$

$$R_{\text{avg}} = 0.200 \text{ m} \text{ (round)} \quad R_{\text{range}} = \sqrt{\frac{200}{100} - \frac{100}{100} - \frac{200}{100}}$$

Least-order modes and resonant frequencies:

Mode	R_{range}	$(kR_{\text{avg}})^2$
$TM_{0,0}$	0.200	2.000000
$TE_{0,0}$	0.117	1.360000
$TE_{1,0}$	0.149	1.480000
$TM_{1,0}$	0.200	2.000000
$TE_{2,0}$	0.200	2.000000
$TM_{2,0}$	0.231	2.140000
$TE_{3,0}$	0.254	2.320000
$TM_{3,0}$	0.274	2.480000
$TE_{4,0}$	0.289	2.640000
$TM_{4,0}$	0.307	2.790000
$TE_{5,0}$	0.322	2.940000

Ex. 20-10 From Ex. 20-9, the quarter-wave resonant mode is $TE_{0,0}$ mode.

$$R_{\text{avg}} = \frac{1}{2} \sqrt{\frac{200}{100} + \frac{100}{100}} = 0.250 \text{ m} \quad (20)$$

b) From Eq. (20-104):

$$\begin{aligned} Q_{\text{avg}} &= \frac{R_{\text{range}} \sqrt{200000000} (200)}{2 \times (100)^2 \times (200) \times (200)} \quad \left(Q_{\text{avg}} = \sqrt{\frac{200000000}{100}} \right) \\ &= \frac{0.117 \times 141421.356237}{2 \times 100^2 \times 200 \times 200} = 0.016 \end{aligned}$$

From Eqs. (20-97) and (20-101):

$$W_0 = \frac{1}{2} \times 0.117^2 \times 200 \times 10^6 = 0.14811225 \text{ W} \quad (21)$$

$$W_{\text{avg}} = \frac{1}{2} \times 0.016 \times \left(\frac{1}{2} \times 10^6 \right) = 0.000800000 \text{ W} = 0.8 \text{ mW} \quad (22)$$

Solve (i) $(V_{\text{ext}})_z = \frac{7}{2} \sqrt{\frac{1}{2} + \frac{1}{2}} = \frac{7}{2} (V_{\text{ext}})_z = 4.9 \text{ eV} \approx 10 \text{ eV}$

(ii) $(V_{\text{ext}})_z = \frac{7}{2} \sqrt{\frac{1}{2} - \frac{1}{2}} = 0 \text{ eV}$

(iii) $(V_{\text{ext}})_z - (V_{\text{ext}})_z = 4.9 \text{ eV} \approx 10 \text{ eV}$
 $= (V_{\text{ext}})_z$

Solve (a) Considering Eqs. (20-29) and (20-30)

$$V_{\text{ext}} = \frac{qV}{R^2} \frac{b \cos^2 \theta + d \sin^2 \theta}{(2b \cos^2 \theta + d \sin^2 \theta)^{3/2}}$$

$\rightarrow V_{\text{ext}}$ has a symmetrical dependence on θ and θ' . It will be maximum when $\theta = \theta'$, which gives a max. value-to-surface ratio

(b) When $\theta = \theta'$, $V_{\text{ext}} = \frac{qV}{R^2} \frac{1}{2R(1+b/a)}$

Solve (i) $V_{\text{ext}} = \frac{qV \cos^2 \theta + d \sin^2 \theta}{2a \cos^2 \theta + d \sin^2 \theta}$

For $a=d=10$, $V_{\text{ext}} = \frac{qV}{2(10)} \sqrt{\frac{1}{2} + \frac{1}{2}} = 0.5 \text{ eV} \left(\frac{1}{2} \right)$

$V_{\text{ext}} = 0.5 \sqrt{2}$

(ii) For $V_{\text{ext}} = 1.414 V_{\text{ext}}$, $V_{\text{ext}}/a = 1.414$

Solve (i) From the field configuration in the cavity we see that the V_{ext} made with respect to a is the same as the V_{ext} made with respect to b . Thus, $(V_{\text{ext}})_{\text{ext}}$ can be obtained from $(V_{\text{ext}})_z$ in Eq. (20-29) by changing b to a and a to b .

(ii) (b) If for the V_{ext} in (a) can be derived from the field expression in Eqs. (20-29), (20-30), (20-31) by setting $\theta = \theta'$, and using Eq. (20-30).

$$V = 2V_0 = \frac{qV}{R^2} \left(\frac{2b^2}{2a} \right) \sin^2 \theta \quad \text{or } V_{\text{ext}}$$

$$V_0 = \frac{1}{2} \left(\frac{qV}{R^2} \right) \sin^2 \theta = \frac{1}{2} \left(\frac{qV}{R^2} \right) \sin^2 \theta$$

$$\begin{aligned}
 R_1 &= R_1 \left(\int_0^1 \frac{1}{x} dx \right)^2 = R_1 \ln^2 2 = R_1 \ln^2 2 \\
 &= R_1 \left(\int_0^1 \frac{1}{x} dx + \int_0^1 \frac{1}{x} dx \right) = R_1 \ln^2 2 \\
 &= R_1 \left(\frac{1}{2} \ln^2 2 + \frac{1}{2} \ln^2 2 \right) = R_1 \ln^2 2
 \end{aligned}$$

$$R_{\text{eff}} = \frac{R_1 \ln^2 2}{2} = \frac{R_1 \ln^2 2 (1 + \ln^2 2)}{2(1 + \ln^2 2)} = R_1 \frac{\ln^2 2}{2}$$

Example 11 TM₁₀ mode:

$$E_z^2 = E_0^2 \sum_{m=0,2,4,\dots}^{\infty} \cos^2 \left(\frac{m\pi x}{a} \right)$$

$$m = 0, 2, 4, \dots \quad n = 0, 1, 2, \dots \quad \beta = m\pi/a$$

$$C_{\text{TM}_{10}} = \frac{1}{2} \left(\frac{1}{2} + \frac{1}{2} \right) = \frac{1}{2}$$

$$C_{\text{TM}_{10}} = \frac{1}{2} \left(\frac{1}{2} + \frac{1}{2} \right)$$

$$\text{TM}_{10} \text{ mode: } E_z^2 = E_0^2 \sum_{m=0,2,4,\dots}^{\infty} \cos^2 \left(\frac{m\pi x}{a} \right)$$

$$m = 0, 2, 4, \dots \quad n = 0, 1, 2, \dots \quad \beta = m\pi/a$$

$$C_{\text{TM}_{10}} = \frac{1}{2} \left(\frac{1}{2} + \frac{1}{2} \right) = \frac{1}{2}$$

$$C_{\text{TM}_{10}} = \frac{1}{2} \left(\frac{1}{2} + \frac{1}{2} \right)$$

1) For TM₁₀, the dominant mode is TM₁₀ / C_{TM₁₀} = $\frac{1}{2} \frac{1}{2} = \frac{1}{4}$

The first three modes with equal power are:

Mode	TM ₁₀	TM ₂₀	TM ₃₀	TM ₄₀	TM ₅₀	TM ₆₀	TM ₇₀
E_z^2	1.00	0.25	0.11	0.06	0.04	0.03	0.02

$$\text{Example 12} \quad C = \frac{1}{2} = \frac{1}{2} \left(\frac{1}{2} + \frac{1}{2} \right) \quad L = \frac{1}{2} \ln \left(\frac{1}{2} \right)$$

$$1) L_1 = \frac{1}{2} \ln \left(\frac{1}{2} \right) = \frac{1}{2} \ln \left(\frac{1}{2} \right)$$

$$2) L_2 = \frac{1}{2} \ln \left(\frac{1}{2} \right) = \frac{1}{2} \ln \left(\frac{1}{2} \right)$$

Chapter II

Antennas and Radiating Systems

Ex 1 Maxwell equations for dipole antenna:

$$\nabla \times \mathbf{E} = -\mu_0 \frac{\partial \mathbf{H}}{\partial t} \quad \text{--- (1)}$$

$$\nabla \times \mathbf{H} = \mathbf{J} + \frac{\partial \mathbf{D}}{\partial t} \quad \text{--- (2)}$$

$$\nabla \cdot \mathbf{E} = \frac{\rho}{\epsilon_0} \quad \text{--- (3)}$$

$$\nabla \cdot \mathbf{H} = 0 \quad \text{--- (4)}$$

a) $\nabla \times (2)$: $\nabla \times (\nabla \times \mathbf{E}) = \mu_0 \frac{\partial}{\partial t} (\mathbf{J} + \frac{\partial \mathbf{D}}{\partial t})$
 $= \mu_0 \frac{\partial \mathbf{J}}{\partial t} + \mu_0 \frac{\partial^2 \mathbf{D}}{\partial t^2}$ --- (5)

But $\nabla \times (\nabla \times \mathbf{E}) = \nabla(\nabla \cdot \mathbf{E}) - \nabla^2 \mathbf{E}$
 $= \frac{1}{\epsilon_0} \nabla \rho - \nabla^2 \mathbf{E}$ --- (6)

Combining (5) and (6), we obtain

$$\nabla^2 \mathbf{E} - \mu_0 \frac{\partial^2 \mathbf{D}}{\partial t^2} = \frac{1}{\epsilon_0} \nabla \rho - \mu_0 \frac{\partial \mathbf{J}}{\partial t}$$

b) Similarly, we have $\nabla^2 \mathbf{H} - \mu_0 \frac{\partial^2 \mathbf{J}}{\partial t^2} = -\nabla \times \mathbf{J}$

Ex 2 \mathbf{A}_D (vector) $\mathbf{A} = -\nabla \psi - \mu_0 \int \mathbf{J}_D \cdot d\mathbf{l}_D + \mu_0 \mathbf{A}_D + \mu_0 \mathbf{A}_B$

$$\mathbf{A}_D = -\frac{\mu_0}{4\pi} \int \frac{\mathbf{J}_D}{r} dV_D$$

$$\mathbf{A}_B = -\frac{\mu_0}{4\pi} \int \frac{\mathbf{J}_B}{r} dV_B$$

$$\mathbf{A}_S = -\frac{\mu_0}{4\pi} \int \frac{\mathbf{J}_S}{r} dV_S$$

The expansion of $\mathbf{A}_D, \mathbf{A}_B$,
and \mathbf{A}_S are given in Ex.
13-15(a, b, c, d).



$$\psi = \frac{1}{4\pi\epsilon_0} \int \frac{\rho}{r} dV = \frac{1}{4\pi\epsilon_0} \int \frac{\rho}{r} dV$$

$$\mathbf{A}_D = \mu_0 \int \frac{\mathbf{J}_D}{r} dV_D$$

$$\mathbf{A}_B = \mu_0 \int \frac{\mathbf{J}_B}{r} dV_B$$

$$\mathbf{A}_S = \mu_0 \int \frac{\mathbf{J}_S}{r} dV_S$$

$$\psi = \frac{1}{4\pi\epsilon_0} \int \frac{\rho}{r} dV = \frac{1}{4\pi\epsilon_0} \int \frac{\rho}{r} dV$$

$$= \frac{1}{4\pi\epsilon_0} \int \frac{\rho}{r} dV = \frac{1}{4\pi\epsilon_0} \int \frac{\rho}{r} dV$$

$$V = \frac{1}{2} \int_{-a}^a \frac{d^2}{dx^2} (1 + 2x^2 + 3x^4) dx$$

$$= \frac{1}{2} \int_{-a}^a (4x + 12x^3) dx$$

Using A_1, A_2, A_3 and V in R_1, R_2 and R_3 , we obtain the same results as given in Eqs. (1)-(3).

Ex. 1



$$d = \sqrt{a^2 + b^2}$$

$$= \sqrt{a^2 + b^2}$$

$$d^2 = a^2 + b^2$$

$$2d^2 = 2a^2 + 2b^2$$

$$d^2 - a^2 = b^2$$

$$d^2 - a^2 = b^2$$

$$d^2 - a^2 = b^2$$

$$d^2 = a^2 + b^2$$

$$d = \sqrt{a^2 + b^2}$$

$$d^2 = a^2 + b^2$$

$$d^2 - a^2 = b^2$$

$$\frac{d^2 - a^2}{d^2} = \frac{b^2}{d^2}$$

$$= \frac{b^2}{a^2 + b^2}$$

In the same manner, we have

$$\frac{d^2 - b^2}{d^2} = \frac{a^2}{a^2 + b^2}$$

$$\therefore \frac{d^2 - a^2}{d^2} = \frac{b^2}{a^2 + b^2}$$

$$\frac{d^2 - b^2}{d^2} = \frac{a^2}{a^2 + b^2}$$

Let $m = \frac{a}{b} = \frac{d}{b}$

$$d = \frac{a}{m} = \frac{b}{m}$$

$$= \frac{a}{m} = \frac{b}{m}$$

$$\begin{aligned} \text{a) } \vec{D} &= \frac{1}{\epsilon_0} \epsilon_0 \vec{E} = \epsilon_0 \vec{E}_0 + \epsilon_0 \vec{E}_p && \text{Expression for } \vec{E}_0, \vec{E}_p \\ \text{b) } \vec{E} &= \frac{1}{\epsilon_0 \epsilon_0} \vec{F} = \vec{E}_0 + \vec{E}_p && \text{and } \vec{E}_p \text{ given as above} \\ &&& \text{given in Ex. 3 (b), (c), (d)} \end{aligned}$$

In the far zone, $r \gg r_0$, $\vec{E}(\vec{r}, t)$ and $\vec{H}(\vec{r}, t)$ terms can be neglected. We have the following instantaneous expressions assuming $\vec{E}(\vec{r}, t) \approx \vec{E}_0(\vec{r}, t)$:

$$\begin{aligned} \vec{E}(\vec{r}, t) &= \epsilon_0 \frac{\partial^2 \vec{p}(t - r/c)}{\partial t^2} \text{ (in the far zone)} \\ \vec{H}(\vec{r}, t) &= \epsilon_0 \frac{\partial^2 \vec{p}(t - r/c)}{\partial t^2} \times \hat{r} \text{ (in the far zone)} \\ \vec{H}(\vec{r}, t) &= \epsilon_0 \frac{\partial^2 \vec{p}(t - r/c)}{\partial t^2} \text{ (in the far zone)} \end{aligned}$$

Ex. 10 Far-zone electric field of elemental electric dipole

$$\vec{E}_0(\vec{r}, t) = \frac{1}{4\pi\epsilon_0} \left(\frac{\partial^2 \vec{p}(t - r/c)}{\partial t^2} \right) \frac{1}{r^2} \text{ (in the far zone)} = \frac{1}{4\pi\epsilon_0} \left(\frac{\partial^2 \vec{p}(t - r/c)}{\partial t^2} \right) \frac{1}{r^2} \text{ (in the far zone)}$$

For the elemental magnetic dipole:

$$\vec{E}_0(\vec{r}, t) = \frac{1}{4\pi\epsilon_0} \left(\frac{\partial^2 \vec{p}(t - r/c)}{\partial t^2} \right) \frac{1}{r^2} \text{ (in the far zone)} = \frac{1}{4\pi\epsilon_0} \left(\frac{\partial^2 \vec{p}(t - r/c)}{\partial t^2} \right) \frac{1}{r^2} \text{ (in the far zone)}$$

$$\text{a) (b) } \frac{\frac{\partial^2 \vec{p}(t - r/c)}{\partial t^2}}{\left(\frac{\partial^2 \vec{p}(t - r/c)}{\partial t^2} \right) / r^2} = \frac{\frac{\partial^2 \vec{p}(t - r/c)}{\partial t^2}}{\left(\frac{\partial^2 \vec{p}(t - r/c)}{\partial t^2} \right) / r^2} = 1$$

————— Electric polarization

b) Circular polarization of $\vec{E} = E_0 \hat{e}_\theta$

$$\text{Ex. 11} \text{ Equation of continuity: } \vec{\nabla} \cdot \vec{J} + \dot{\rho} = 0 \text{ — } \vec{J} = \frac{1}{\epsilon_0} \frac{\partial \vec{p}}{\partial t}$$

$$\text{a) } \vec{\nabla} \cdot \vec{J} = \dot{\rho} \text{ (in the far zone)} \text{ — } \vec{J} = -j \frac{1}{\epsilon_0} \vec{E}_0 \sin \theta \text{ — } j \frac{1}{\epsilon_0} \vec{E}_0 \sin \theta$$

$$\text{b) } \vec{\nabla} \cdot \vec{J} = \dot{\rho} \text{ (in the far zone)} \text{ — } \vec{J} = \left(\frac{1}{\epsilon_0} \frac{\partial \vec{p}}{\partial t} \right) \sin \theta$$

$$\text{Ex. 12} \quad \vec{E} = \frac{1}{4\pi\epsilon_0} \frac{\partial^2 \vec{p}(t - r/c)}{\partial t^2} \text{ (in the far zone)} \text{ — } \frac{\partial^2 \vec{p}}{\partial t^2} = \frac{\partial^2 \vec{p}}{\partial t^2} \text{ (in the far zone)}$$

$$\text{a) Radiation resistance } R_r = 80\pi^2 \left(\frac{d}{\lambda} \right)^2 \text{ (in the far zone)}$$

$$\text{b) } R_r(\text{in the far zone}) = 80\pi^2 \left(\frac{d}{\lambda} \right)^2 = 80\pi^2 \left(\frac{d}{\lambda} \right)^2 \text{ (in the far zone)}$$

$$R_r(\text{in the far zone}) = 80\pi^2 \left(\frac{d}{\lambda} \right)^2 = 80\pi^2 \left(\frac{d}{\lambda} \right)^2 \text{ (in the far zone)}$$

$$\text{c) } R_r(\text{in the far zone}) = 80\pi^2 \left(\frac{d}{\lambda} \right)^2 \text{ (in the far zone)} \text{ — } R_r(\text{in the far zone}) = 80\pi^2 \left(\frac{d}{\lambda} \right)^2 \text{ (in the far zone)}$$

$$R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \left[\sin^{-1} x \right]_0^1 = \frac{\pi}{2}$$

$$= \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx$$

$$= \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$$

$$R_2 = \frac{\pi}{2} = \frac{1}{\sqrt{1-x^2}} dx$$

$$dx = \frac{1}{\sqrt{1-x^2}} dx$$

In case $\theta = \sin^{-1} x$, $\frac{d\theta}{dx} = \frac{1}{\sqrt{1-x^2}}$, and $dx = \sqrt{1-x^2} d\theta$.

$$\therefore R_2 = \int_0^{\pi/2} \frac{1}{\sqrt{1-x^2}} \sqrt{1-x^2} d\theta = \int_0^{\pi/2} 1 d\theta = \theta \Big|_0^{\pi/2} = \frac{\pi}{2}$$

$$R_2 = \int_0^{\pi/2} 1 d\theta = \frac{\pi}{2}$$

$$R_2 = \int_0^{\pi/2} 1 d\theta = \frac{\pi}{2}$$

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$$R_2 = \int_0^{\pi/2} 1 d\theta = \frac{\pi}{2}$$

$$R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{\pi}{2}$$

- $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{\pi}{2}$, over constant growth.
- From $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx$, $\frac{dR_2}{dx} = \frac{1}{\sqrt{1-x^2}}$, $R_2 = \frac{1}{\sqrt{1-x^2}}$, $R_2 = \frac{1}{\sqrt{1-x^2}}$.
- $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{\pi}{2}$, $R_2 = \frac{1}{\sqrt{1-x^2}}$, $R_2 = \frac{1}{\sqrt{1-x^2}}$.
- $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{\pi}{2}$.

$$R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{\pi}{2}$$

- $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{\pi}{2}$, $R_2 = \frac{1}{\sqrt{1-x^2}}$, $R_2 = \frac{1}{\sqrt{1-x^2}}$, $R_2 = \frac{1}{\sqrt{1-x^2}}$.
- $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{\pi}{2}$, $R_2 = \frac{1}{\sqrt{1-x^2}}$, $R_2 = \frac{1}{\sqrt{1-x^2}}$.

From problem 1, $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{\pi}{2}$, $R_2 = \frac{1}{\sqrt{1-x^2}}$.

- For $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{\pi}{2}$, $R_2 = \frac{1}{\sqrt{1-x^2}}$, $R_2 = \frac{1}{\sqrt{1-x^2}}$.

Ex. 10-20 a) $L_1 = \int_{-\pi/2}^{\pi/2} (\frac{d}{dt} \sin t)^2 dt = 2 \int_{-\pi/2}^{\pi/2} \cos^2 t dt$

b) $L_2 = \int_{-\pi/2}^{\pi/2} \cos^2 t dt$, $L_3 = \int_{-\pi/2}^{\pi/2} \sin^2 t dt$

c) For $\theta = t$ ($0 \leq t \leq \pi/2$), $L_1 = L_2 = \int_0^{\pi/2} \cos^2 t dt$ and $L_3 = \int_0^{\pi/2} \sin^2 t dt$
 $\theta = 2\pi - t$ ($\pi/2 \leq t \leq \pi$), $L_1 = \int_{\pi/2}^{\pi} \cos^2 t dt$ from part a,
 $L_2 = \int_{\pi/2}^{\pi} \sin^2 t dt$, $L_3 = \int_{\pi/2}^{\pi} \cos^2 t dt$
 $\rightarrow L_1 = 2 \int_0^{\pi/2} \cos^2 t dt$

Ex. 10-21 $R = \int_0^{\pi} r^2 dr = \int_0^{\pi} \frac{1}{2} (2a \cos t)^2 dt = a^2 \int_0^{\pi} \cos^2 t dt$

$= \frac{a^2}{2} \int_0^{\pi} (1 + \cos 2t) dt = \frac{a^2}{2} \left[\int_0^{\pi} 1 dt + \int_0^{\pi} \cos 2t dt \right]$
 $= \frac{a^2}{2} \left[\pi + \frac{\sin 2t}{2} \Big|_0^{\pi} \right]$

$= \frac{a^2}{2} \left[\pi + \frac{\sin 2\pi}{2} - \frac{\sin 0}{2} \right]$, which is the same as Eq. (10-44)

Ex. 10-22 $r_{\text{avg}} = \frac{1}{\pi} \int_0^{\pi} r dt = \frac{2a}{\pi}$



For $\theta = \pi/2$, $r = 2a$

$r_{\text{avg}} = \frac{1}{\pi} \int_0^{\pi} 2a(1 - \cos t) dt$

Width of major beam between the first nulls
 $= 2 \times \frac{2a}{\pi} = \frac{4a}{\pi}$

Ex. 10-23 $L = \int_0^{\pi} r^2 dt = \int_0^{\pi} a^2 (1 - \cos t)^2 dt$

From Eq. (10-44), $L_1 = \frac{a^2}{2} \int_0^{\pi} (1 + \cos 2t) dt = a^2 \int_0^{\pi} \cos^2 t dt$

$= \frac{a^2}{2} \int_0^{\pi} (1 + \cos 2t) dt$ (use part a)

$= \frac{a^2}{2} \left[\pi + \frac{\sin 2t}{2} \Big|_0^{\pi} \right]$

Maximum L_1 occurs at $\theta = \frac{\pi}{2}$, where

$L_1 \left(\frac{\pi}{2} \right) = \pi - \frac{\pi^2}{4}$

Ex 11-10 a) $V_{\text{ext}} = -kx^2 = -\frac{d}{2} \left[\frac{\cos(\frac{1}{2} \pi \cos \theta)}{\frac{1}{2} \pi \cos \theta} \right]$

b) $V = \frac{1}{2} \left[\frac{d^2 V_{\text{ext}}}{dx^2} \right]_{x_0} = \frac{d^2 V_{\text{ext}}}{dx^2} = \frac{d^2}{dx^2} \left[\frac{\cos(\frac{1}{2} \pi \cos \theta)}{\frac{1}{2} \pi \cos \theta} \right]$

which has a maximum value $2k^2 \cos^2(\theta) \sin^2(\theta)$ at $\theta = \frac{\pi}{4}$

c) For $\lambda = \frac{d^2 V_{\text{ext}}}{dx^2} = 2d^2 \cos^2 \theta \sin^2 \theta$ and $V_0 = \frac{d^2}{2} \cos^2 \theta$ (value)

$\theta = \frac{\pi}{4} \rightarrow V_0 \left(\frac{\pi}{4} \right) = \frac{d^2}{8} = 0.125 d^2, \quad V_{\text{ext}} = -0.125 d^2$

$\lambda = \frac{d^2 V_{\text{ext}}}{dx^2} = 0.125 d^2 = 0.125 d^2$

$\theta = \frac{\pi}{4} \rightarrow V_0 \left(\frac{\pi}{4} \right) = \frac{d^2}{8} \left[\frac{\cos(\frac{1}{2} \pi \cos \theta)}{\frac{1}{2} \pi \cos \theta} \right] = 0.125 d^2$

$V_{\text{ext}} = -0.125 d^2, \quad \lambda = 0.125 d^2$

Ex 11-11



if $V = d - d \cos \theta$

so $V_{\theta} = \frac{d^2 V}{d\theta^2} = d \sin \theta \cos \theta$

(for $\theta = \frac{\pi}{2}$)

$= \frac{d^2 V}{d\theta^2} = d \sin \theta \cos \theta = 0$

where $V(\theta) = d - d \cos(\theta)$

a) $d = 1.0$

$V(\theta) = (1 - \cos \theta)$



c) $d = 1$

$V(\theta) = (1 - \cos 2\theta)$



Ex-10



From Ex-9 we have
 $L_1 = \frac{a_1x + b_1y + c_1}{\sqrt{a_1^2 + b_1^2}}$ and $L_2 = \frac{a_2x + b_2y + c_2}{\sqrt{a_2^2 + b_2^2}}$

where

$$\begin{aligned} \sin \alpha &= |L_1 - L_2| \\ &= \left| \frac{a_1x + b_1y + c_1}{\sqrt{a_1^2 + b_1^2}} - \frac{a_2x + b_2y + c_2}{\sqrt{a_2^2 + b_2^2}} \right| \\ &= \left| \frac{a_1x + b_1y + c_1}{\sqrt{a_1^2 + b_1^2}} - \frac{a_2x + b_2y + c_2}{\sqrt{a_2^2 + b_2^2}} \right| \\ &= \frac{|c_1\sqrt{a_2^2 + b_2^2} - c_2\sqrt{a_1^2 + b_1^2}|}{\sqrt{a_1^2 + b_1^2} \sqrt{a_2^2 + b_2^2}} \end{aligned}$$

$$L_1 - L_2 = \frac{a_1x + b_1y + c_1}{\sqrt{a_1^2 + b_1^2}} - \frac{a_2x + b_2y + c_2}{\sqrt{a_2^2 + b_2^2}}$$

As the distance from the origin to the perpendicular line is 'd',

- In the region: $d = 0$, $L_1 = L_2 = 0$
- In the region: $d > 0$, $L_1 > L_2$ (not in region)
- In the region: $d < 0$, $L_1 < L_2$ (not in region and not in region)
- It is a line, $d = 0$



$L_1 > L_2$ (not in region)



$L_1 < L_2$ (not in region)

Ex 217 From Eq. (20.11) $M(\psi) = \frac{d^2}{dt^2} [\cos(\psi) \cos(\omega t)]$, where

$$\psi = \beta \cos(\omega t) \text{ (small)}$$

In the absence of a signal, $\omega = 0$, $M(\psi) = 0$.
 With $\omega = \frac{1}{2} \beta$:

$$M(\psi) = \cos(\frac{1}{2}\beta \cos(\omega t))$$



$$M(\psi) = \cos(\frac{1}{2}\beta \cos(\omega t))$$

$$M(\psi) = \cos(\frac{1}{2}\beta \cos(\omega t))$$



Ex 218 a) Relative deviation amplitude: $\beta = 0.1 \text{ rad}$.

b) Array factor: $M(\psi) = \cos(\frac{1}{2}\beta \cos(\omega t))$



$$M(\psi) = \cos(\frac{1}{2}\beta \cos(\omega t))$$

$$\approx 1 - \frac{1}{8}\beta^2 \cos^2(\omega t)$$

$$\approx 1 - \frac{1}{8}\beta^2 \cos(2\omega t)$$

$$\approx 1 - \frac{1}{8}\beta^2 \cos(2\omega t)$$

$$\approx 1 - \frac{1}{8}\beta^2 \cos(2\omega t)$$

For a signal array, from Eq. (20.11):

$$\frac{d^2}{dt^2} \left[\frac{\cos(\frac{1}{2}\beta \cos(\omega t))}{\cos(\frac{1}{2}\beta \cos(\omega t))} \right] = \frac{d^2}{dt^2} \cos(\frac{1}{2}\beta \cos(\omega t))$$

With power beamwidth for 3-dB signal without array
 with $\beta = 0.1$ (spacing = 2λ) $\approx 2\lambda \cos^2(\omega t) = 2\lambda \cos^2(\omega t)$

Ex 219 a) From Eq. (20.11) the array: $M(\psi) = \frac{d^2}{dt^2} \left[\frac{\cos(\frac{1}{2}\beta \cos(\omega t))}{\cos(\frac{1}{2}\beta \cos(\omega t))} \right]$



1) Resonance Operation $\phi = \phi_{res} = 0$

$$|Z_{eq}| = \frac{1}{\omega} \left| \frac{R + j\omega L}{1 - \omega^2 LC} \right| = \left| \frac{R}{\omega} \right| \quad \text{for } \phi = 0$$

where $Z = R \parallel \omega L$

At half-power points: $\left| \frac{R}{\omega} \right| = \frac{1}{\sqrt{2}} R \implies \omega = 0.707 R$

(For each bandwidth, another operating)

For bandwidth operation, the half-power bandwidth

$$\Delta \omega_{BW} = 0.707 \left(\frac{R}{L} \right) \text{ rad/s}$$

$$= 0.707 \left(\frac{R}{L} \right) \text{ rad/s}$$

For $\omega = 1$, $\Delta \omega_{BW} = 0.707 \left(\frac{R}{L} \right) \text{ rad/s}$

From Eq. (1) & (2): $\Delta \omega_{BW} = 0.707 \left(\frac{R}{L} \right) \text{ rad/s}$

2) Resonance Operation $\phi = \phi_{res} = 0$

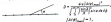
$$|Z_{eq}| = \frac{1}{\omega} \left| \frac{R + j\omega L}{1 - \omega^2 LC} \right| = \frac{R}{\omega} \text{ rad/s}$$

For $\omega = 1$, $|Z_{eq}| = \frac{R}{\omega} \text{ rad/s}$

From Eq. (1) & (2): $|Z_{eq}| = \frac{R}{\omega} \text{ rad/s}$

3) Resonance Operation $\phi = \phi_{res} = 0$

Resonance operation: $\phi = \phi_{res} = 0$



$$\int_0^{\infty} \left(\frac{R}{\omega} \right)^2 \sin^2 \omega t \, d\omega = \frac{R^2}{\omega} \int_0^{\infty} \left(\frac{1}{\omega} \right)^2 \sin^2 \omega t \, d\omega = \frac{R^2}{\omega}$$

$\therefore Z = \frac{R}{\omega} = \frac{R}{\omega}$, where $\omega = \text{angular frequency}$

4) Construction follows the steps outlined in pp. 107-108



ϕ is at $\theta = \frac{\pi}{4}$

Radius of circle is
 $R = \frac{R}{\omega} = \frac{R}{\omega}$

Ex-21 a) $A(x) = r + qz$, $z = e^{i\theta}$ ψ at π

b) $A(z) = (r + qz)^2$ A double zero $\psi_1 = \psi_2 = \pi$

c) $A(z) = \frac{z^2 + 1}{z - 1}$ Simple ψ_1 at π , $\psi_2 = \pi + 2\pi$

d) $A(z) = (r + qz)^2 = \left(\frac{z^2 + 1}{z - 1}\right)^2$

Double zeros $\psi_1 = \psi_2 = \pi$

$\psi_3 = \psi_4 = \pi + 2\pi$

- e) The zeros of an array polynomial specify the nulls in the array pattern as θ changes from 0 to 2π . If $\theta = (\pi + 2\pi)$ gives the main beam. The regions between the nulls (except the main beam region) are side lobes. The double zeros of the array in part (d) are more widely spaced leading to a wider beamwidth and lower side lobes (width) are half lower than those of a three-element uniform array.

Ex-22 From Eqs. (1)-(3) and (3)-(4):

$$|E_{\theta}| = \frac{2\pi \eta_0 I_0 a^2 \cos^2 \theta}{r^2} \left| \sum_{n=0}^{N-1} A_n e^{i(n\theta - \beta r_n)} \right|$$

where $A_n = \frac{1}{a} \frac{\sin(\beta a \cos \theta)}{\beta \cos \theta}$, $\beta = \frac{2\pi}{\lambda} \sin \theta$

$A_n = \frac{1}{a} \frac{\sin(\beta a \cos \theta)}{\beta \cos \theta}$, $\beta = \frac{2\pi}{\lambda} \sin \theta$

$$|E_{\theta}| = \frac{2\pi \eta_0 I_0 a^2}{r^2} \left| \left[\frac{\sin(N\beta a \cos \theta)}{\sin \beta a \cos \theta} \right] \frac{\sin(\beta a \cos \theta)}{\beta \cos \theta} \right|$$

Ex-23 From Eqs. (1) and (2) $E_{\theta}(z) = \frac{2\pi \eta_0 I_0 a^2}{r^2} \left[\frac{\sin(N\beta a \cos \theta)}{\sin \beta a \cos \theta} \right]$ (1)

Using Eqs. (2) & (3) $E_{\theta}(z) = \frac{2\pi \eta_0 I_0 a^2}{r^2} \left[\frac{\sin(N\beta a \cos \theta)}{\sin \beta a \cos \theta} \right]$ (2)

d) Substituting (2) in Eq. (1) $E_{\theta}(z) = \frac{2\pi \eta_0 I_0 a^2}{r^2} \left[\frac{\sin(N\beta a \cos \theta)}{\sin \beta a \cos \theta} \right]$

e) Also value of $A_n(z)$ and $A_n(r) = \frac{2\pi \eta_0 I_0 a^2}{r^2}$

For $\beta = \frac{2\pi}{\lambda} \cos \theta$, $\theta = 0^\circ$, $A_n(z) = \frac{2\pi \eta_0 I_0 a^2}{r^2}$

Ex:12 a) $E_p = 10^{-12} \text{ J}$, $E_p = \frac{1}{2} m v^2 \left(\frac{1}{\sqrt{1 - \frac{v^2}{c^2}}} \right)$ (1)

$$E_p = \frac{1}{2} m_0 v^2 \left(\frac{1}{\sqrt{1 - \frac{v^2}{c^2}}} \right) = \frac{1}{2} m_0 v^2 \gamma$$

$$= \frac{1}{2} m_0 v^2 \left(\frac{1}{\sqrt{1 - \frac{v^2}{c^2}}} \right) \quad (2)$$

$$\therefore E_p = \frac{1}{2} m_0 v^2 = \frac{1}{2} m_0 v^2 \left(\frac{1}{\sqrt{1 - \frac{v^2}{c^2}}} \right) = \frac{1}{2} m_0 v^2$$

Ex:13 From eq (1) & (2): $\frac{E_p}{E_0} = \left(\frac{1}{\sqrt{1 - \frac{v^2}{c^2}}} \right) \frac{E_0}{E_0}$

a) For half mass at poles: $E_p = E_0 = 100 \text{ J}$

$$E = 2 \times 10^2 \text{ J} \quad \therefore m_0 v^2 = 100 \text{ J} \quad \therefore v = \sqrt{100} \text{ m/s}$$

$$E_p = 2 \times 10^2 \text{ J} = 2 \times 10^2 \text{ J} \quad (2) = 100 \text{ J} \text{ (given)}$$

b) For Newtonian dynamics: $E_p = E_0 = 100 \text{ J}$

$$E_p = 100 \text{ J} \text{ (given)}$$

Ex:14 From given Earth radius = 6370 km

Gravitational acceleration at 6370 km = 9.8 m/s²

$$g = \frac{GM}{R^2} \Rightarrow \frac{GM}{(6370 \times 10^3)^2} = 9.8 \text{ m/s}^2$$

$$g = \frac{GM}{R^2} \Rightarrow \frac{GM}{R^2} = 9.8 \text{ m/s}^2$$

a) For particles would never stay

$$\therefore (2g)^2 - (2\pi R)^2 < 0$$

Use above condition to solve

$$R = 6370 \times 10^3 \text{ m}$$

$$g = 9.8 \text{ m/s}^2 \Rightarrow \text{cannot escape the planet surface}$$



Gravitational field

b) Let $E_1 =$ Power transmitted by satellite antenna

$$E_2 = \text{Power density within the cone} = \frac{E_1}{4\pi R^2 \sin^2 \theta}$$

$$\text{Area of cone cap on earth} = \pi \int_0^{\theta} r^2 \sin \theta d\theta$$

$$= \pi r^2 (1 - \cos \theta) \text{ (area of cap)} = \pi r^2 (1 - \cos \theta)$$

$$\therefore E_2 = \frac{E_1}{4\pi R^2 \sin^2 \theta} \Rightarrow \theta = \frac{1}{2} \sqrt{\frac{E_1}{E_2 R^2}} = \frac{1}{2} \sqrt{\frac{E_1}{E_2 R^2}}$$

$$\text{Major axis diameter} = 2R = 4R \sin^2 \theta$$

EX-11 (i) From Eq. (1) and (2) $A_1 = \frac{2.0 \times 10^4}{2.0 \times 10^4} A_2$

$$A_1 = \frac{2.0 \times 10^4}{2.0 \times 10^4} A_2 = 2.0 \times 10^4 \text{ cm}^2, \quad A_2 = 2.0 \times 10^4 \text{ cm}^2$$

$$A_3 = \frac{2.0 \times 10^4}{2.0 \times 10^4} A_2 = 2.0 \times 10^4 \text{ cm}^2, \quad A_4 = 2.0 \times 10^4 \text{ cm}^2$$

$$r = 2.0 \times 10^4 \text{ cm}, \quad R_1 = 2.0 \times 10^4 \text{ cm}$$

$$\text{--- } R_2 = 2.0 \text{ cm}$$

(ii) From Eq. (1) and (2) $A_1 = \frac{2.0 \times 10^4}{2.0 \times 10^4} A_2$

$$A_1 = \frac{2.0 \times 10^4}{2.0 \times 10^4} A_2 = 2.0 \text{ cm}^2, \text{--- } R_1 = 2.0 \text{ cm}$$

EX-12 (i) From Eq. (1) and (2) $A_1 = A_2 \left(\frac{R_1}{R_2} \right)^2$

where, from Eq. (1) and (2), $R_1 = 2 \left[\frac{2.0 \times 10^4}{2.0 \times 10^4} \right] R_2$
 $= 2.0 \times 10^4 \text{ cm} = 2.0 \text{ cm}$

Using (i) $A_1 = 2.0 \times 10^4 \text{ cm}^2 = 2.0 \text{ cm}^2$

(ii) $A_1 = A_2 \left(\frac{R_1}{R_2} \right)^2 = 2.0 \text{ cm}^2$

$$\text{--- } R_1 = 2.0 \text{ cm}$$

EX-13



$$R \sin \theta = F \cos \theta$$

$$\text{or } R = F \frac{\cos \theta}{\sin \theta} = F \frac{2.0 \times 10^4 \text{ cm}}{2.0 \times 10^4 \text{ cm}}$$

In the horizontal
 direction

$$R \cos \theta = F \sin \theta \Rightarrow R = F \frac{\sin \theta}{\cos \theta}$$

$$= F \frac{2.0 \times 10^4 \text{ cm}}{2.0 \times 10^4 \text{ cm}} = 2.0 \times 10^4 \text{ cm}$$

where $R \cos \theta = \frac{2.0 \times 10^4 \text{ cm}}{2.0 \times 10^4 \text{ cm}}$

(i) $A_1 = A_2 \sin \theta, \quad A_2 = A_1 \cos \theta, \quad A_3 = A_1$

$$E = \int_0^{\pi} E \sin \theta d\theta = E_0 \int_0^{\pi} \left(\frac{d}{2} \cos \theta \right) \sin \theta d\theta$$

$$\text{In the far zone, } \frac{d}{2} \cos \theta = \frac{d}{2} \rightarrow E = E_0 \int_0^{\pi} \frac{d}{2} \sin \theta d\theta$$

$$E_{\text{far}} = E_0 \frac{d}{2} \int_0^{\pi} \sin \theta d\theta = E_0 \frac{d}{2} \left[-\cos \theta \right]_0^{\pi}$$

$$E_{\text{far}} = E_0 \frac{d}{2} (2) = E_0 d$$

4)



Radiation pattern
for $d \ll \lambda$
→ Plot of $|E_{\text{far}}|$

Ex 11.1 From Eq. (11-24a) $E_{\text{far}} = \frac{d}{4\pi r} (E_0 - E_0 \cos^2 \theta)$

a) $E_{\text{far}} = E_0 (E_0 + E_0 \cos^2 \theta) \sin^2 \theta$

From Eq. (11-24b) $|E_{\text{far}}| = E_0 \frac{d}{4\pi r} (1 + \cos^2 \theta) \sin^2 \theta$

b) For $E_{\text{far}} = E_0 (E_0 - E_0 \cos^2 \theta) \sin^2 \theta$

$$|E_{\text{far}}| = E_0 \frac{d}{4\pi r} (1 - \cos^2 \theta) \sin^2 \theta$$

c) For $E_{\text{far}} = E_0 E_0 \sin^2 \theta$, $|E_{\text{far}}| = E_0 \frac{d}{4\pi r} \sin^2 \theta$

Ex 11.2 Eq. (11-24a) $E = \frac{d \sin^2 \theta}{4\pi r} \left(\frac{E_0}{r} \right) [E_0 (1 + \cos^2 \theta) \sin^2 \theta]$

$$E = \frac{d}{4\pi} E_0 \frac{E_0}{r} \left(\frac{E_0 \sin^2 \theta}{r} \right) [1 + \cos^2 \theta] \sin^2 \theta$$

For circular polarization, $r = \frac{d}{2} \sin \theta$

a) $E = E_0 \frac{d}{4\pi} = E_0 \frac{d}{4\pi} \frac{E_0}{r} = E_0 \frac{d}{4\pi} \left(\frac{2}{d \sin \theta} \right) \sin^2 \theta$

$$= \frac{E_0 \sin \theta}{2\pi}$$

$$E = \frac{E_0 \sin^2 \theta}{2\pi} \text{ or } \sin \theta = \frac{E}{E_0} \text{ or } \theta = \sin^{-1} \left(\frac{E}{E_0} \right)$$

$$\therefore E_0 = \frac{E}{\sin \theta} = \frac{E}{\sin \left(\sin^{-1} \left(\frac{E}{E_0} \right) \right)} = \frac{E}{\frac{E}{E_0}} = E_0$$

b) $E_0 = \frac{E \sin \theta}{2\pi} = \frac{E \sin \left(\sin^{-1} \left(\frac{E}{E_0} \right) \right)}{2\pi} = \frac{E \left(\frac{E}{E_0} \right)}{2\pi} = \frac{E^2}{2\pi E_0}$

Ex-13 Assume $L_2(x, y) = k_2 y$.

$$\begin{aligned}
 E_{2,2}(t) &= \int_{x_0}^{x_1} \int_{y_0}^{y_1} e^{-(\lambda_1 + \lambda_2)t} \lambda_1 \lambda_2 L_2(x, y) dx dy \\
 &= \int_{x_0}^{x_1} \int_{y_0}^{y_1} e^{-(\lambda_1 + \lambda_2)t} \lambda_1 \lambda_2 k_2 y dx dy \\
 &= k_2 \lambda_1 \lambda_2 \int_{x_0}^{x_1} \left[\frac{e^{-(\lambda_1 + \lambda_2)t} y}{\lambda_2} \right]_{y_0}^{y_1} dx \\
 E_{2,2}(t) &= \frac{k_2 \lambda_1}{\lambda_2} e^{-(\lambda_1 + \lambda_2)t} \left[\frac{e^{-(\lambda_1 + \lambda_2)t} (y_1 - y_0)}{\lambda_2} \right]_{x_0}^{x_1} = k_2 (y_1 - y_0)
 \end{aligned}$$

Ex-14 From Ex-13, we have

$$E_{2,2}(t) = \int_{x_0}^{x_1} \int_{y_0}^{y_1} L_2(x, y) e^{-(\lambda_1 + \lambda_2)t} dx dy$$

(a) In the xy -plane, $\phi = t^{-1}$

$$\begin{aligned}
 E_{2,2}(t) &= k \int_{x_0}^{x_1} \int_{y_0}^{y_1} y e^{-(\lambda_1 + \lambda_2)t} dx dy \\
 &= k \int_{x_0}^{x_1} \left[\frac{e^{-(\lambda_1 + \lambda_2)t} y^2}{2} \right]_{y_0}^{y_1} dx \\
 &= \frac{k}{2} \frac{e^{-(\lambda_1 + \lambda_2)t} (y_1^2 - y_0^2)}{(\lambda_1 + \lambda_2)t} \quad \text{Let } \phi = \frac{1}{\lambda_1 + \lambda_2} \text{ then } \frac{d\phi}{dt} = -\phi^2
 \end{aligned}$$

$$E_{2,2}(t) = \frac{k}{2} \left[\frac{e^{-(\lambda_1 + \lambda_2)t}}{(\lambda_1 + \lambda_2)t} \right]$$

(b) Let $\left[\frac{e^{-(\lambda_1 + \lambda_2)t}}{(\lambda_1 + \lambda_2)t} \right] = \frac{1}{\phi} \implies \frac{d\phi}{dt} = -\phi^2$

Half-power intervals $(t_{1/2})_2 = t_{1/2}(t_{1/2})_1 \sqrt{2}$

For $\lambda_1 = \lambda_2 = \lambda$, $(t_{1/2})_2 = t_{1/2}(t_{1/2})_1 \sqrt{2}$
 $= t_{1/2} \sqrt{2} \log 2$

(c) Let $\frac{1}{\phi} = t \implies E_{2,2} = \frac{k}{2} \sqrt{2} \frac{1}{\phi} = \frac{k}{2} \sqrt{2} \log 2$
 $= t_{1/2} \sqrt{2} \log 2$

(d) First absolute error at $\phi = 1/2$, $t = t_{1/2}$

$$\text{where } E_{2,2} = \frac{k}{2} \left[\frac{1}{\sqrt{2}} \right]$$

\therefore Level of first absolute error, $E_{2,2} = \frac{k}{2} \sqrt{2} \log 2 = 0.49 \log 2$

Comparison of Results

	Uniform State	Triangular State
Asymptotic Frequency	$\ln\left(\frac{2000}{1000}\right)$	$\frac{2000}{1000} \ln\left(\frac{2000}{1000}\right)$
Half-power bandwidth	$20 \frac{1}{2}$ Hz	$7.14 \frac{1}{2}$ Hz
Location of first null	$40 \frac{1}{2}$ Hz	$14.28 \frac{1}{2}$ Hz
First-null-to-first-zero	21.2 dB	24.7 dB

EXERCISE 10 In the network, $\phi = 0^\circ$:

$$\begin{aligned} \hat{L}_{22}(s) &= 1 + \int_0^{2000} \cos\left(\frac{\omega t}{1000}\right) \cos(\phi) \cos(\omega t) dt \\ &= \int_0^{2000} \left[\frac{\cos\left(\frac{\omega t}{1000} + \omega t\right) + \cos\left(\frac{\omega t}{1000} - \omega t\right)}{2} \right] dt \quad \phi = 0^\circ \Rightarrow \cos \phi = 1 \end{aligned}$$

$$\text{d) Let } \frac{\left(\frac{\omega T}{2}\right)^2 \cos \phi}{\left(\frac{\omega T}{2}\right)^2 - \phi^2} = \frac{1}{\sqrt{2}} \quad \phi = 1.107 \text{ rad}$$

$$\text{Half-power bandwidth } (2.028)_{dB} = 20 \log_{10} \left(\frac{1}{\sqrt{2}} \right)$$

$$\text{For } \phi = 0 \text{ rad, } (2.028)_{dB} = 20 \log_{10} \left(\frac{1}{\sqrt{2}} \right) \text{ (rad)}$$

$$= 20 \log_{10} \left(\frac{1}{\sqrt{2}} \right) \text{ (rad)}$$

$$\text{e) Let } \phi = \frac{\pi}{4} \quad \phi = 0.785 \text{ rad} \quad \phi = 2\pi^{-1} \left(\frac{10}{1000} \right) = \frac{10}{1000} \text{ (rad)}$$

$$= 10 \frac{1}{1000} \text{ (rad)}$$

d) At first null, $\phi = \pi$ rad:

$$\hat{L}_{22} = -20 \log_{10} \frac{\left(\frac{\omega T}{2}\right)^2}{\left(\frac{\omega T}{2}\right)^2 - \pi^2} = 20 \log_{10} 17 = 24.7 \text{ dB}$$

	Uniform State	Constant State
Asymptotic Frequency	$\ln\left(\frac{2000}{1000}\right)$	$\frac{2000}{1000} \ln\left(\frac{2000}{1000}\right)$
Half-power bandwidth	$20 \frac{1}{2}$ Hz	$42.1 \frac{1}{2}$ Hz
Location of first null	$40 \frac{1}{2}$ Hz	$84.2 \frac{1}{2}$ Hz
First-null-to-first-zero	21.2 dB	22.7 dB

The following corrections should be made to PHYSICS and PHYSICS LABORATORY by David E. Gray. An explanation for any inconsistency that may result.

INDEX, 1951

Table of New Corrections by David E. Gray (July, 1951)

- P. 114, 1st paragraph, 2nd line: ~~Change~~ → ~~Change~~.
- P. 1, Eq. (1-10): Add the first square under the first square—see in Eq. (1-10).
- P. 11, Fig. 1-1: The dashed lines for the bottom part of the diagram are ~~not~~ ~~not~~. The dot for \vec{p}_1 should be put at the center of the bottom line. The dot for \vec{p}_2 should be put at the center of the upper line.
- P. 41, Eq. (1-10): $\vec{p}_1 \rightarrow -\vec{p}_1$
- P. 75, problem 1-1-10: $\Delta x_1^2 \rightarrow \Delta x_2^2$.
- P. 111, Fig. 1-11: Add a plus sign.
- P. 141, problem 1-1-11: ~~Change~~ → ~~Change~~.
- P. 151, Eq. (1-11): In the denominator: $\rightarrow \rightarrow$.
- P. 151, problem 1-1, 1st line: ~~add~~ ~~add~~ \vec{p}_1 and \vec{p}_2 before \vec{p}_1 .
- P. 151, problem 1-1: (1-11) → (1-11).
- P. 151, Eq. (1-11): ~~change~~ ~~change~~ \vec{p}_1 after the \vec{p}_1 sign.
- P. 151, table 1-1, 1st line: the letter E in the 1st row.
- P. 211, Eq. (2-11): ~~change~~ ~~change~~ $\frac{1}{2} \frac{d^2x}{dt^2}$ for $\frac{1}{2} \frac{d^2x}{dt^2}$ in the first term: $\frac{1}{2} \frac{d^2x}{dt^2}$.
- P. 211, line 1: ~~change~~ ~~change~~ $\frac{1}{2} \frac{d^2x}{dt^2}$ after the word $\frac{1}{2} \frac{d^2x}{dt^2}$.
- P. 211, problem 2-1-11: ~~change~~ ~~change~~ $\frac{1}{2} \frac{d^2x}{dt^2}$ before the word $\frac{1}{2}$.
- P. 211, problem 2-1-11, 11: $1/2 \rightarrow 1/2$.
- P. 211, problem 2-1: ~~line~~ ~~line~~ $\frac{1}{2} \frac{d^2x}{dt^2}$ the word $\frac{1}{2}$ is to be ~~added~~ ~~added~~ $\frac{1}{2} \frac{d^2x}{dt^2}$ and a new $\frac{1}{2}$ should be ~~added~~ ~~added~~.
- P. 211, line 1: $\frac{1}{2} \frac{d^2x}{dt^2} \rightarrow \frac{1}{2} \frac{d^2x}{dt^2}$ problem 2-1-11. ~~line~~ ~~line~~ $\frac{1}{2} \frac{d^2x}{dt^2}$ → $\frac{1}{2} \frac{d^2x}{dt^2}$.
- P. 211, Fig. 2-11: ~~change~~ ~~change~~ the segment between E and \vec{p}_1 —(see Fig. 2-11, p. 211).
- P. 211, line 1: ~~line~~ ~~line~~ $\frac{1}{2} \frac{d^2x}{dt^2}$ → Fig. 2-11.
- P. 211, problem 2-1-11: ~~line~~ ~~line~~ $\frac{1}{2} \frac{d^2x}{dt^2}$ → $\frac{1}{2} \frac{d^2x}{dt^2}$ in Fig. 2-11.
- P. 211, Fig. 2-11: ~~change~~ ~~change~~ $\frac{1}{2} \frac{d^2x}{dt^2}$ before $\frac{1}{2} \frac{d^2x}{dt^2}$ and $\frac{1}{2}$ in $\frac{1}{2} \frac{d^2x}{dt^2}$ before $\frac{1}{2} \frac{d^2x}{dt^2}$.

P. 224, eq. (20-11): $\vec{E} \rightarrow \vec{E}'$ (2 added); P. 224-225: $\vec{E} \rightarrow \vec{E}'$

P. 224, problem 20-44: last line Change "to the field" to "to";
and last before "axis of support".

P. 224, problem 20-44(a): Change "vector" to "moment".

P. 224, paragraph 4, line 4: $\vec{E} \rightarrow \vec{E}'$

P. 224, last. sentence, line 2: 20-223 \rightarrow 20-224.

P. 224, line 10: $\vec{E} \rightarrow \vec{E}'$

P. 224, problem 20-47 (a): $\vec{E} \rightarrow \vec{E}'$ (more space)

P. 224, problem 20-47 (b): 20-47 \rightarrow 20-48.

P. 224, problem 20-48: 20-4 \rightarrow 20-48.

Book references: table $\vec{E} \rightarrow \vec{E}'$ (see eq. 20-11, p. 22 and eq. 20-12, p. 224)
Energy density

Request to check the explanation when answers are 400

20-2, Fig. 1-1: Molecular lines should be centered on the vertical.

20-4, Fig. 1-4

20-6, Fig. 20-6: \vec{E} should be on the vertical.

20-8, Fig. 2-8

20-9, Fig. 2-9: \vec{E} should be on the vertical.

20-10, Fig. 2-10: \vec{E} and \vec{E}' should be vertical; the arrow should be
drawn up to touch \vec{E} ; the point on \vec{E} should be $\vec{E} \cdot \vec{E}'$ (see Fig. 2-14 on p. 224).

20-11, Fig. 2-11 (the vertical) vertical line should coincide with the $\vec{E} \cdot \vec{E}'$
line. (see Fig. 2-14 on p. 224).

20-12, Fig. 2-12: The arrow should pass through the centers of the circles.

20-13, Fig. 2-13: 20-13, Fig. 2-14: 20-13, Fig. 20-15: 20-13, Fig. 20-16.